

Digitally-Assisted Compensation for Timing Skew in ATE Systems

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Advantest Corporation

Gunma University



Contents

- Research Goal
- Conventional Linear Phase Digital Filter Condition
- New Linear Phase Digital Filter Condition
 - Time-Shift, Impulse Response of Ideal Filter
 - New Linear Phase Digital Filter
- MATLAB Simulation
- Design Considerations
 - Window
 - Gain Adjustment
- Application
- Conclusion



Research Goal

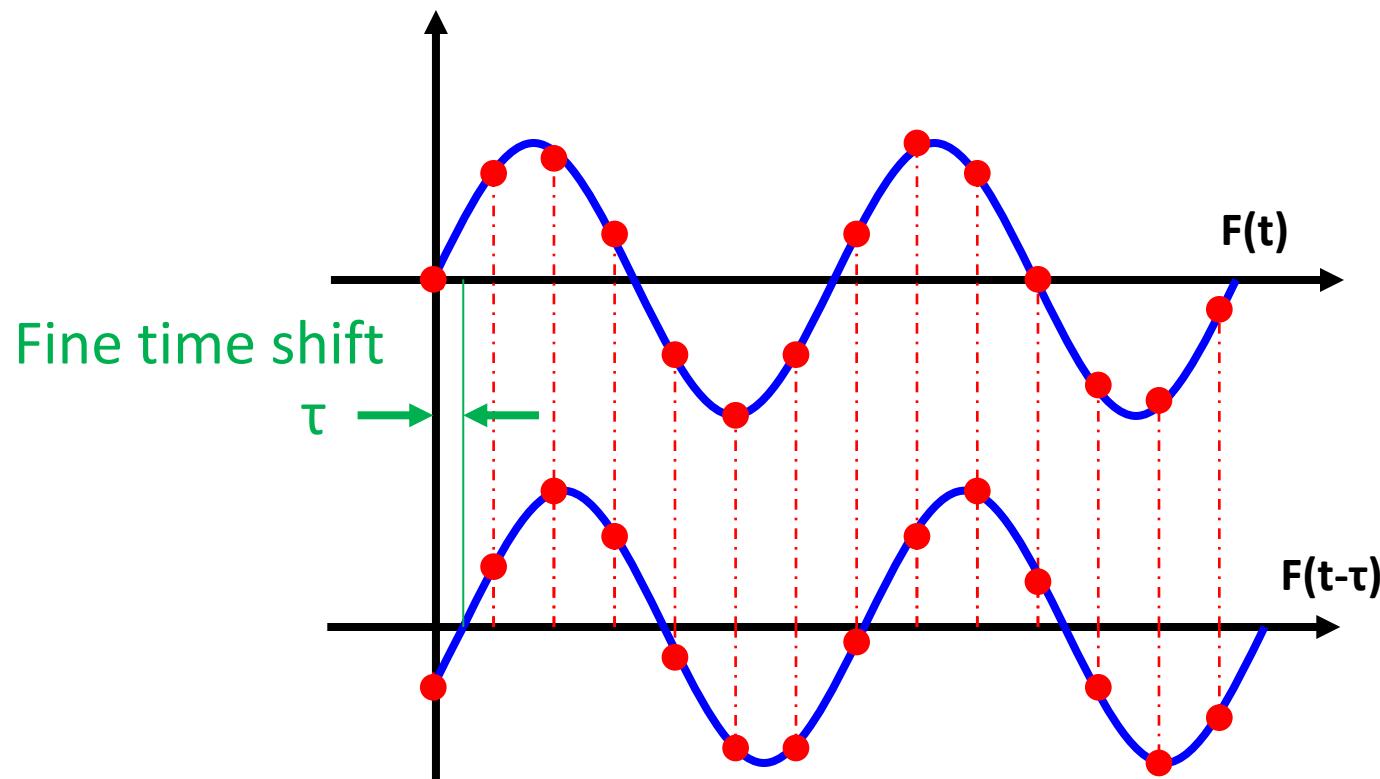
Timing skew is a major problem in ATE systems

Digital compensation for timing skew
⇒ Linear phase is important

Conventional linear-phase digital filter ⇒ coarse timing adjustment

Proposed linear-phase digital filter ⇒ fine timing adjustment

Features of Proposed Digital Filter



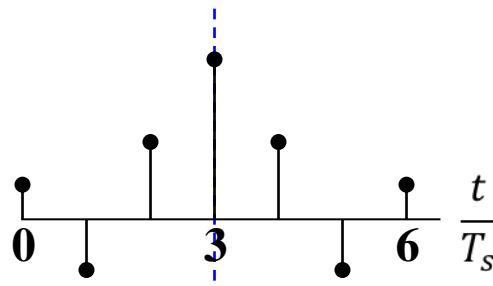
- Fine time resolution
- Linear phase
- Applicable to bandpass signal

Contents

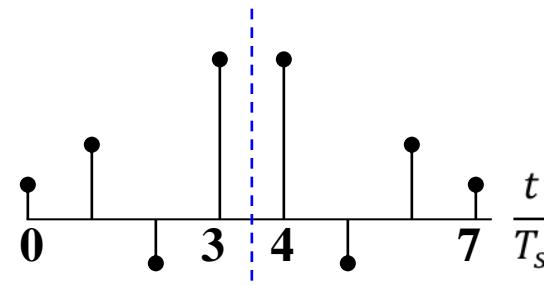
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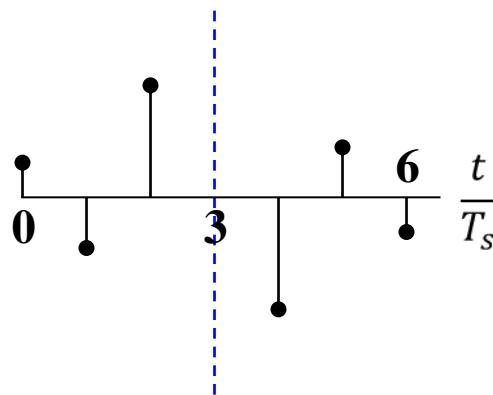
Linear Phase FIR Filter Impulse Response



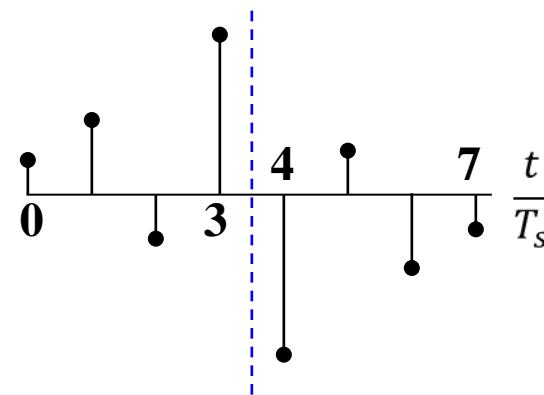
(1) Case 1
odd # of taps · even symmetry



(2) Case 2
even # of taps · even symmetry



(3) Case 3
odd # of taps · odd symmetry



(4) Case 4
even # of taps · odd symmetry

Frequency Characteristics

$h(nT)$	$H(e^{j\omega T})$
Case 1	$e^{-j\omega(N-1)T_s/2} \sum_{k=0}^{(N-1)/2} a_k \cos[\omega k T_s]$
Case 2	$e^{-j\omega(N-1)T_s/2} \sum_{k=1}^{N/2} b_k \cos[\omega(k - 1/2) T_s]$
Case 3	$e^{-j(\omega(N-1)T_s/2-\pi/2)} \sum_{k=0}^{(N-1)/2} a_k \sin[\omega k T_s]$
Case 4	$e^{-j(\omega(N-1)T_s/2-\pi/2)} \sum_{k=1}^{N/2} b_k \sin[\omega(k - 1/2) T_s]$

Phase : proportional to ω (linear phase)

Time resolution of group delay $T_s/2$

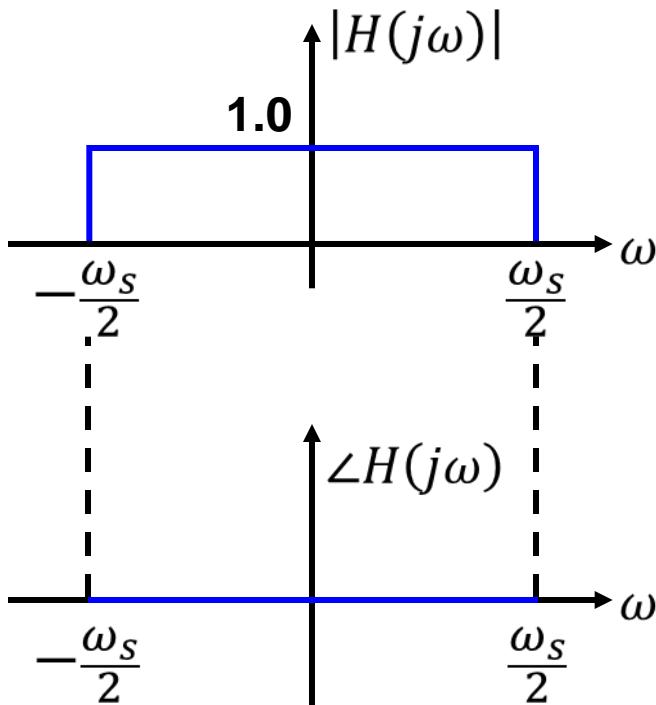
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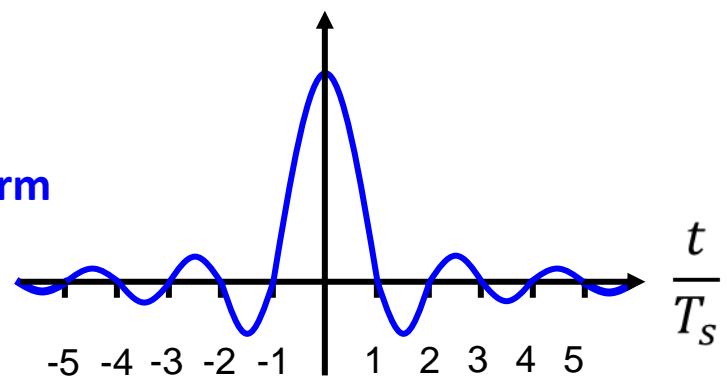


Ideal LPF

Frequency Characteristics



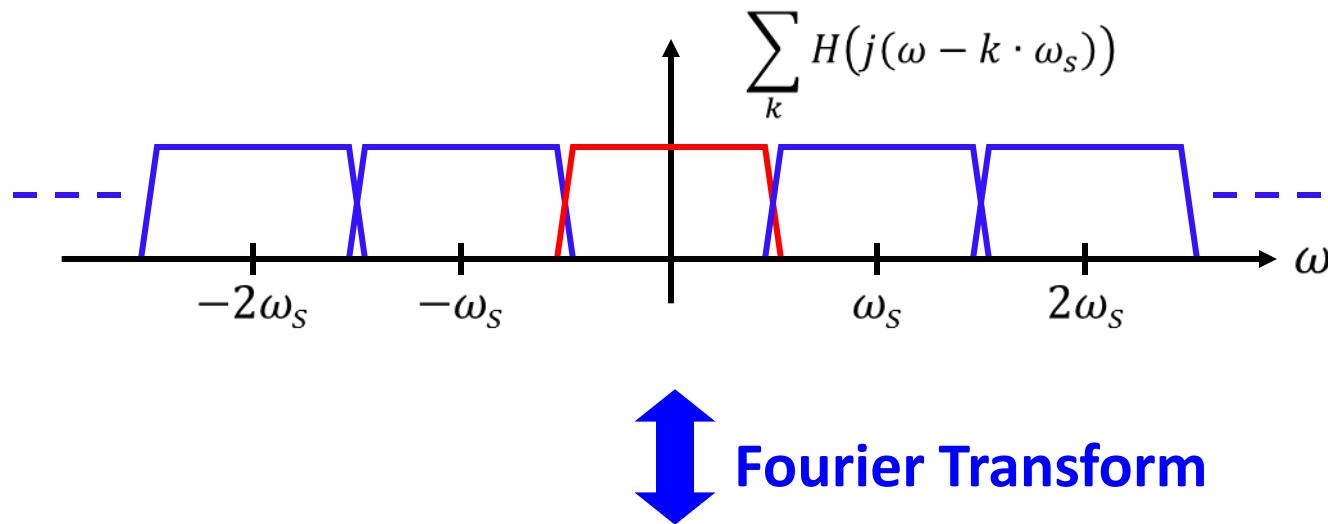
Impulse Response



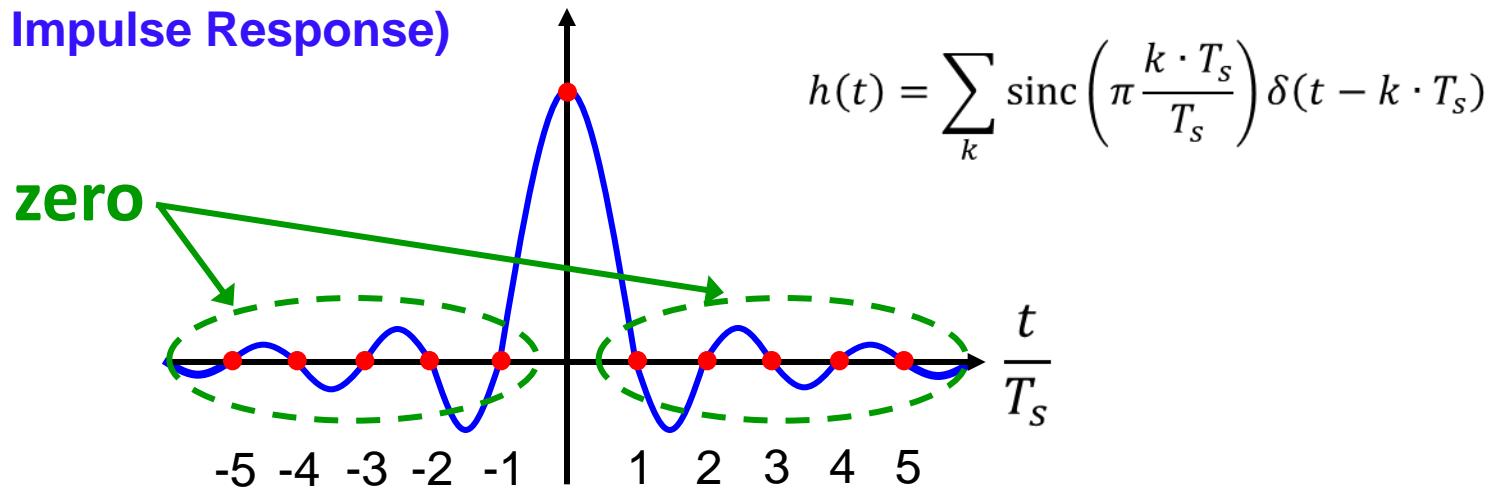
$$h(t) = \frac{1}{T_s} \operatorname{sinc}\left(\pi \frac{t}{T_s}\right)$$

$$\omega_s = \frac{2\pi}{T_s} : \text{Sampling Frequency}$$

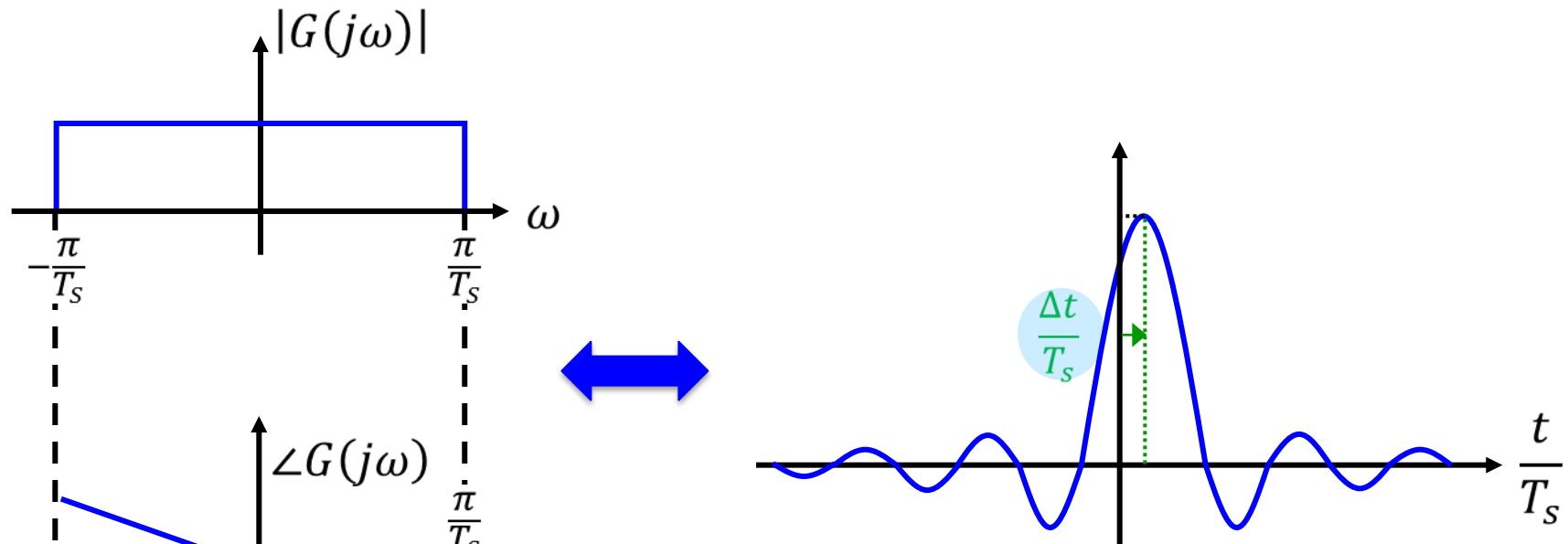
Discrete-Time Representation of Ideal LPF



FIR (Finite Impulse Response)



Impulse Response Time-Shift

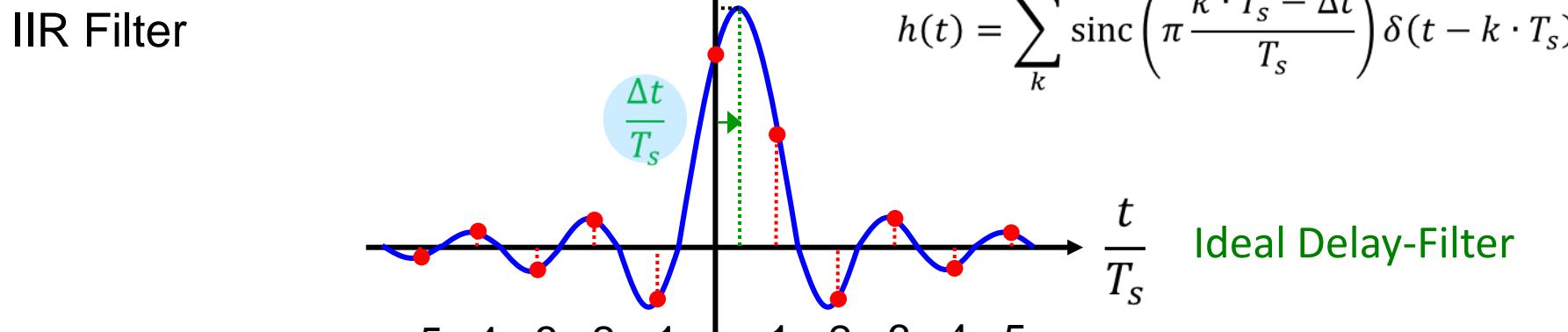
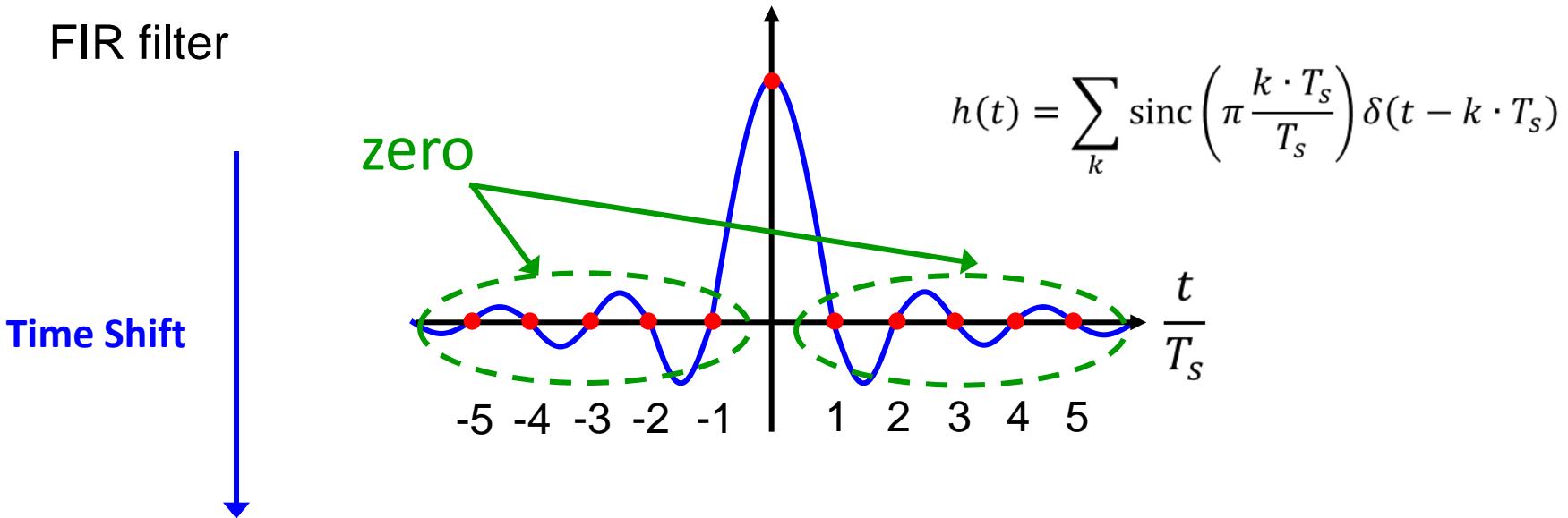


$$\angle G(j\omega) = -\omega \Delta t$$

Δt time-shift of impulse response

No change of Gain

Time-Shift and Filter Coefficients

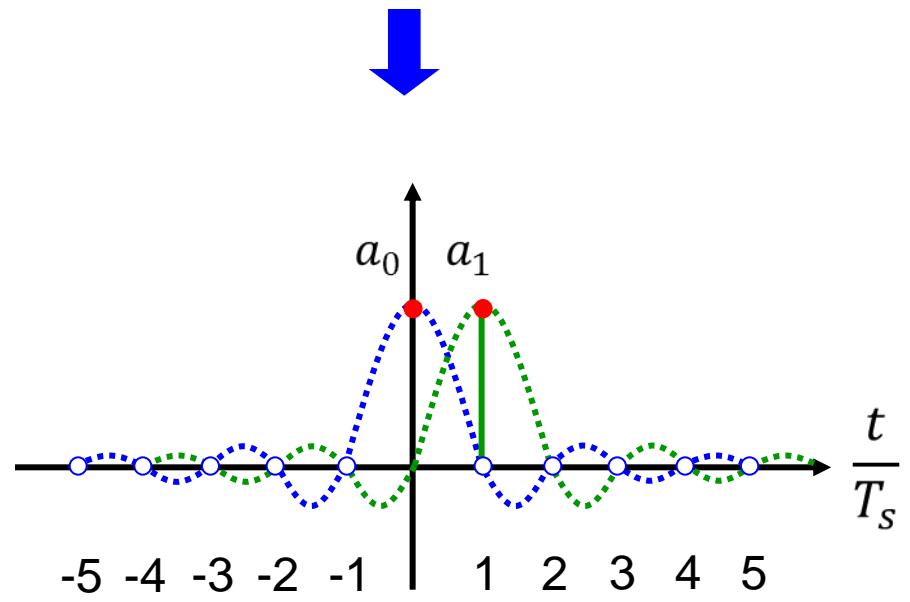
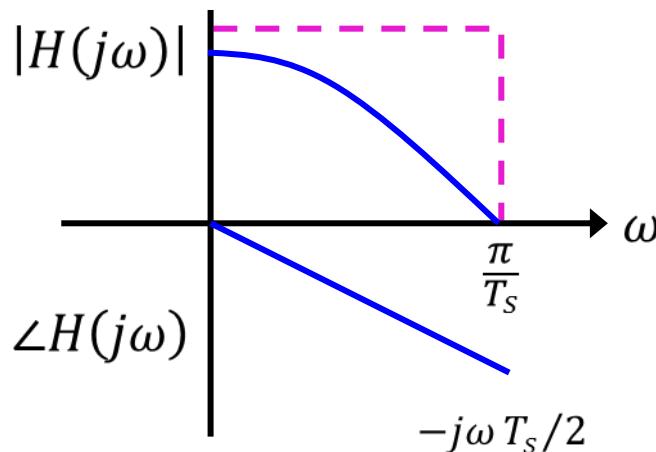
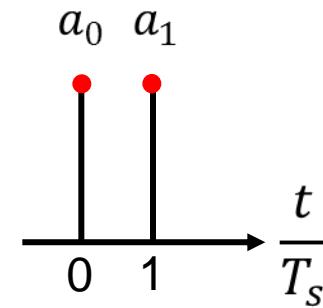
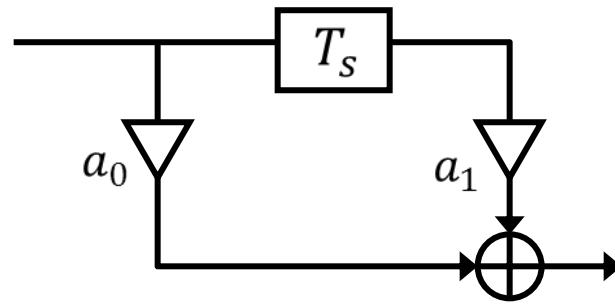


Contents

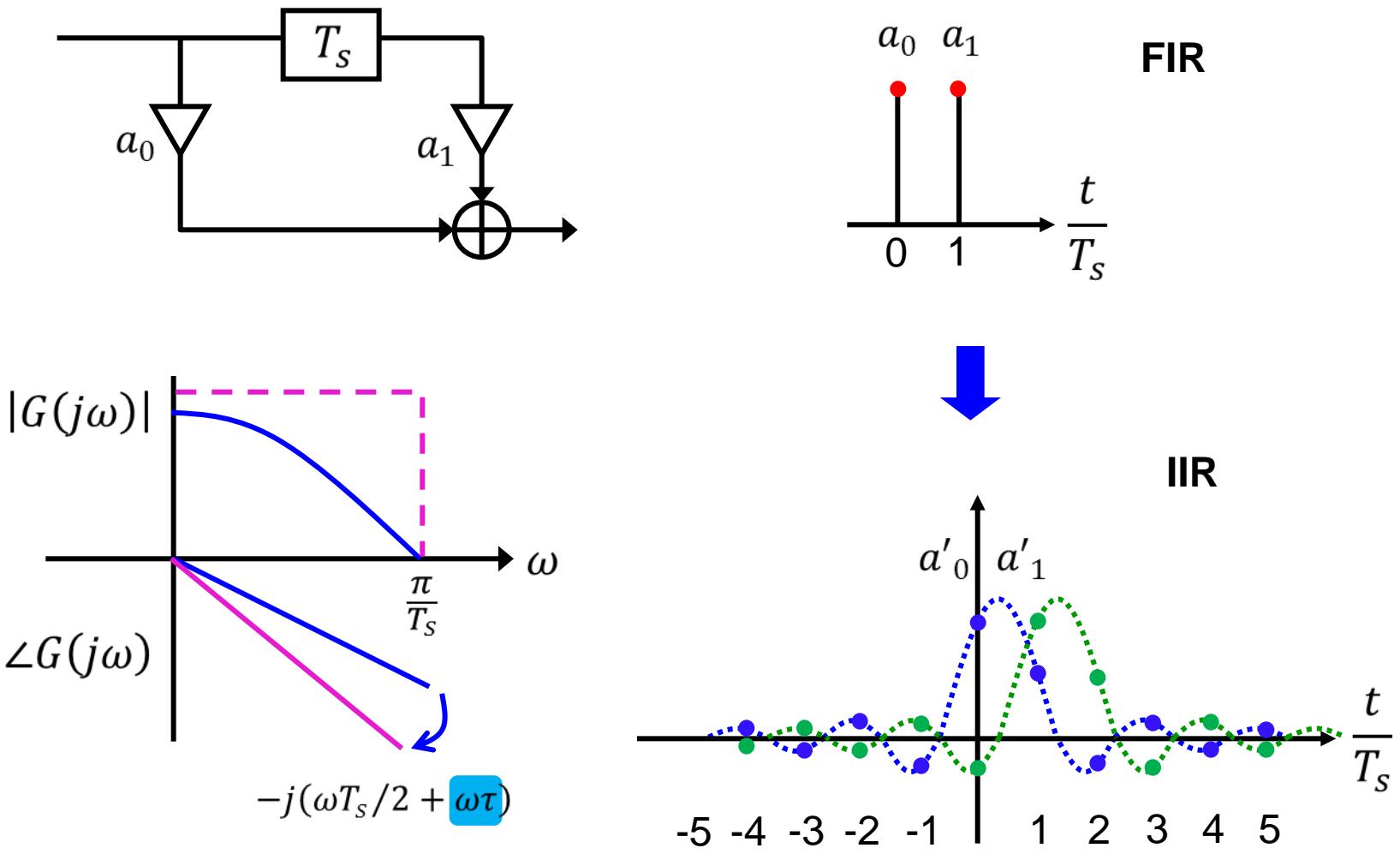
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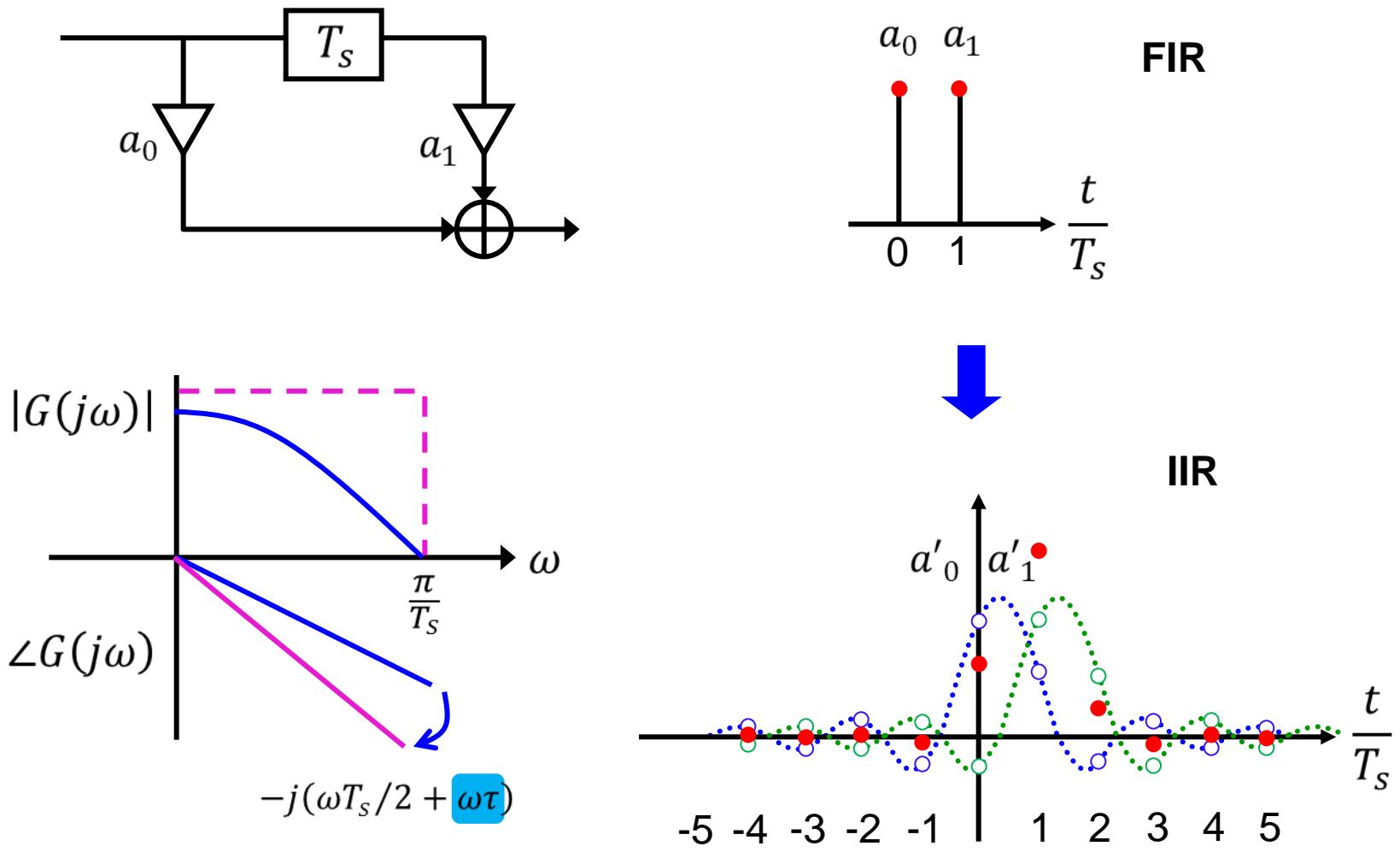
2-Tap Filter: Model



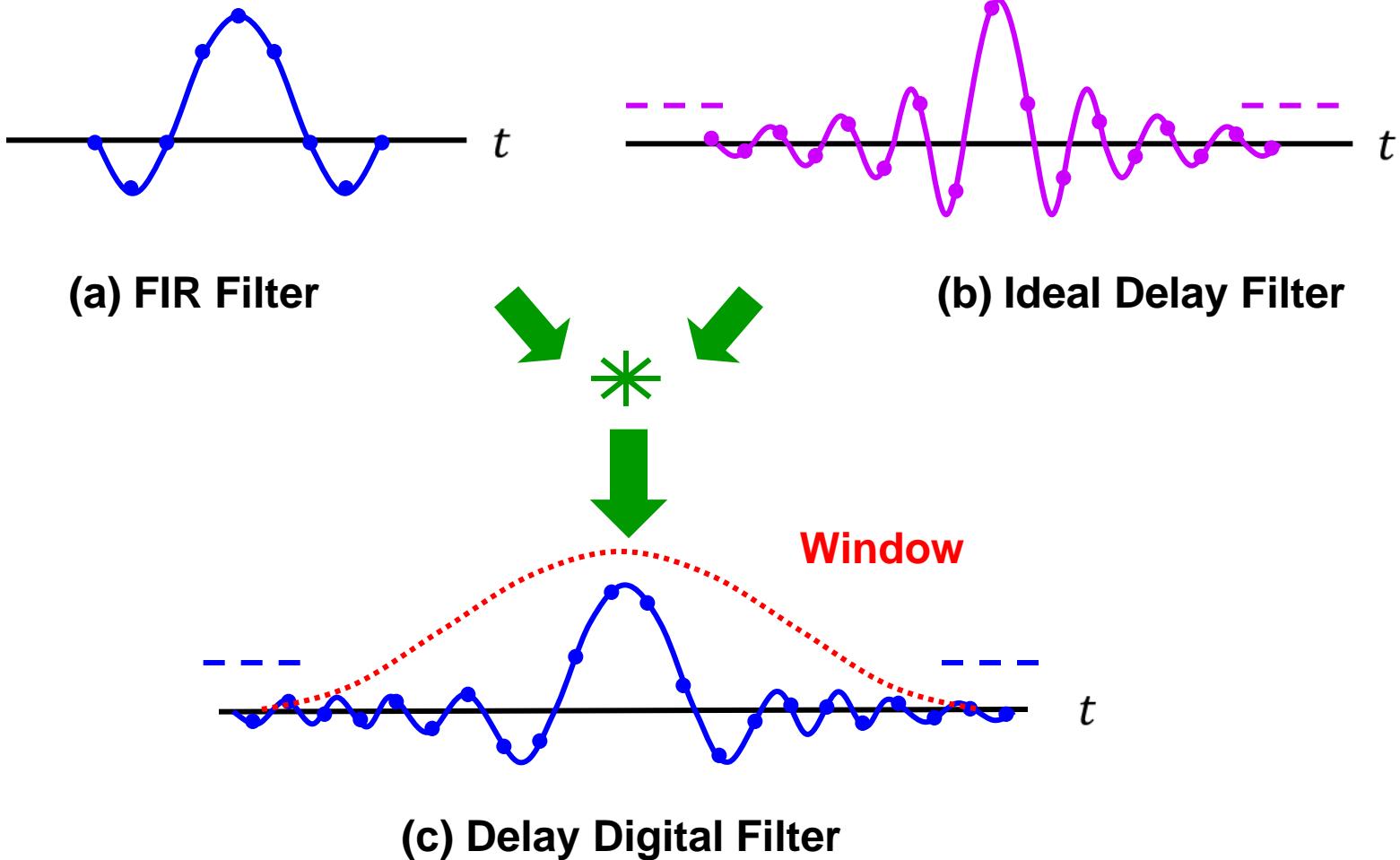
2-Tap Filter: Delay Model



2-Tap Filter: Delay Model



Proposed Delay Digital Filter



Frequency Characteristics of Proposed Delay Digital Filter

$g(nT)$	$G(e^{j\omega T})$
Case 1	$e^{-j\omega(N-1)T_s/2+\omega\tau} \sum_{k=0}^{(N-1)/2} a_k \cos[\omega k T_s]$
Case 2	$e^{-j\omega(N-1)T_s/2+\omega\tau} \sum_{k=1}^{N/2} b_k \cos[\omega(k - 1/2) T_s]$
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Case 4	$e^{-j(\omega(N-1)T_s/2-\pi/2+\omega\tau)} \sum_{k=1}^{N/2} b_k \sin[\omega(k - 1/2) T_s]$

Phase : proportional to ω (linear phase)

Group delay time resolution τ : Arbitrary small

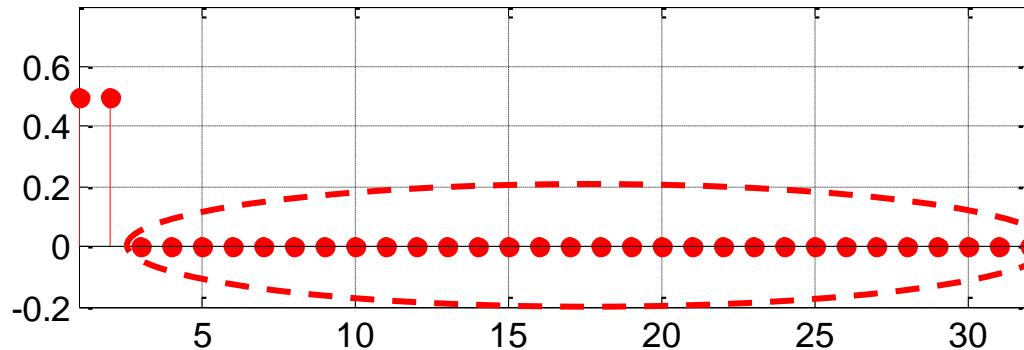
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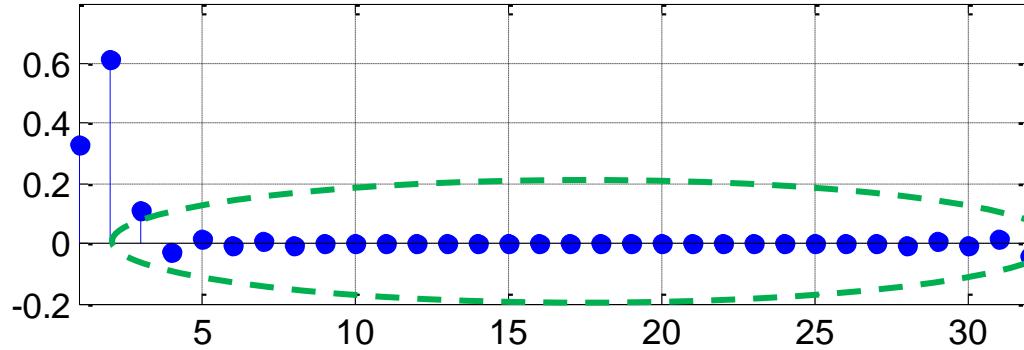
Comparison of 2-Tap Filter Impulse Responses

2-Tap FIR Filter



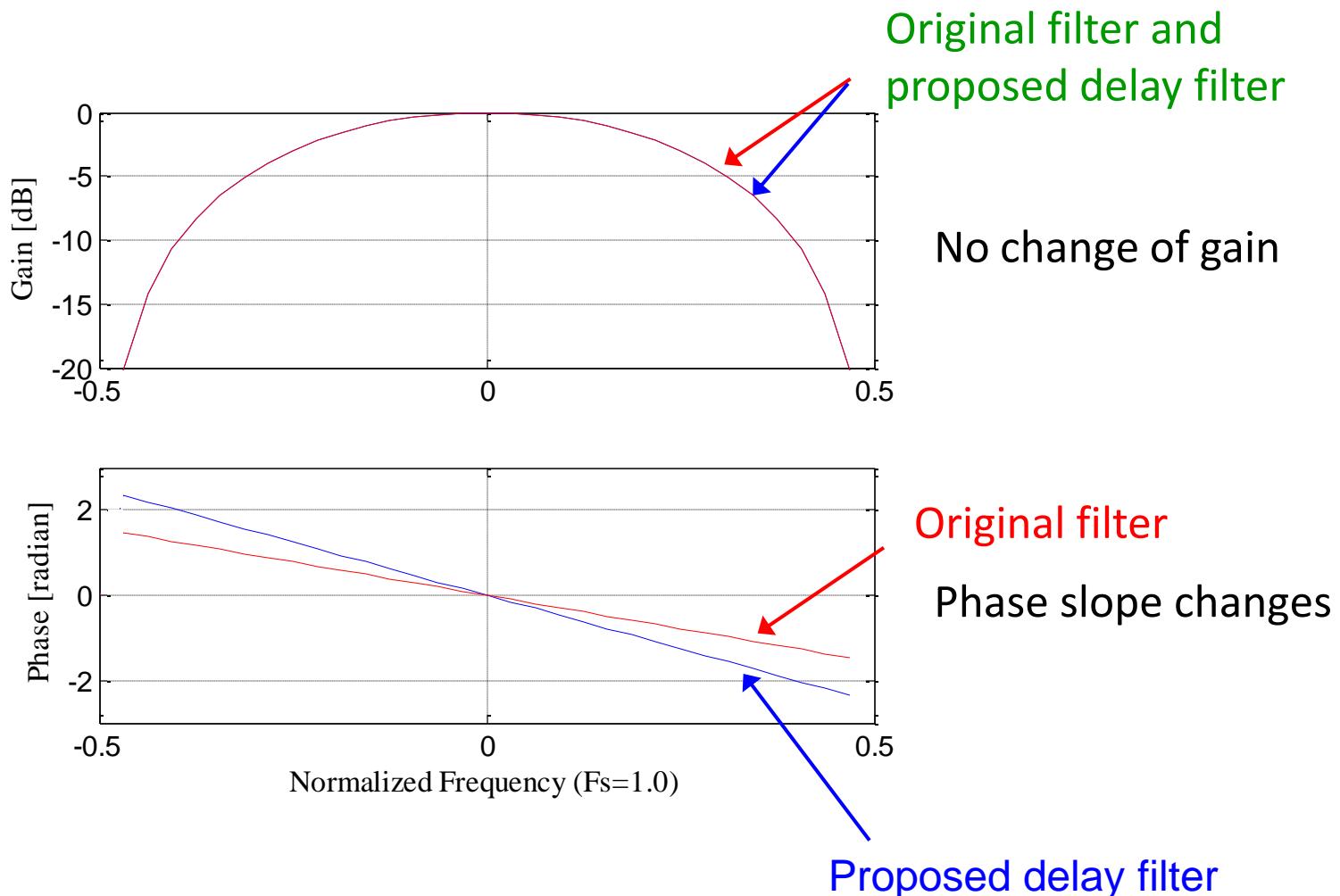
2-tap FIR coefficients
zero

Proposed Delay Filter (0.3 samples delay)



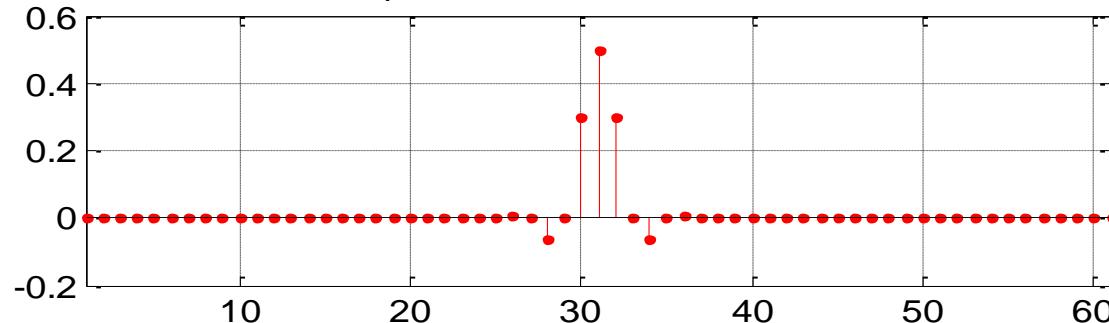
Impulse response changes.
Non-zero

Comparison of 2-Tap Filter Frequency Characteristics

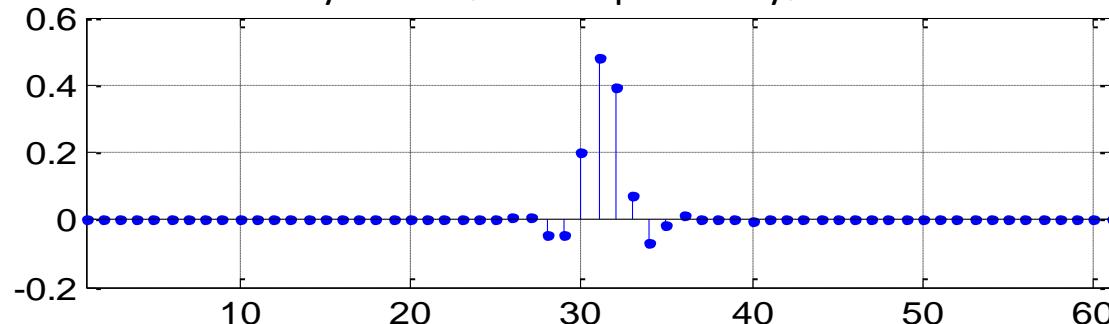


Finite Tap Truncation of Proposed Delay Filter

61-Tap Cosine Roll-off Filter

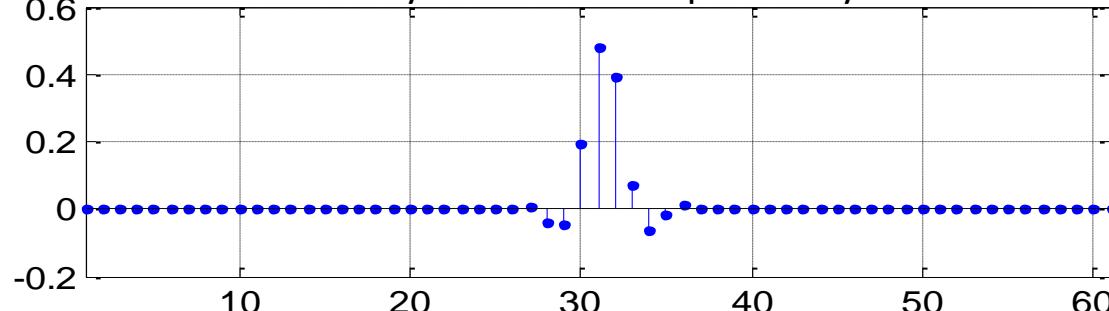


Delay Filter (0.3 samples delay)



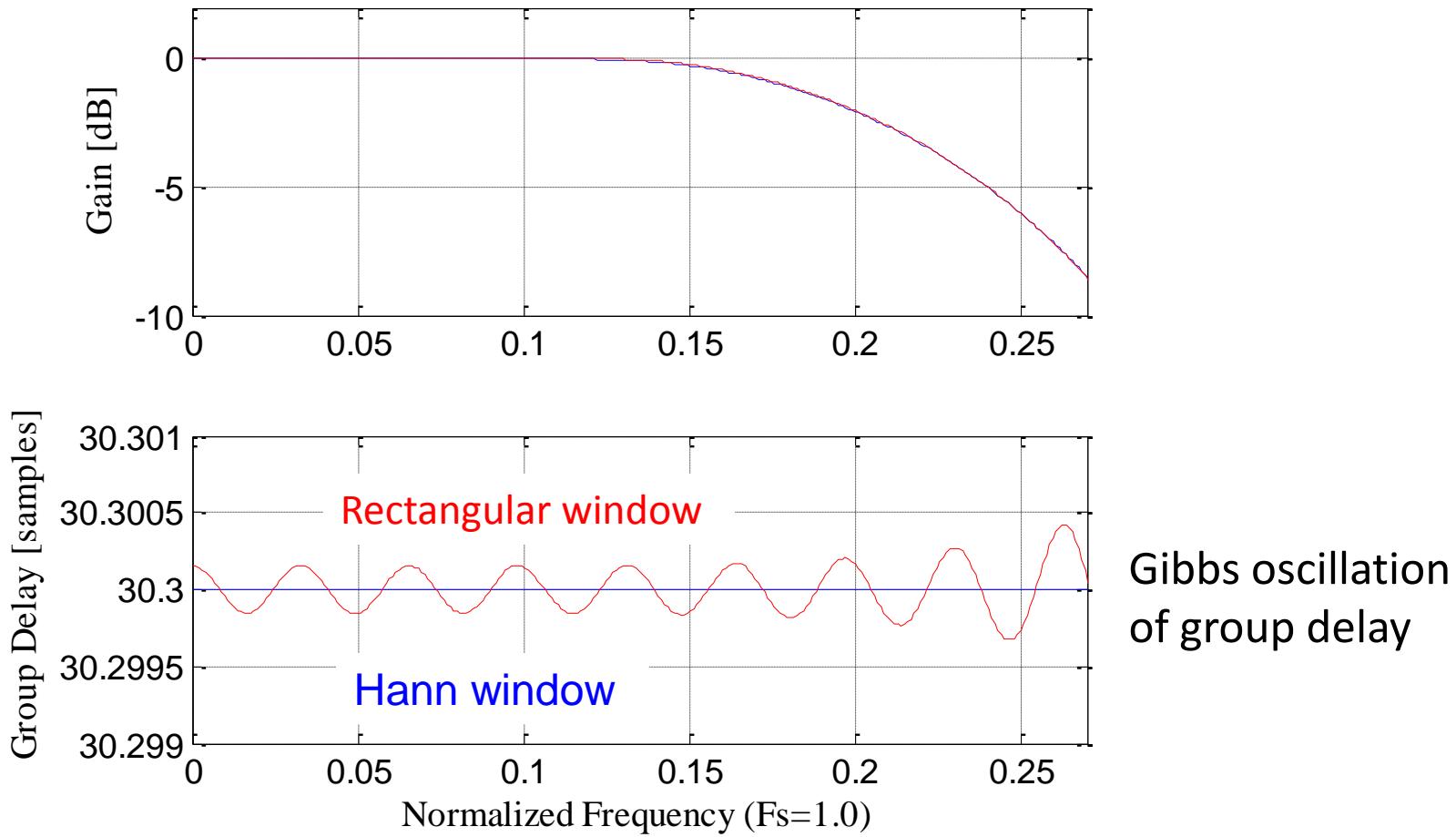
Rectangular
window

Delay Filter (0.3 samples delay)



Hann
window

Effects of Window



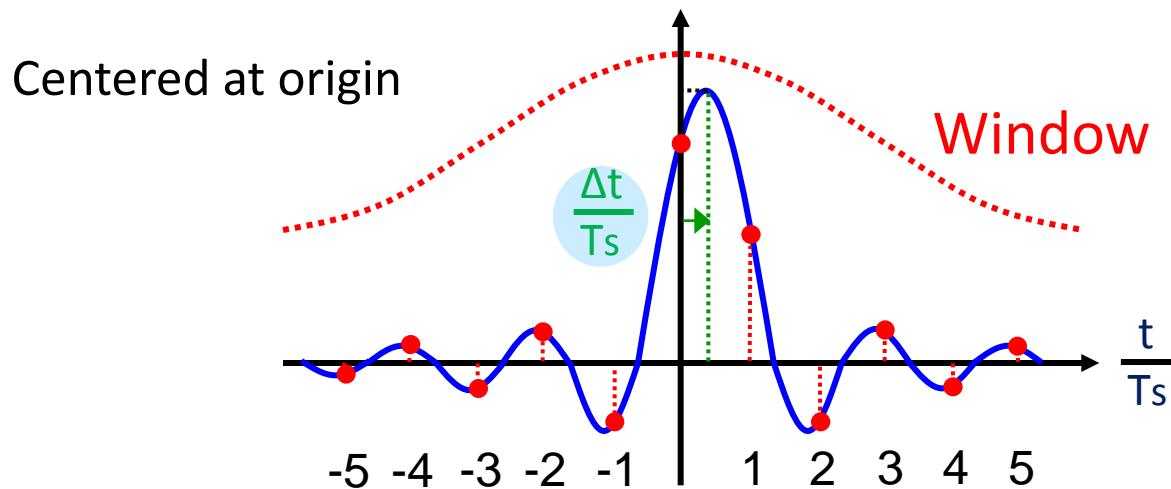
Frequency characteristics of delay filter with 61-tap truncation

Contents

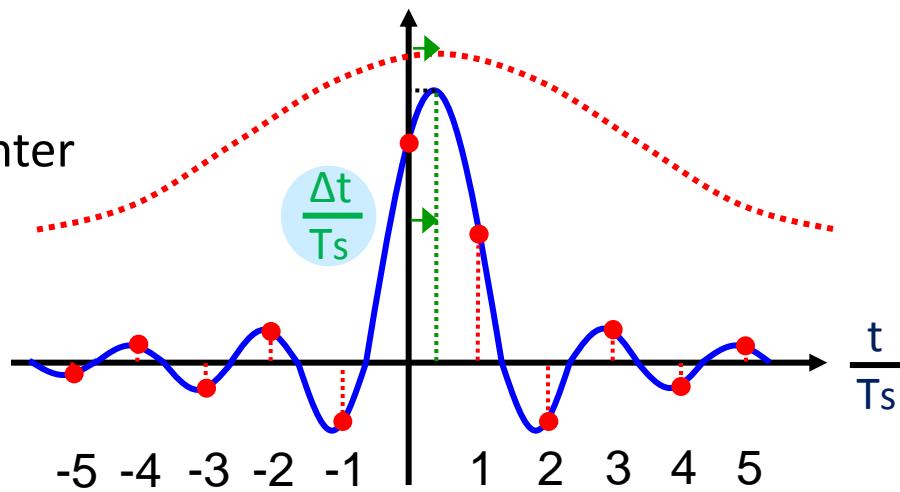
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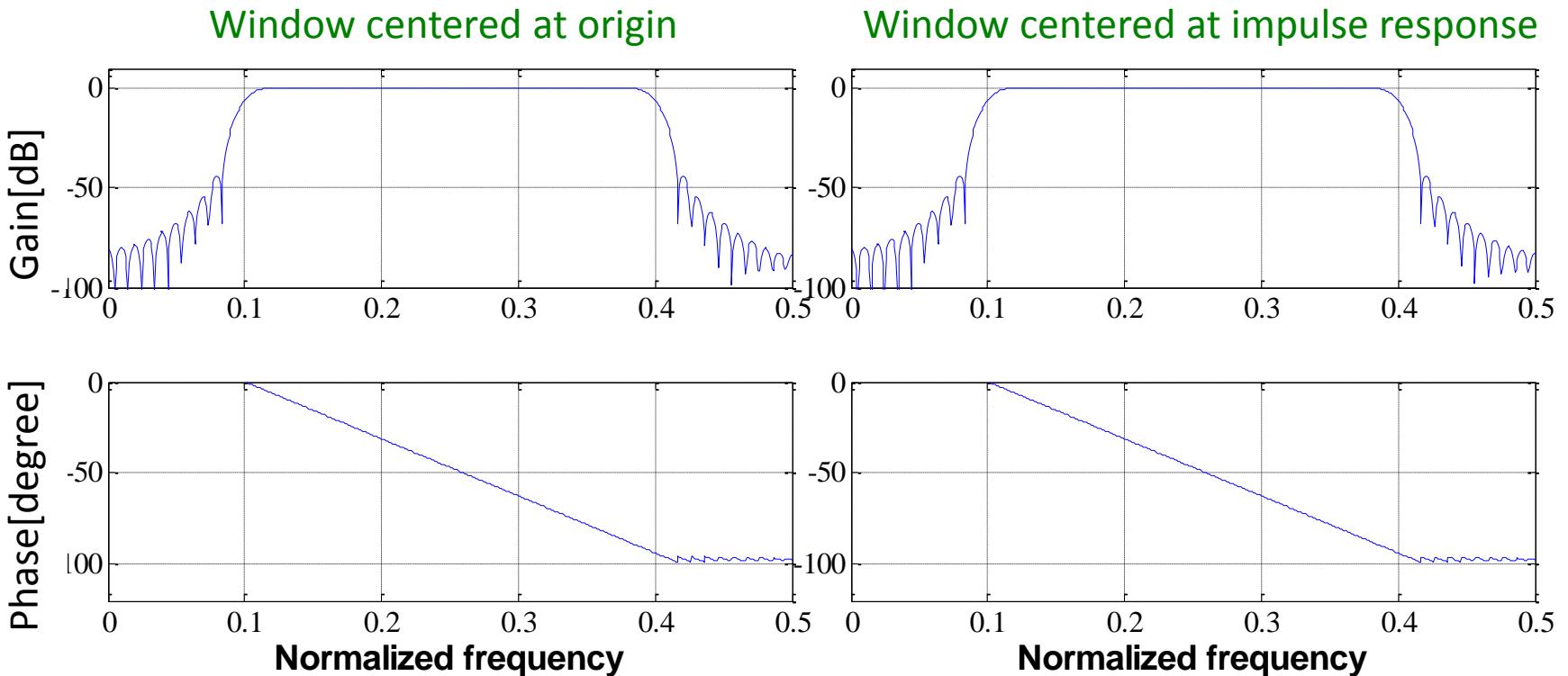
How to Apply Window



Centered at
impulse response center

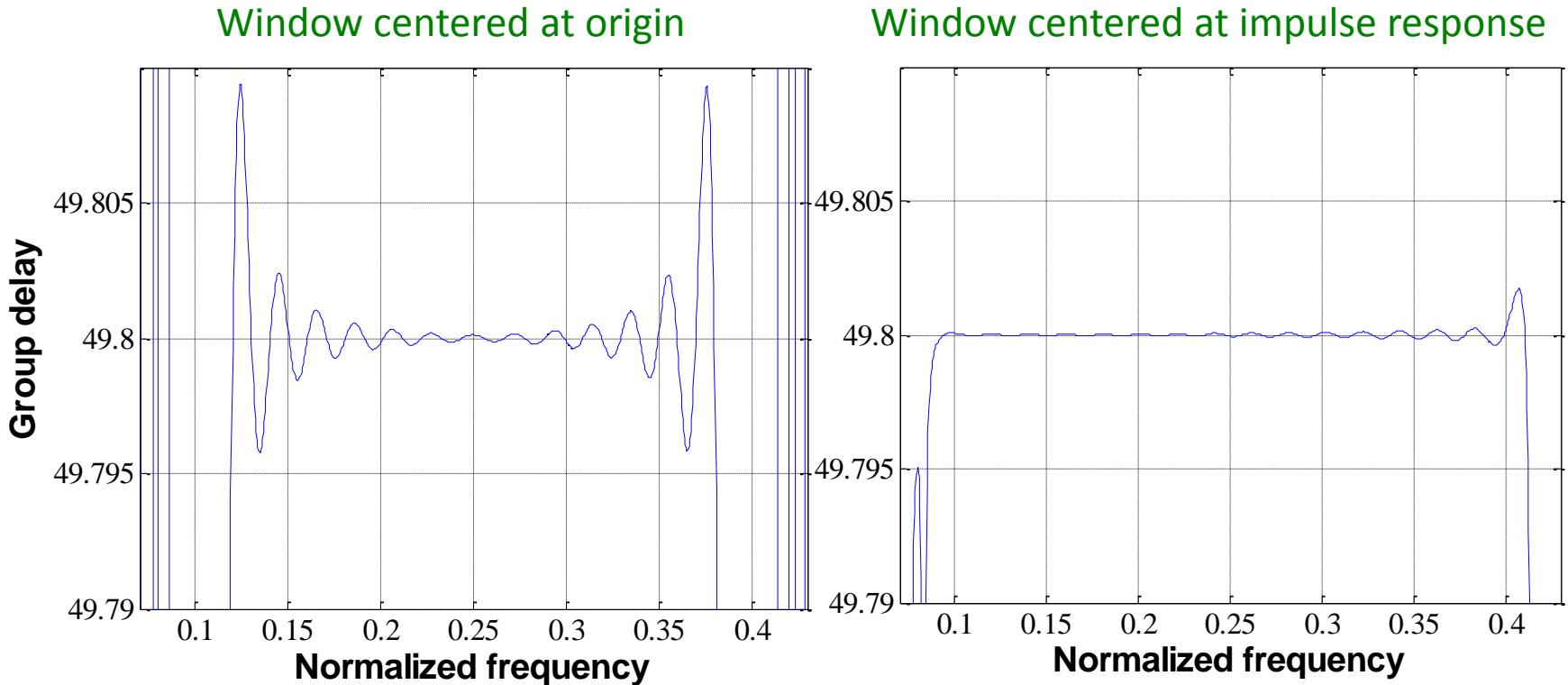


Frequency Characteristics of Delay Filter after Applying Window



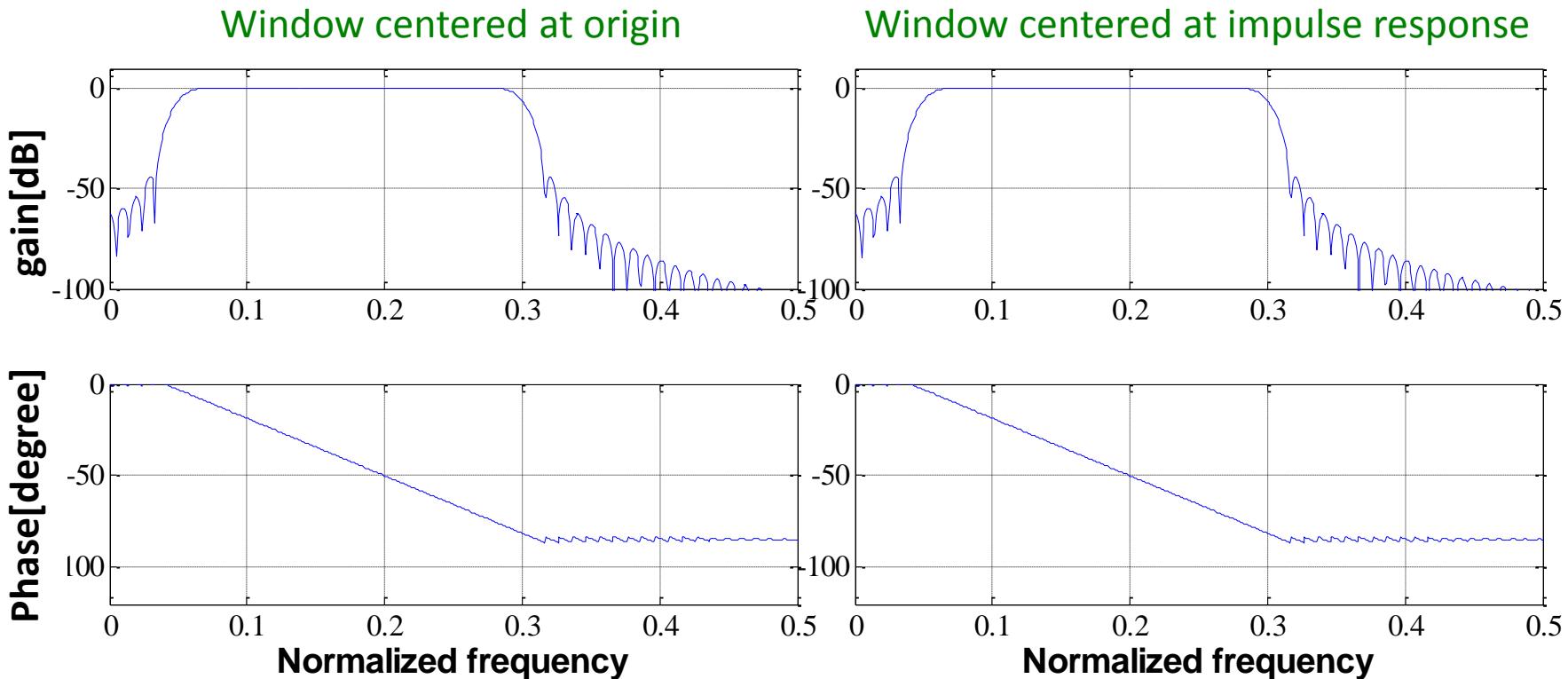
Delay	0.3 samples
Filter Tap	100 taps
Window	Han
Pass band	$(0.1 \sim 0.4) \cdot F_s$
FFT points	1024 points

Group Delay Characteristics of Delay Filter after Applying Window



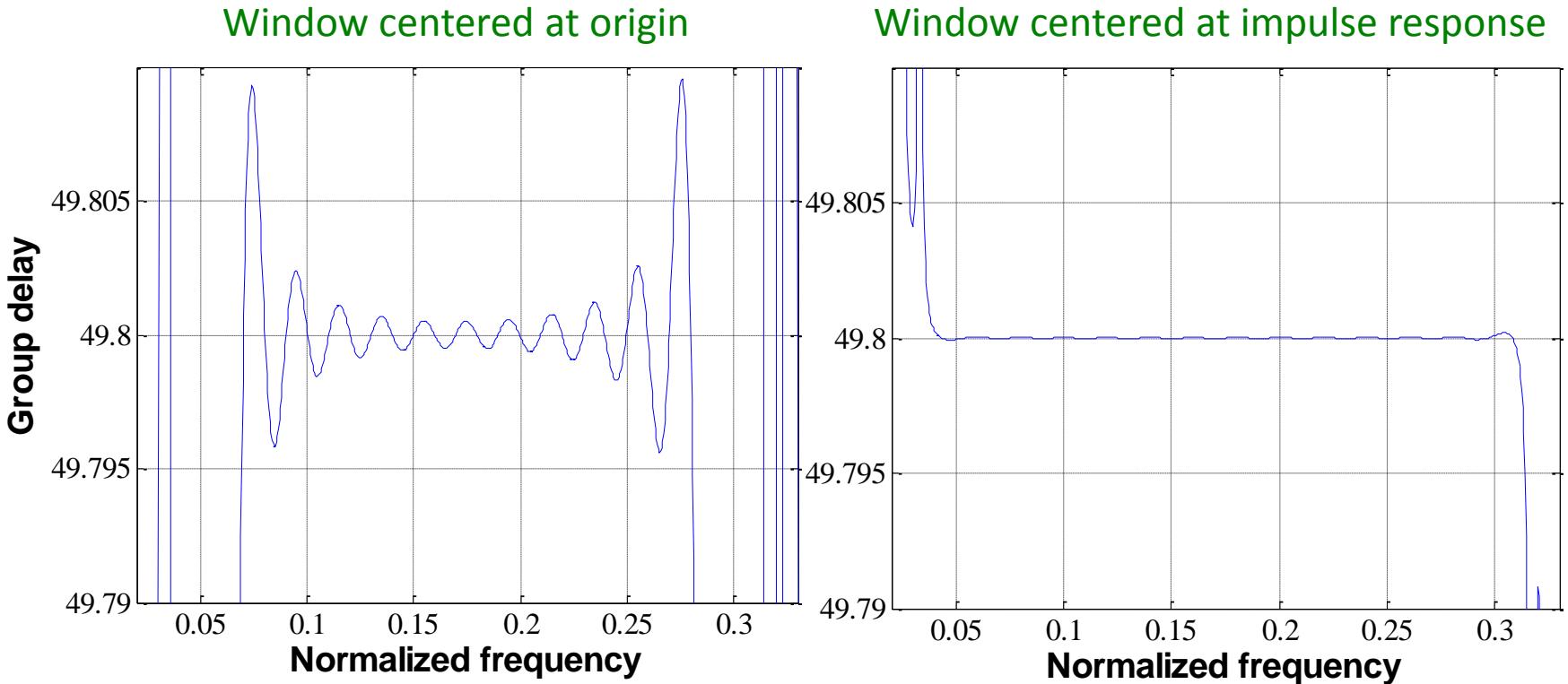
Delay	0.3 samples
Filter Tap	100 taps
Window	Han
Pass band	$(0.1 \sim 0.4) \cdot F_s$
FFT points	1024 points

Frequency Characteristics of Delay Filter after Applying Window



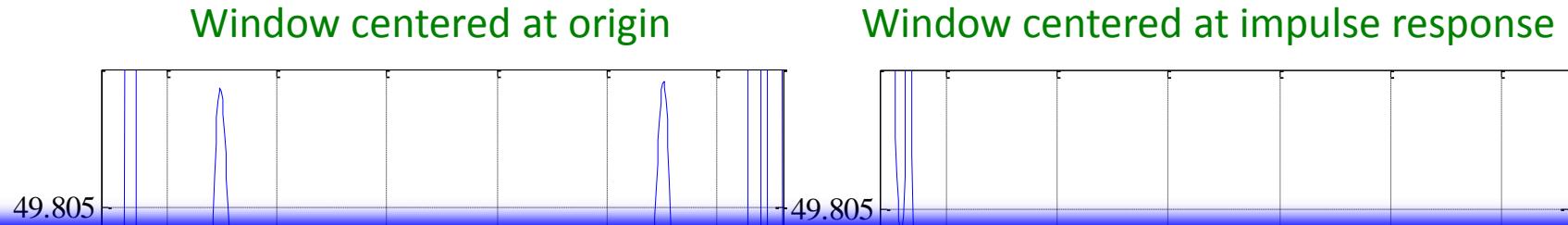
Delay	0.3 samples
Filter Tap	100 taps
Window	Han
Pass band	(0.05~0.3)·Fs
FFT points	1024 points

Group Delay Characteristics of Delay Filter after Applying Window



Delay	0.3 samples
Filter Tap	100 taps
Window	Han
Pass band	(0.05~0.3)·Fs
FFT points	1024 points

Group Delay Characteristics of Delay Filter after Applying Window



Applying window centered at impulse response



Constant group delay over entire passband

Normalized frequency

Normalized frequency

Delay	0.3 samples
Filter Tap	100 taps
Window	Han
Pass band	(0.05~0.3)·Fs
FFT points	1024 points

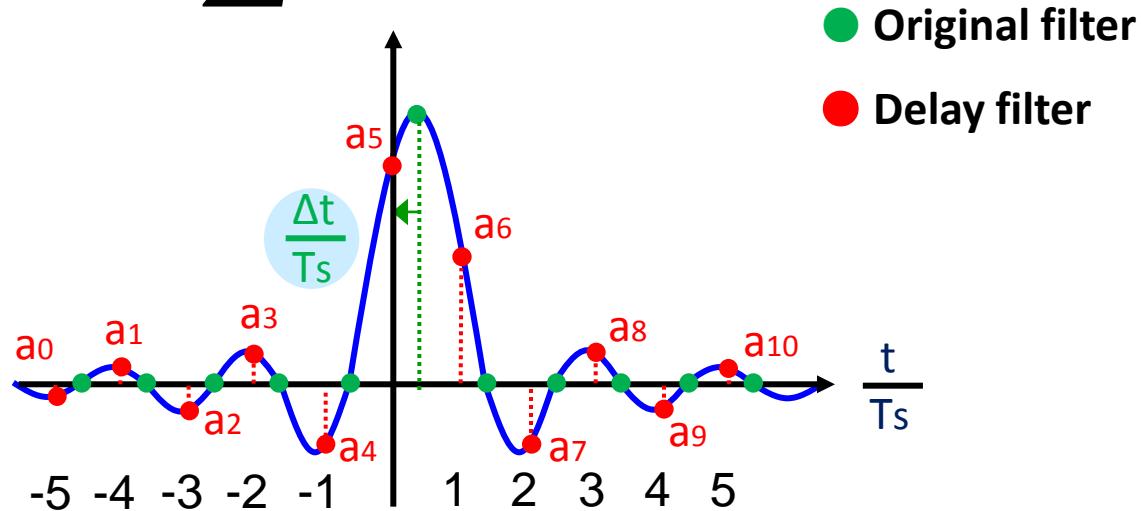
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Proposed Filter DC Gain Adjustment

Digital filter DC gain : $\sum a_n$

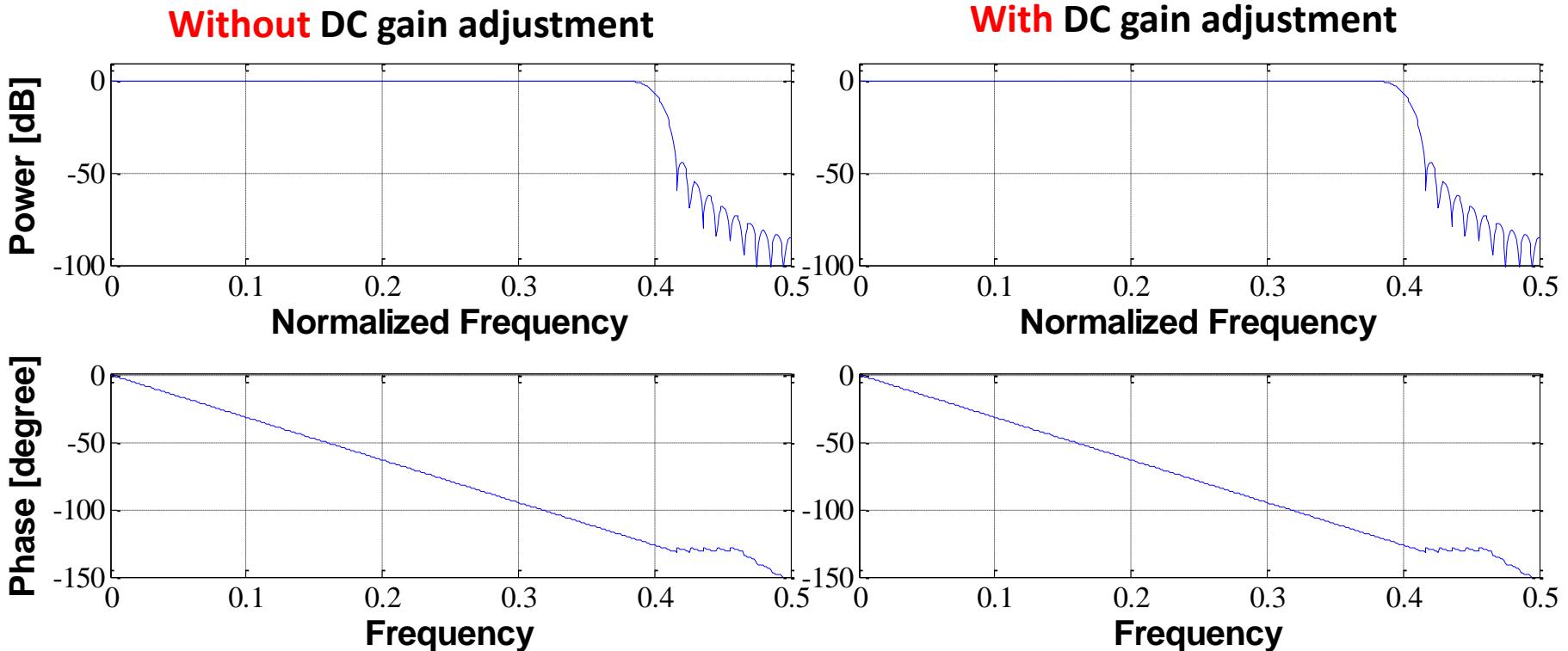


DC gain adjustment due to finite tap truncation
is required



$$\sum_{n=0}^N a'_n = \text{DC gain of original FIR filter}$$

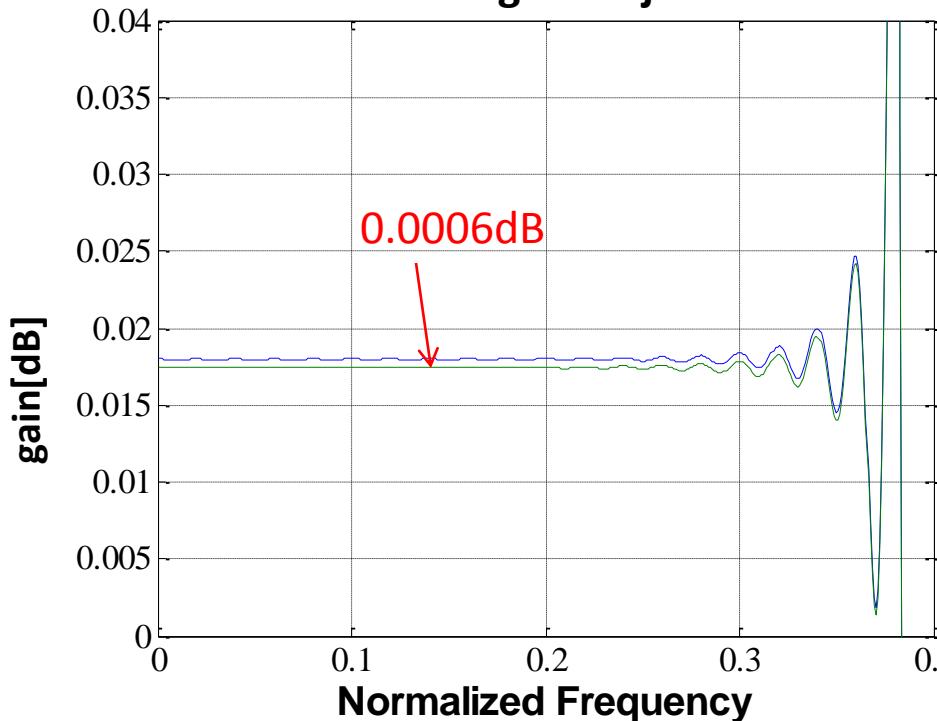
Frequency Characteristics of Proposed Delay Filter



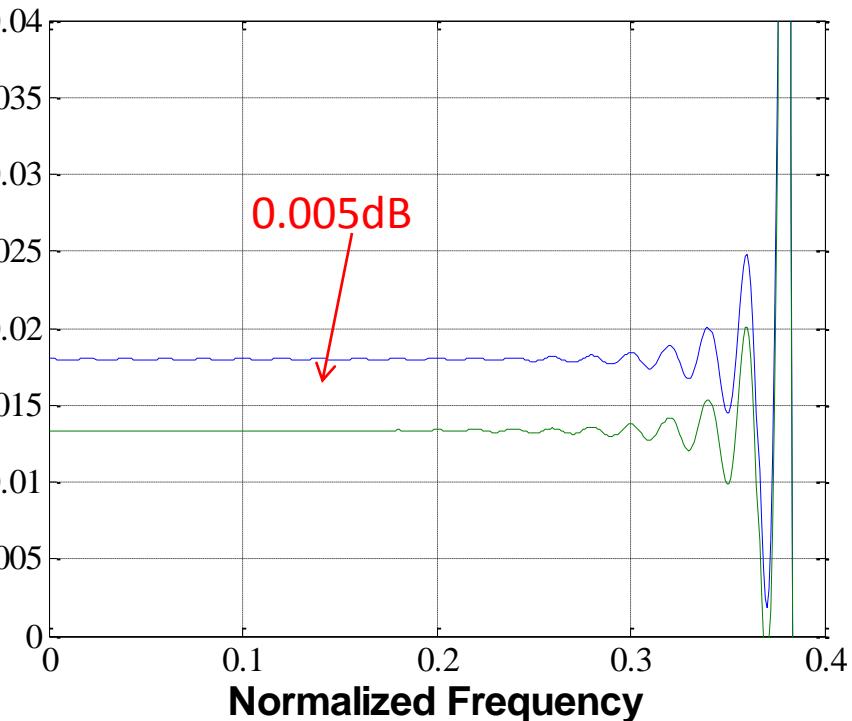
Delay	0.3 samples
Filter Tap	101 taps
Window	Han
Cut-off Freq.	$0.4 \cdot F_s$
FFTpoints	1024 points

Gain Characteristics of Proposed Delay Filter

— With DC gain adjustment
— Without DC gain adjustment



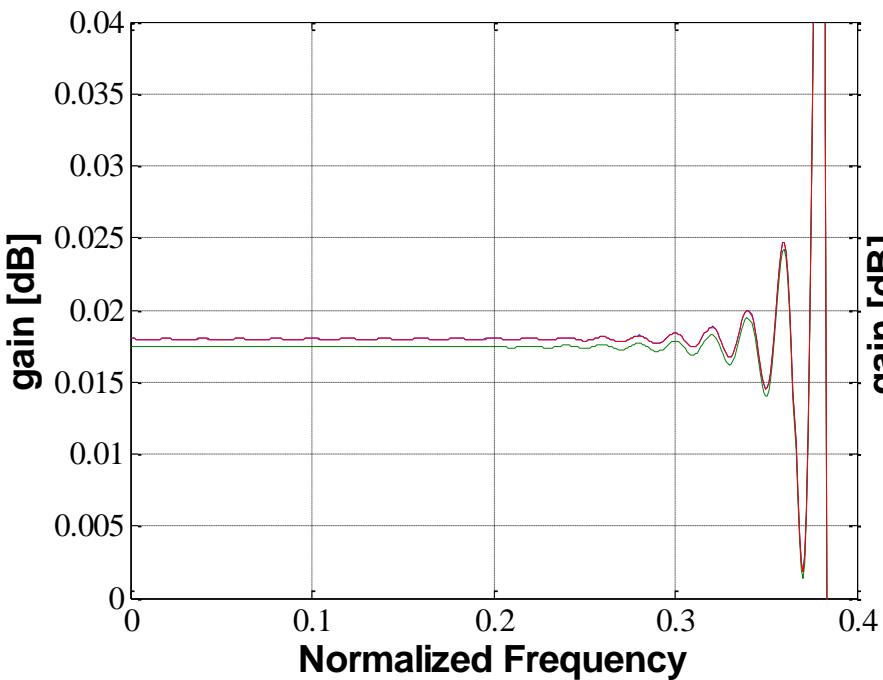
Delay	0.1 samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	$0.4 \cdot F_s$
FFTpoints	1024 points



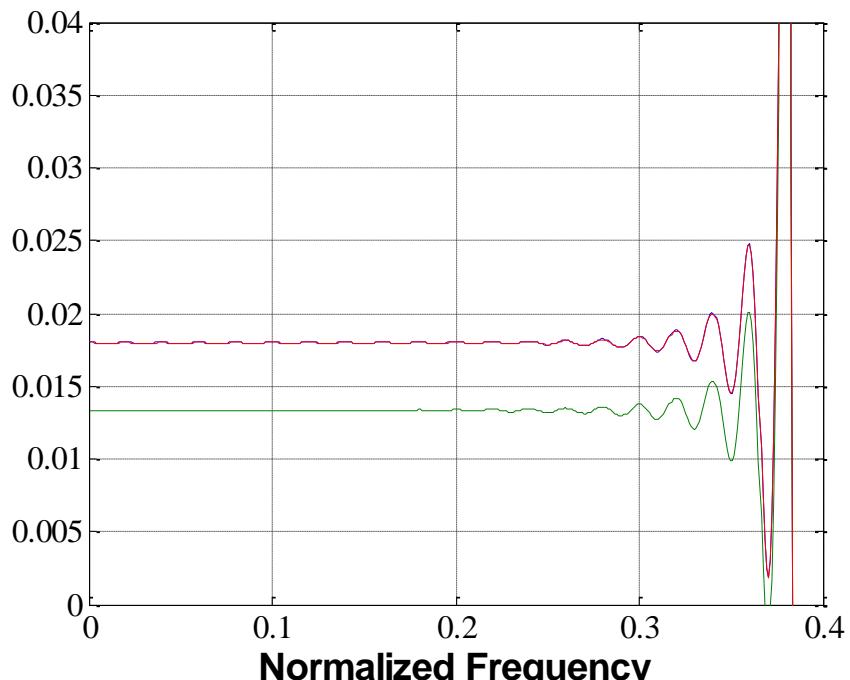
Delay	0.3 samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	$0.4 \cdot F_s$
FFT points	1024 points

Gain Characteristics of Proposed Delay Filter

- Original FIR filter
- With DC gain adjustment
- Without DC gain adjustment



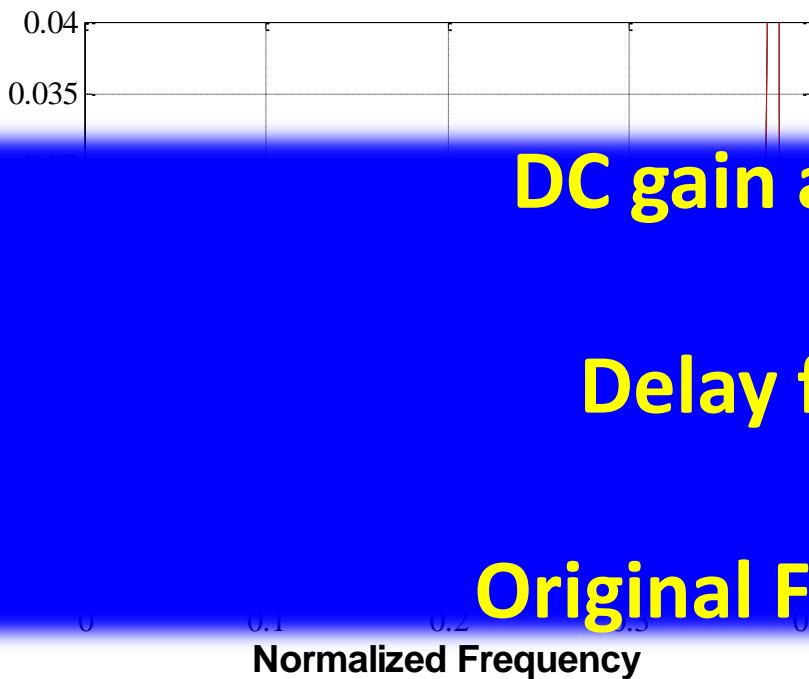
Delay	0.1 samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	$0.4 \cdot F_s$
FFTpoints	1024 points



Delay	0.3 samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	$0.4 \cdot F_s$
FFT points	1024 points

Gain Characteristics of Proposed Delay Filter

- Original FIR filter
- With DC gain adjustment
- Without DC gain adjustment



DC gain adjustment



Delay filter gain



Original FIR filter gain

Delay	0.1 samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	$0.4 \cdot F_s$
FFTpoints	1024 points

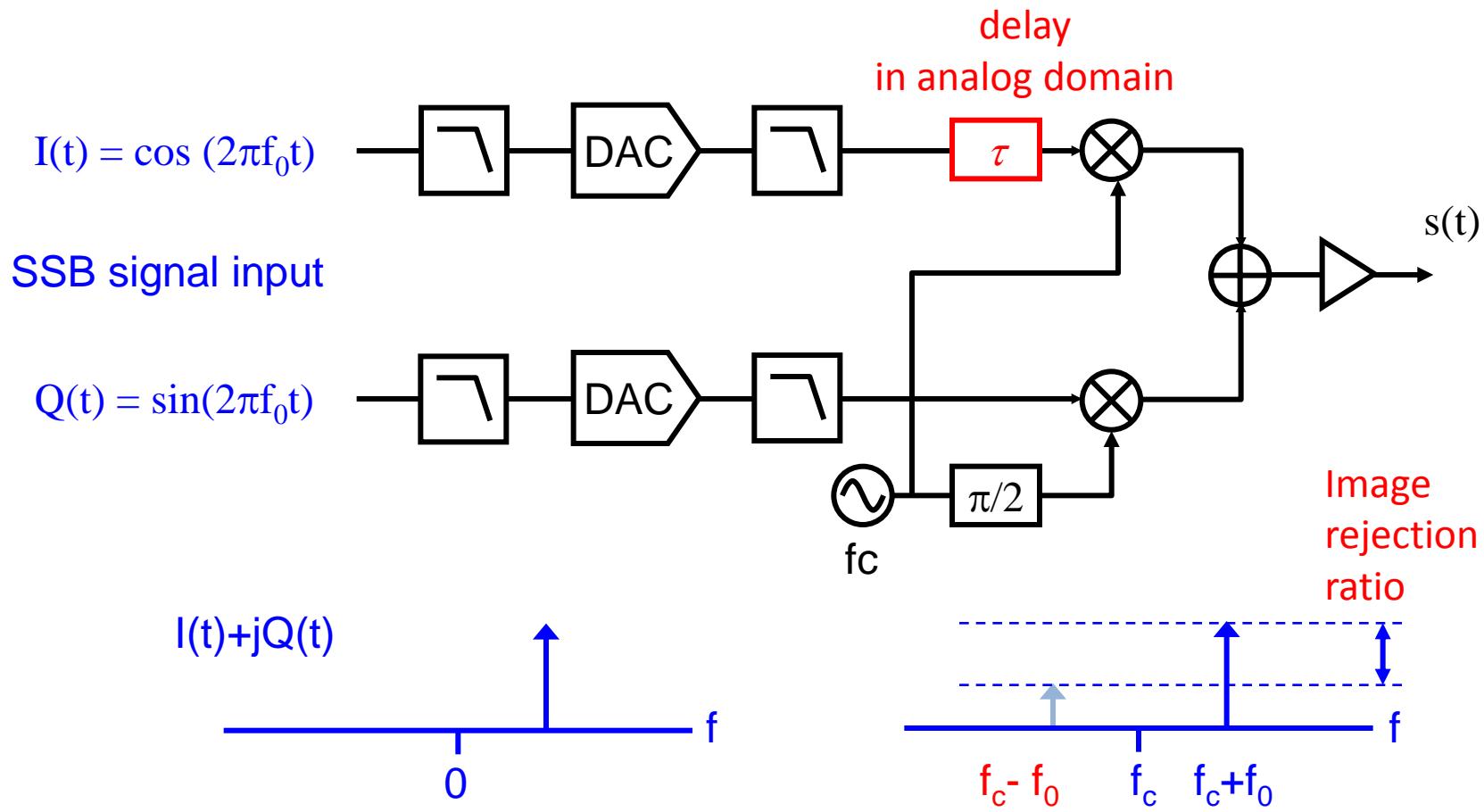
Delay	0.3 samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	$0.4 \cdot F_s$
FFT points	1024 points

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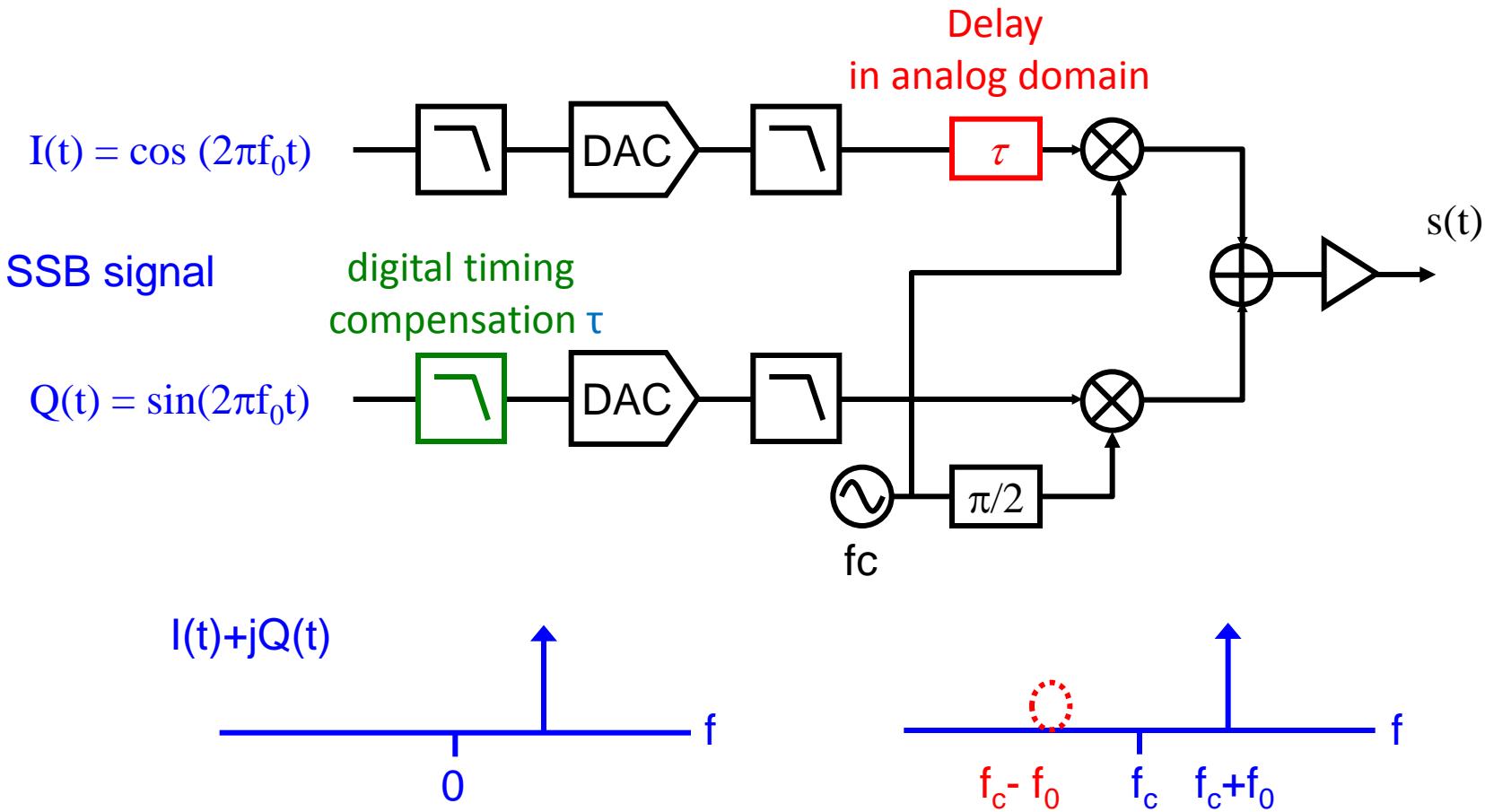
I/Q Delay Mismatch in Quadrature Modulator



SSB : single side band

DAC : digital-to-analog converter

I/Q Delay Mismatch Compensation in Quadrature Modulator

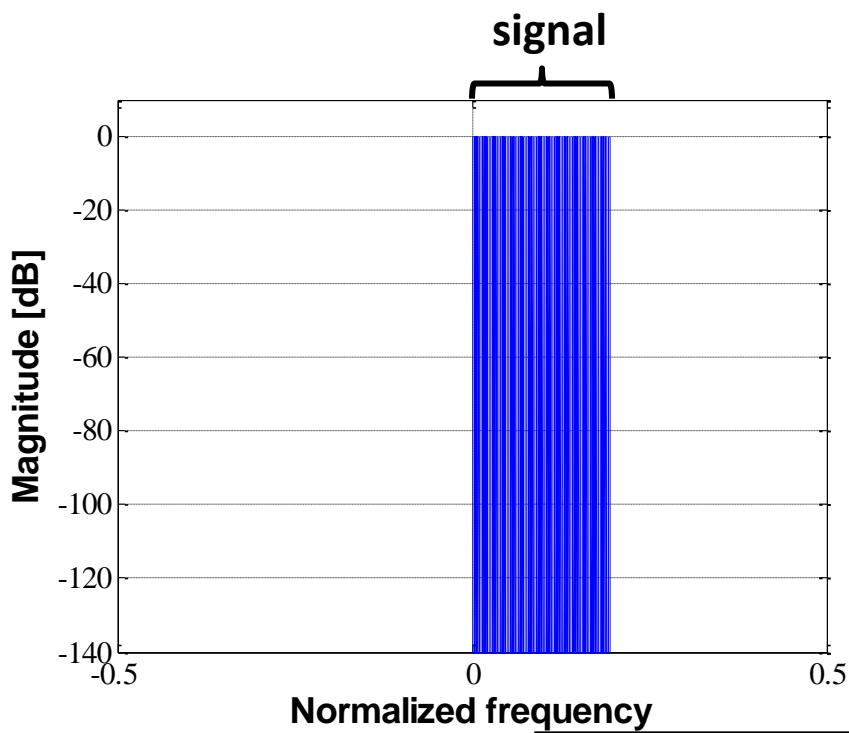


SSB : single side band

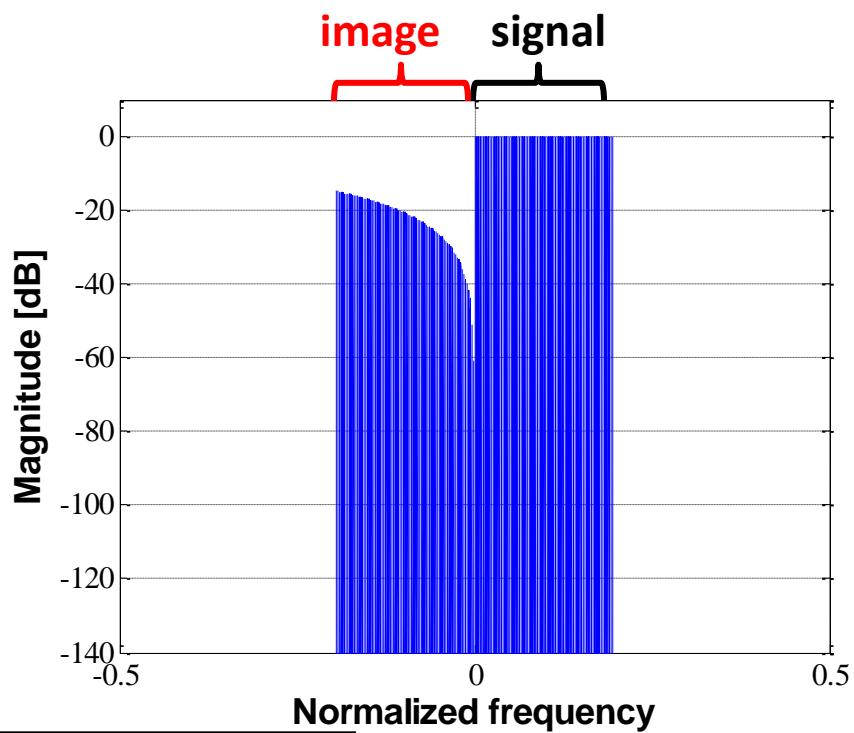
DAC : digital-to-analog converter

Matlab Simulation Results

(a) Ideal case



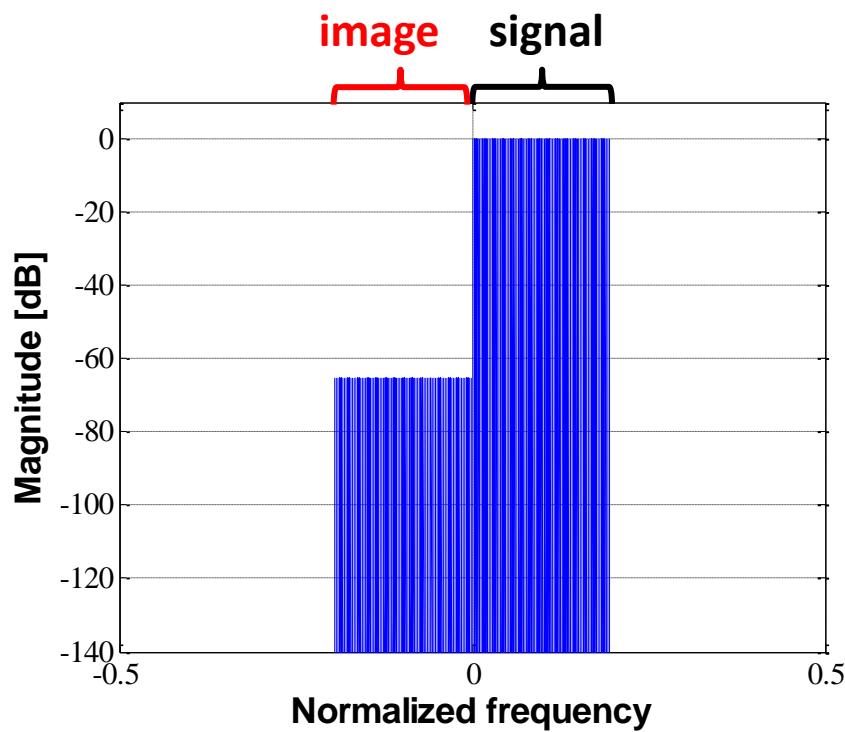
(b) Timing skew case



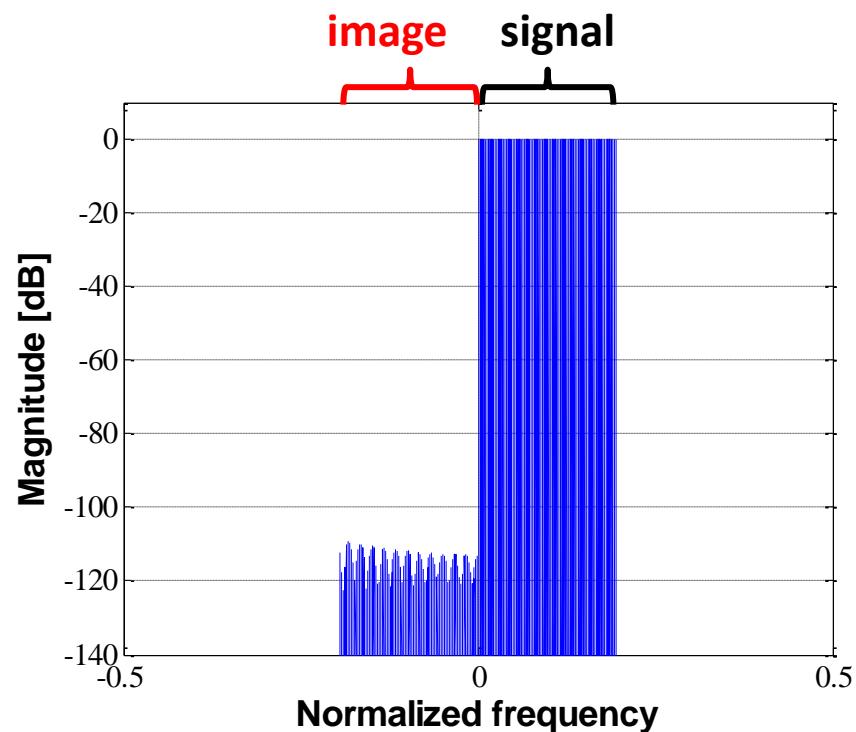
Delay	0.3 samples
Filter tap #	61 taps
Window	Han
FFT points	1024 points

Matlab Simulation Results

(c) Compensation using delay filter
Without adjustment of window, gain



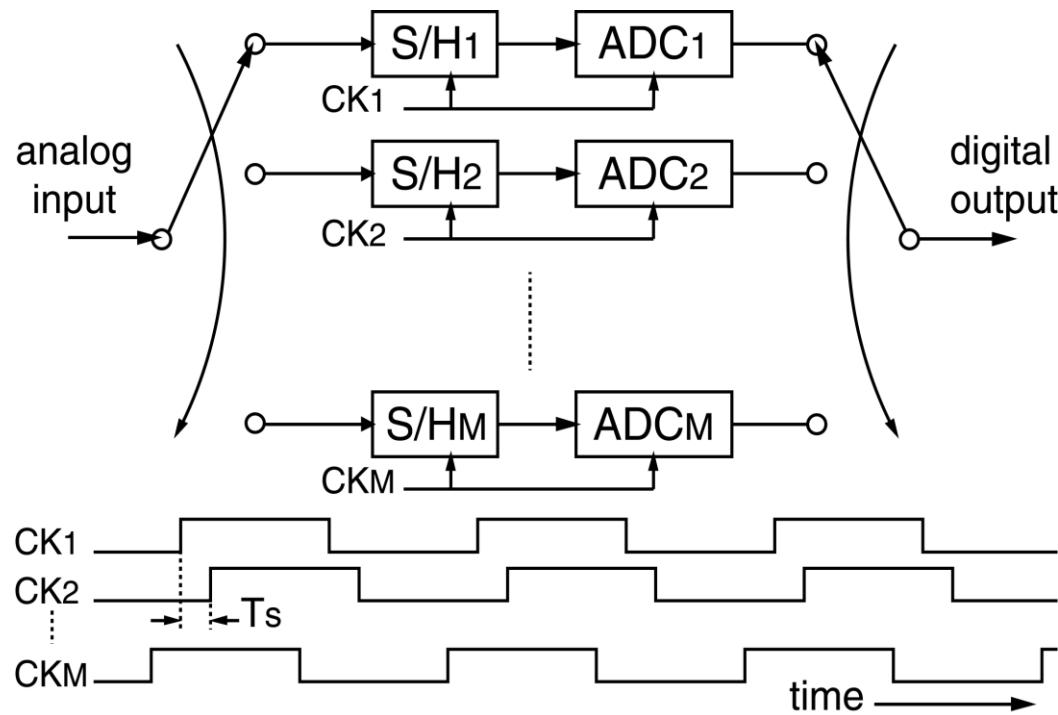
(d) Compensation using delay filter
With adjustment of window, gain



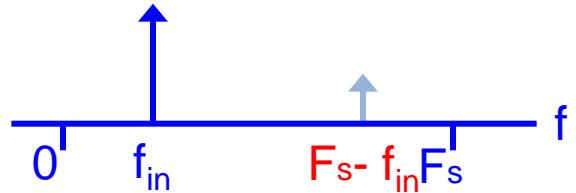
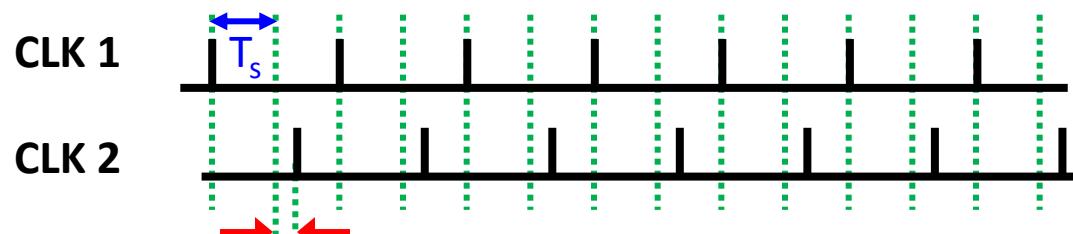
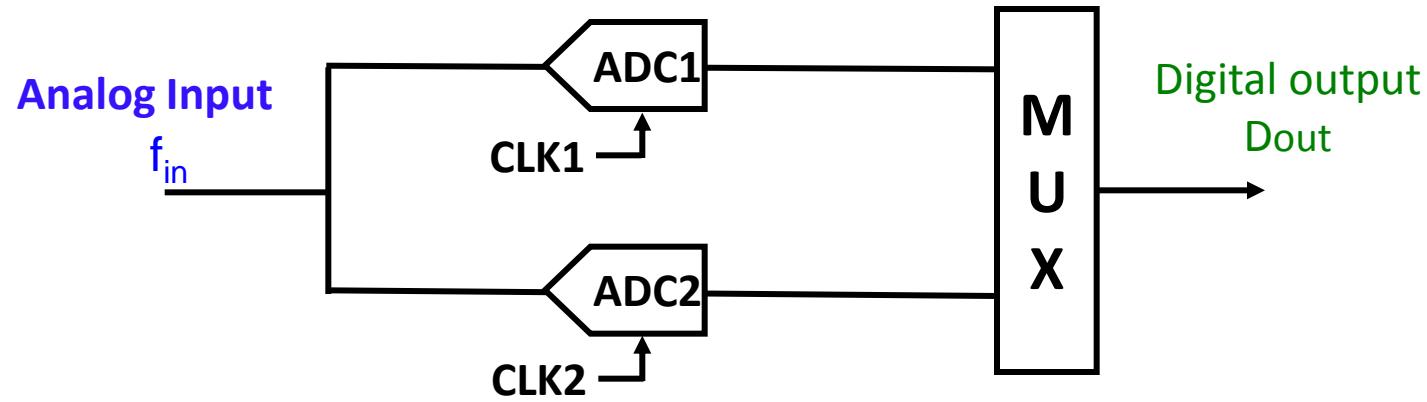
Delay	0.3 samples
Filter tap	61 taps
Window	Han
FFT points	1024 points

Interleaved ADC System

■ M channel ADCs → M-times sampling rate



Timing Skew in Interleaved ADC System

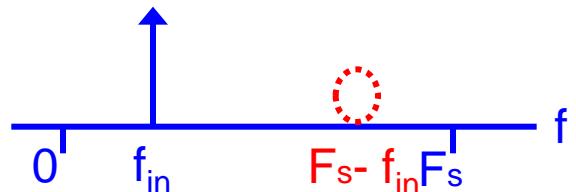
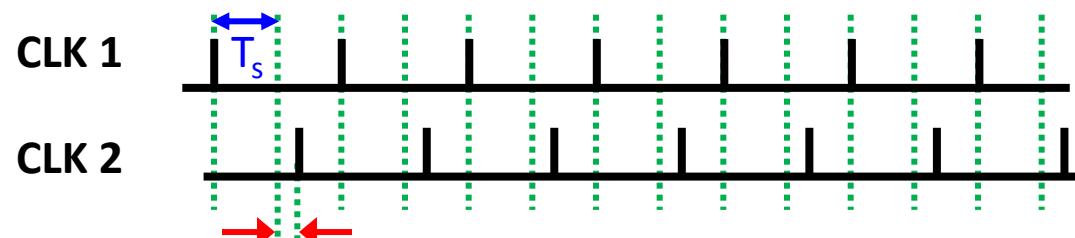
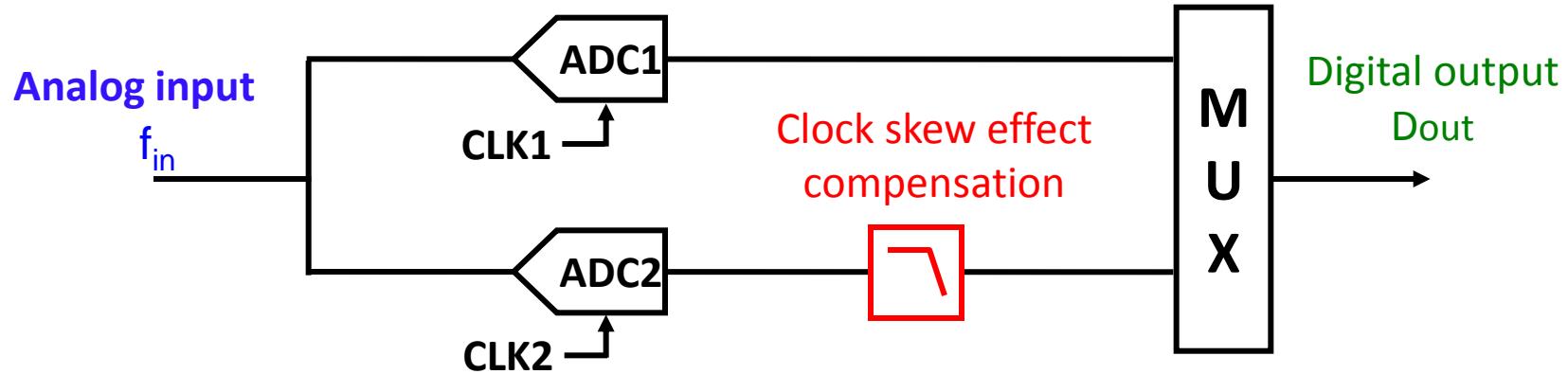


ADC : analog-to-digital converter

$$F_s = 1/T_s$$



Timing Skew Compensation in Interleaved ADC System

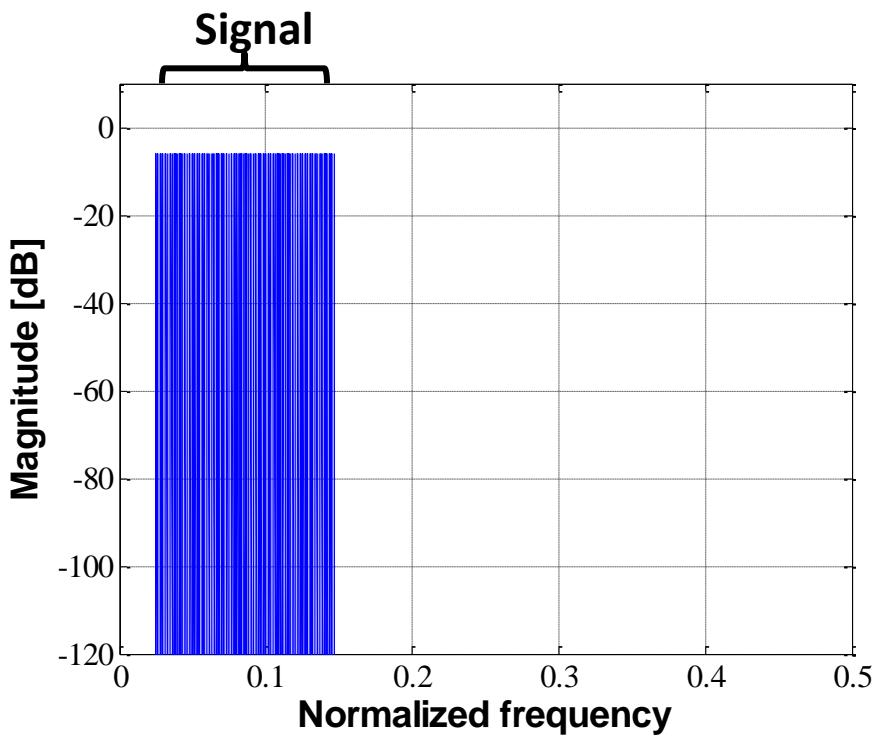


ADC : analog-to-digital converter

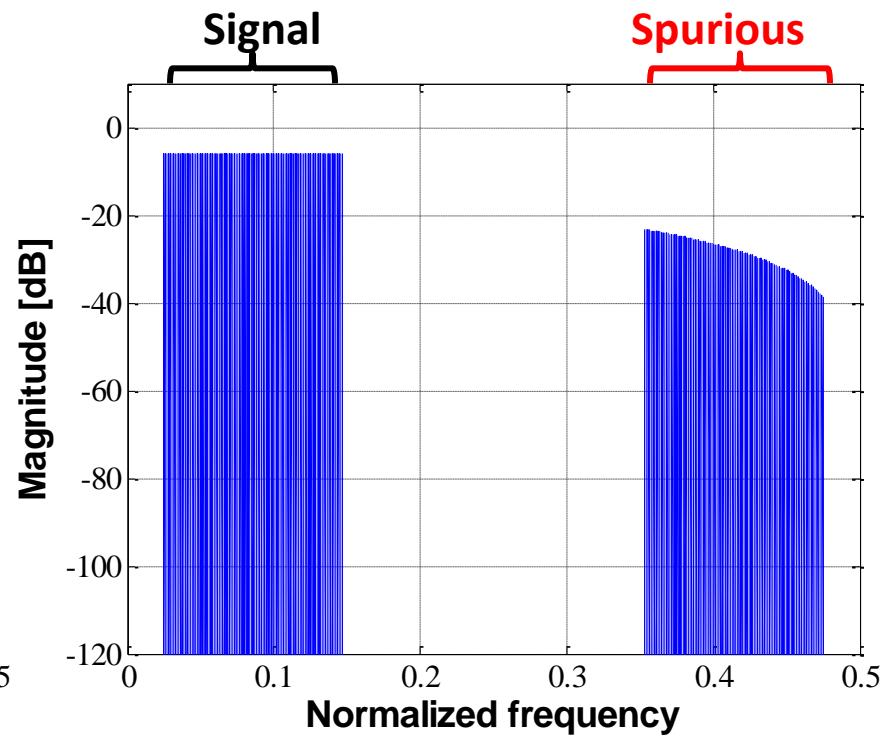


Matlab Simulation Results

(a) Ideal case



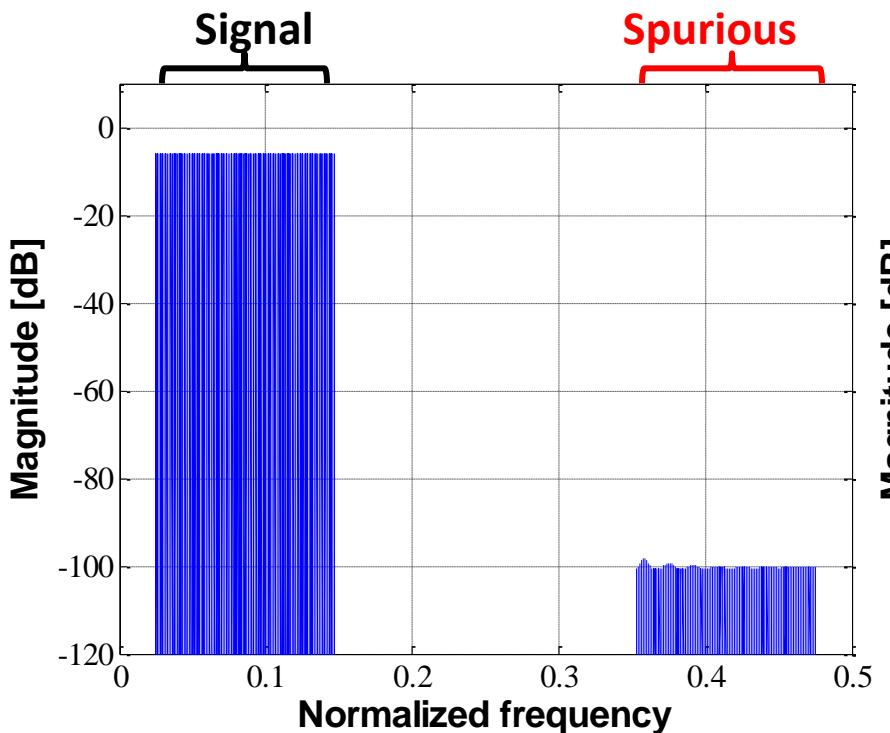
(b) Timing skew case



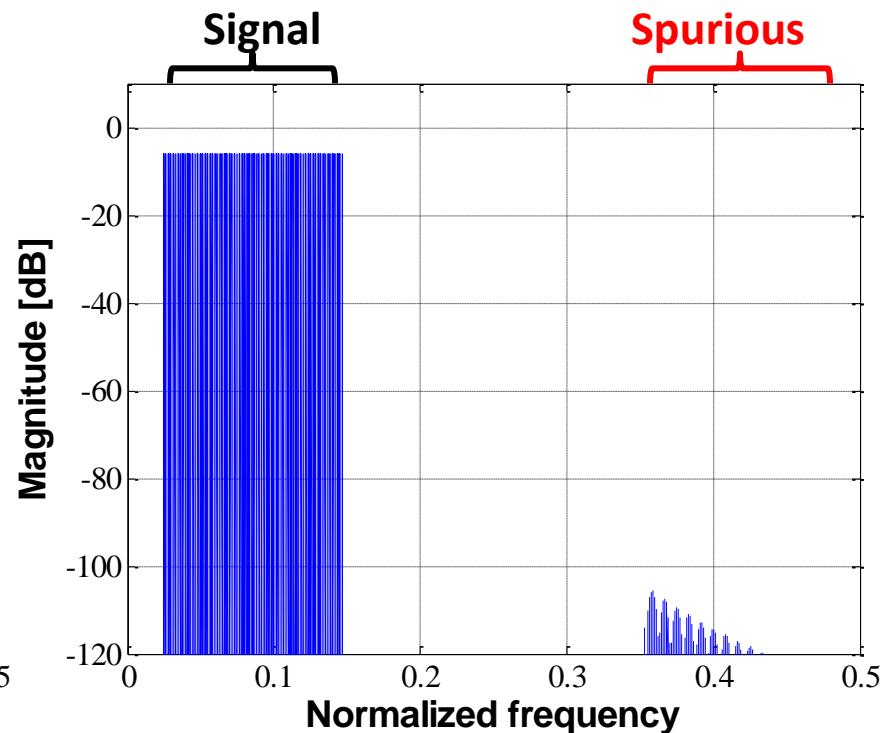
Delay	0.3 samples
Filter tap	61 taps
Window	Han
FFTpoints	1024 points

Matlab Simulation Results

(c) Compensation using delay filter
Without adjustment of window, gain



(d) Compensation using delay filter
With adjustment of window, gain



Delay	0.3 samples
Filter tap	61 taps
Window	Han
FFTpoints	1024 points

Conclusion

- Linear phase digital filter
with fine time resolution of group delay
- Design consideration
 - How to apply window
 - DC gain adjustment
- Application Examples
 - I/Q delay mismatch compensation
in quadrature modulator
 - Timing skew compensation in interleaved ADC system

Future work

- Implementation issues
 - Finite word length, finite tap effects
 - LSI implementation

