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Digitally-Assisted Compensation for Timing Skew in ATE Systems

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Advantest Corporation

Gunma University

Contents

- Research Goal
- Conventional Linear Phase Digital Filter Condition
- New Linear Phase Digital Filter Condition
  - Time-Shift, Impulse Response of Ideal Filter
  - New Linear Phase Digital Filter
- MATLAB Simulation
- Design Considerations
  - Window
  - Gain Adjustment
- Application
- Conclusion

Research Goal

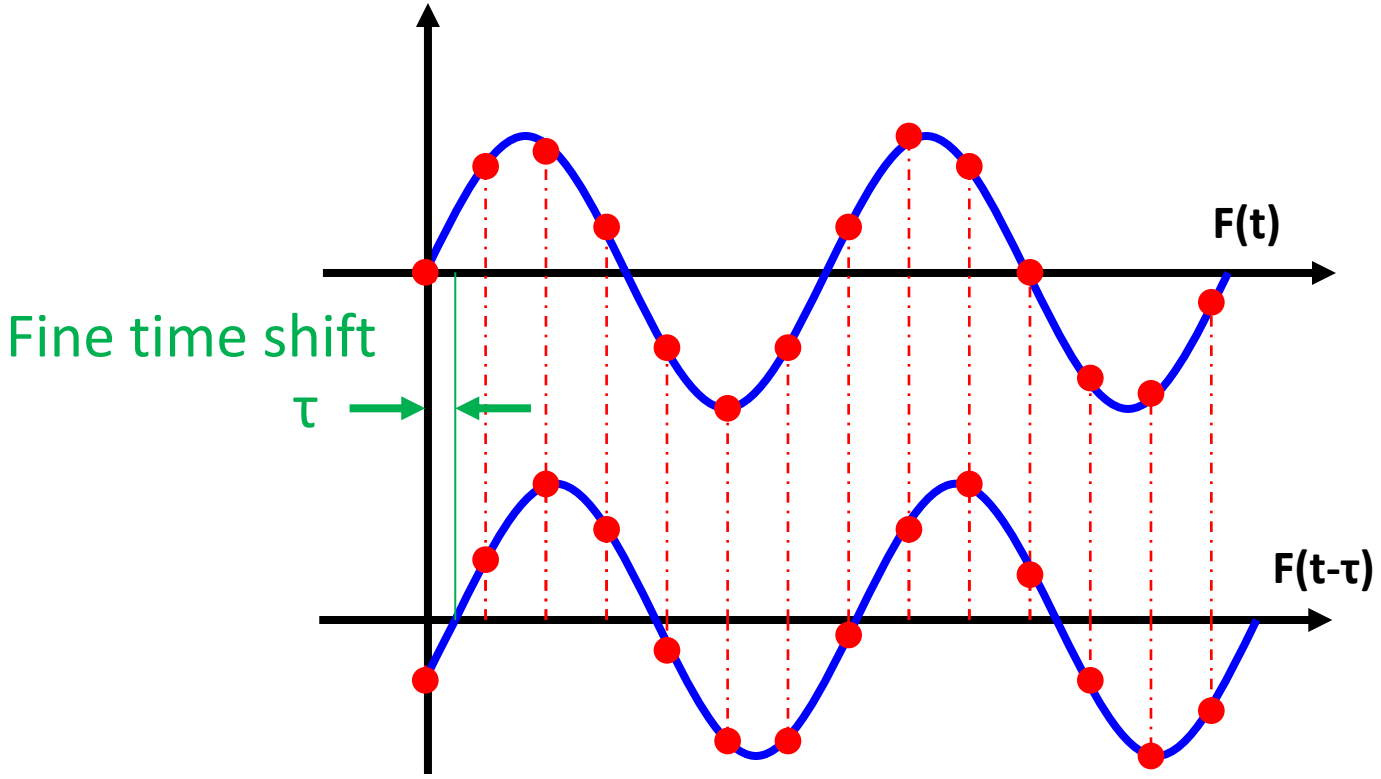
Timing skew is a major problem in ATE systems

Digital compensation for timing skew
⇒ Linear phase is important

Conventional linear-phase digital filter ⇒ coarse timing adjustment

Proposed linear-phase digital filter ⇒ fine timing adjustment

Features of Proposed Digital Filter



- Fine time resolution
- Linear phase (to preserve waveform in time domain)
- Applicable to bandpass signal (as well as lowpass signal)

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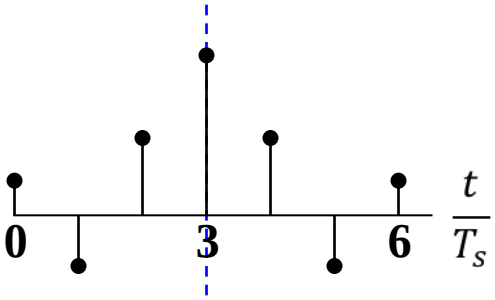
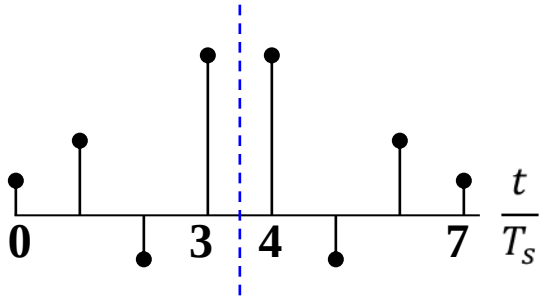
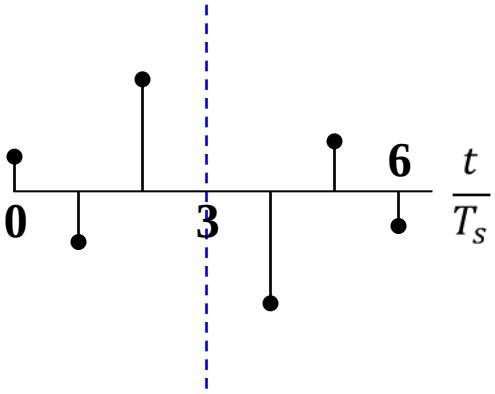
Linear Phase FIR Filter Impulse Response



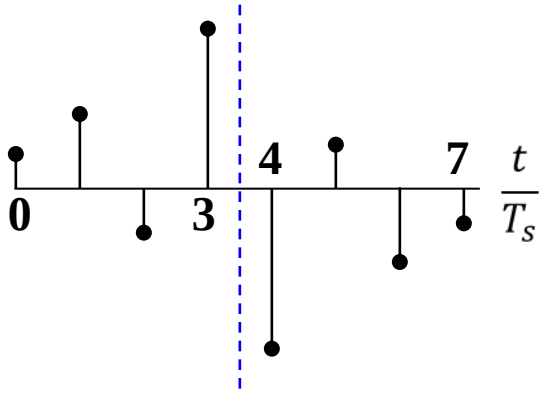
(1)Case 1
odd # of taps ・even symmetry



(2) Case 2
even # of taps・even symmetry



(3) Case 3
odd # of taps・odd symmetry



(4) Case 4
even # of taps・odd symmetry

Frequency Characteristics

$h(nT)$	$H(e^{j\omega T})$
Case 1	$e^{-j\omega(N-1)T_s/2} \sum_{k=0}^{(N-1)/2} a_k \cos[\omega k T_s]$
Case 2	$e^{-j\omega(N-1)T_s/2} \sum_{k=1}^{N/2} b_k \cos[\omega(k-1/2)T_s]$
Case 3	$e^{-j(\omega(N-1)T_s/2-\pi/2)} \sum_{k=0}^{(N-1)/2} a_k \sin[\omega k T_s]$
Case 4	$e^{-j(\omega(N-1)T_s/2-\pi/2)} \sum_{k=1}^{N/2} b_k \sin[\omega(k-1/2)T_s]$

Phase : proportional to ω (linear phase)

Time resolution of group delay Ts/2

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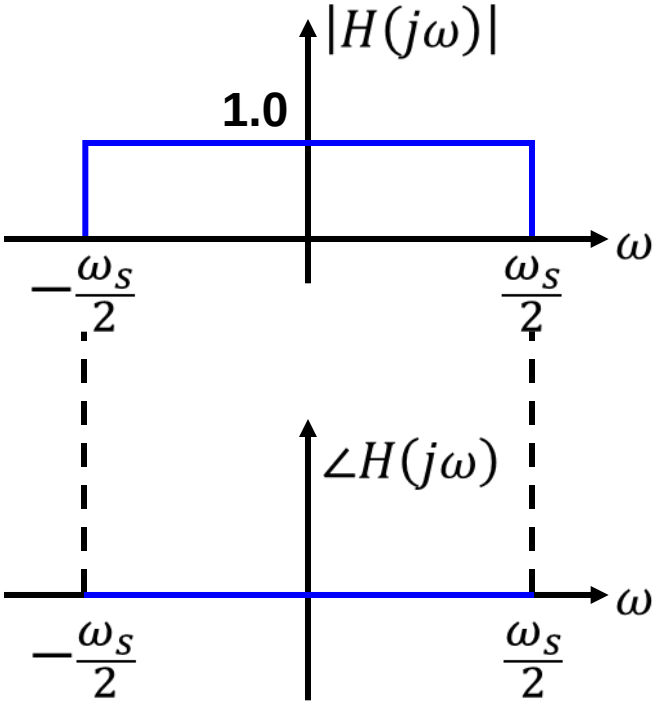
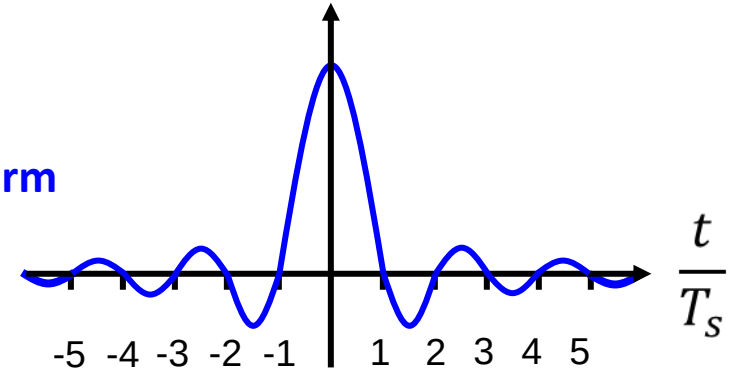
Ideal LPF

Frequency Characteristics



Fourier Transform

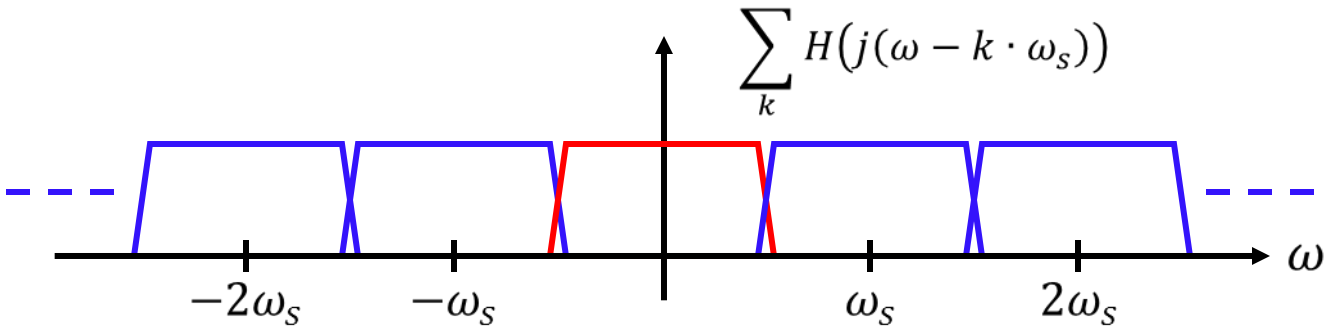
Impulse Response



$$h(t) = \frac{1}{T_s}\mathrm{sinc}\left(\pi\frac{t}{T_s}\right)$$

$\omega_s = \frac{2\pi}{T_s}$: **Sampling Frequency**

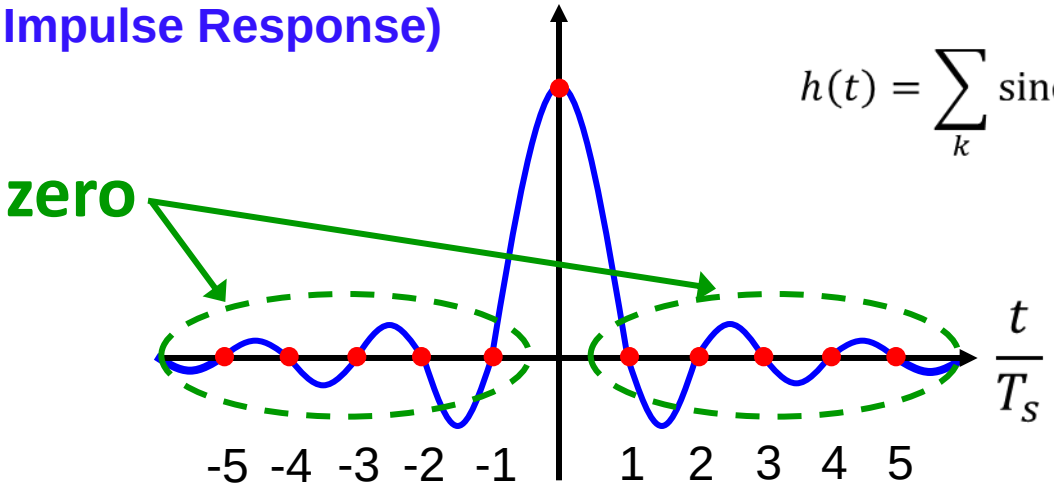
Discrete-Time Representation of Ideal LPF



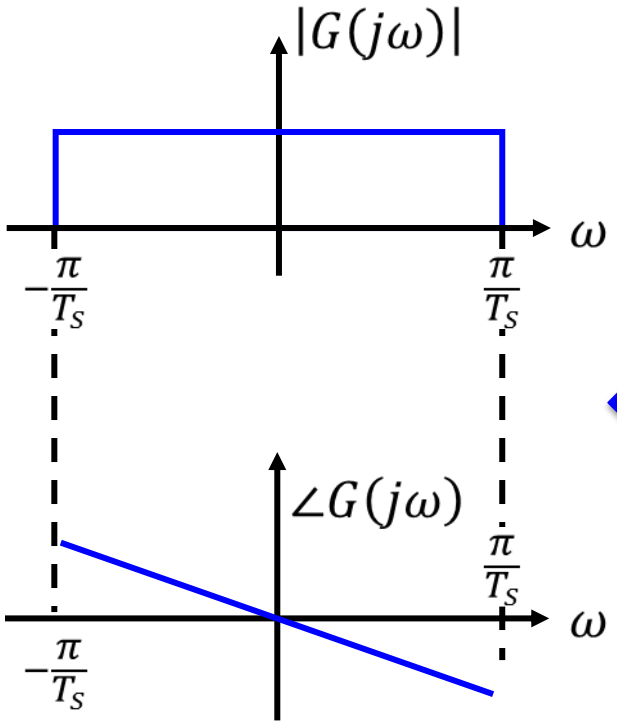
Fourier Transform

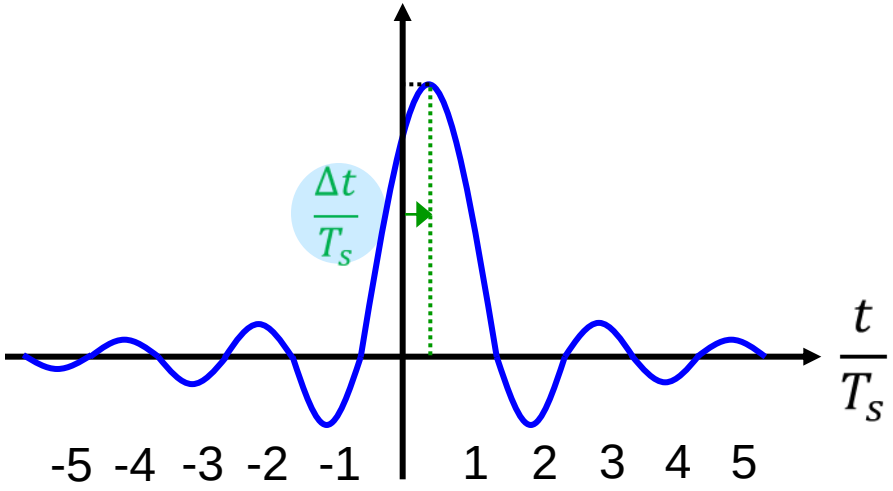
FIR (Finite Impulse Response)

$$h(t) = \sum_k \text{sinc}\left(\pi \frac{k \cdot T_s}{T_s}\right) \delta(t - k \cdot T_s)$$



Impulse Response Time-Shift





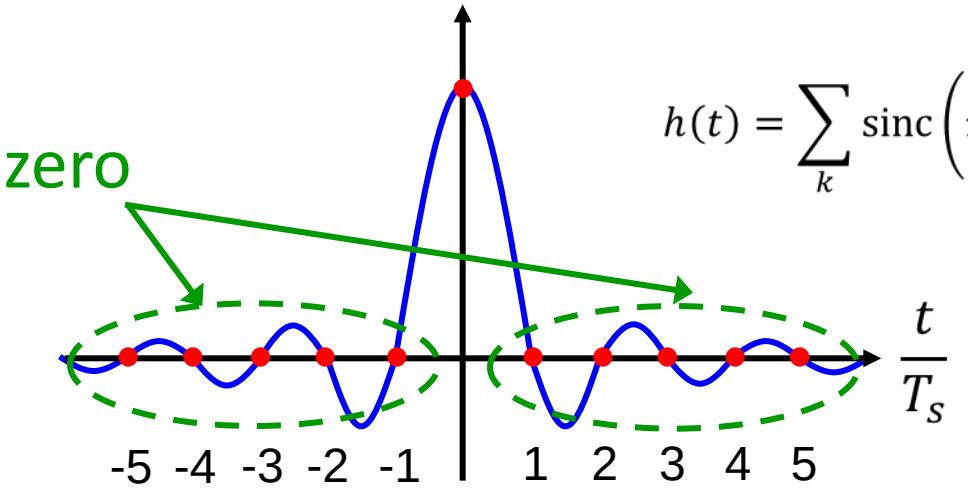
$$\angle G(j\omega) = -\omega\Delta t$$

Δt time-shift of impulse response

No change of Gain

Time-Shift and Filter Coefficients

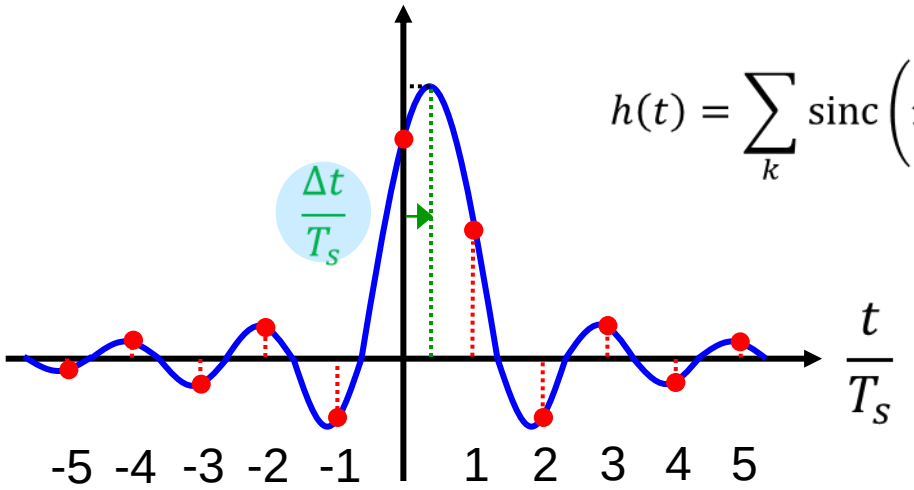
FIR filter



$$h(t) = \sum_k \mathrm{sinc}\left(\pi \frac{k \cdot T_s}{T_s}\right) \delta(t - k \cdot T_s)$$

Time Shift

IIR Filter



$$h(t) = \sum_k \mathrm{sinc}\left(\pi \frac{k \cdot T_s - \Delta t}{T_s}\right) \delta(t - k \cdot T_s)$$

Ideal Delay-Filter

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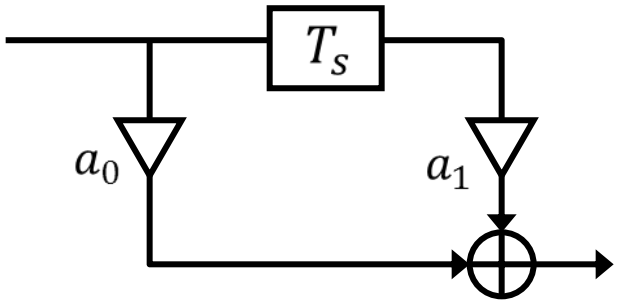
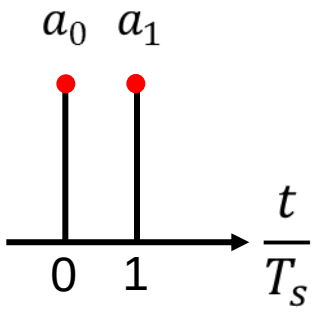
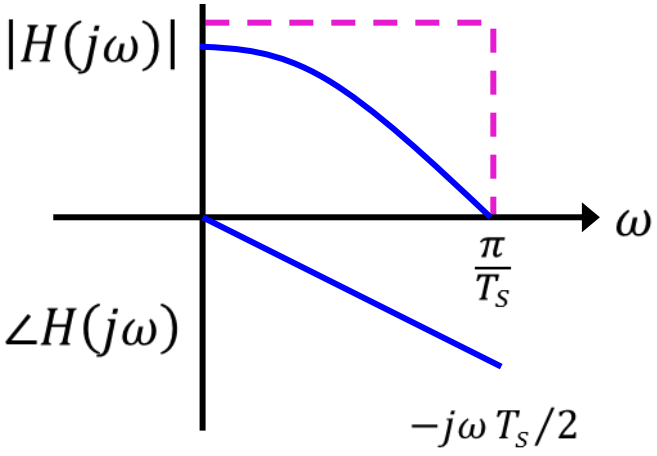
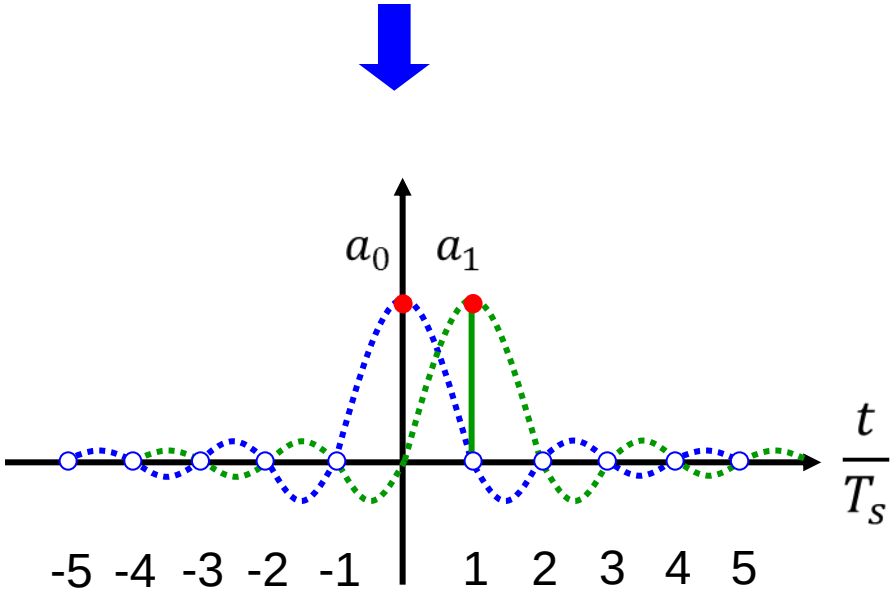
2-Tap Filter: Model

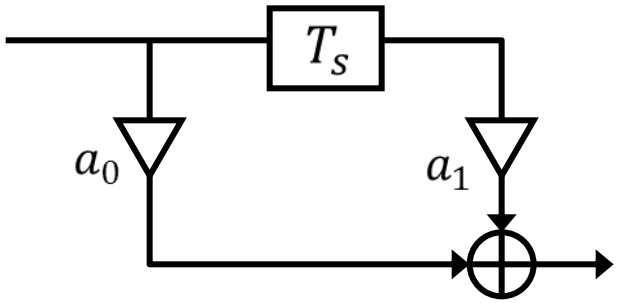


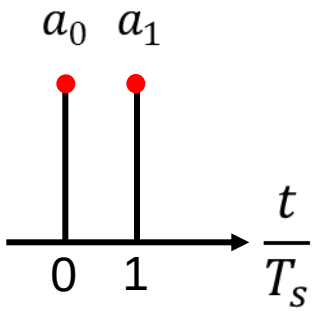




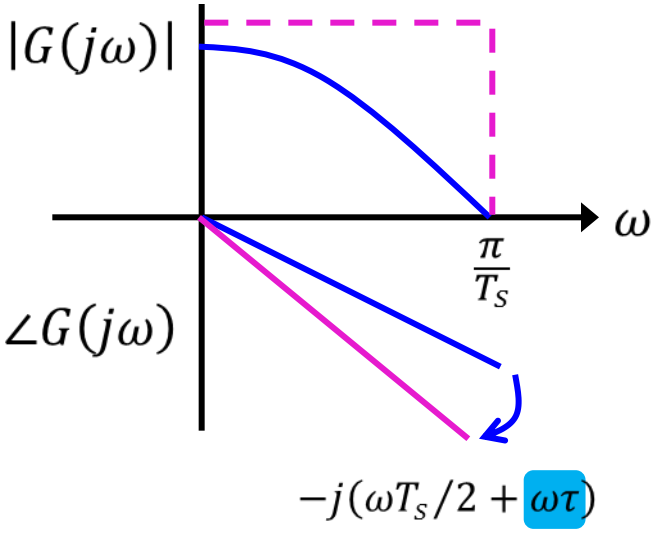


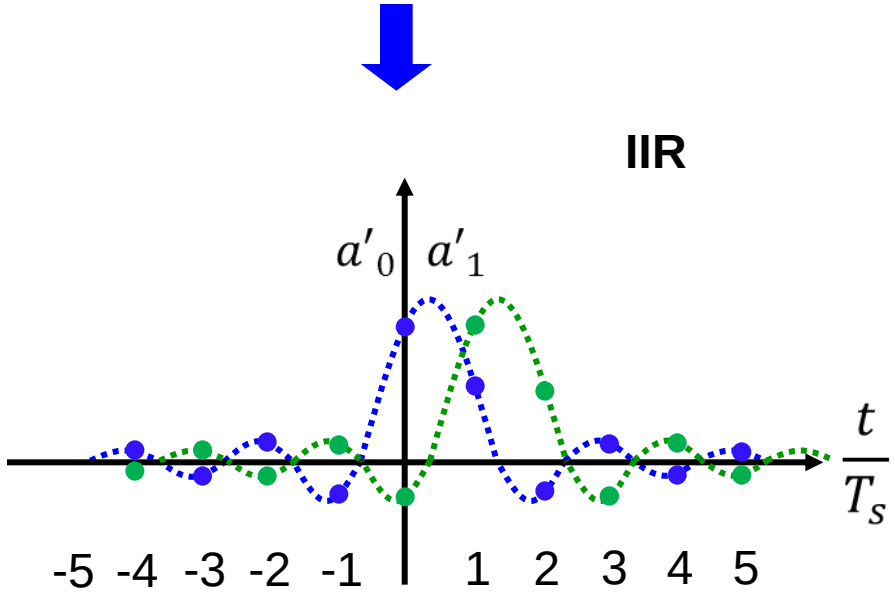
2-Tap Filter: Delay Model





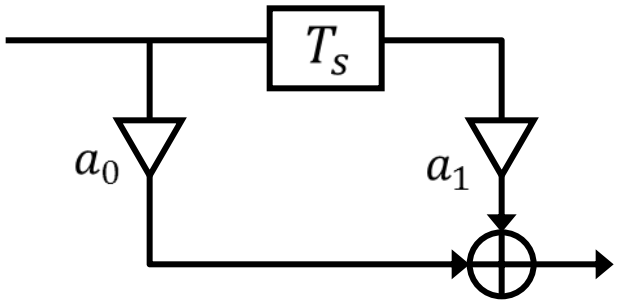
FIR

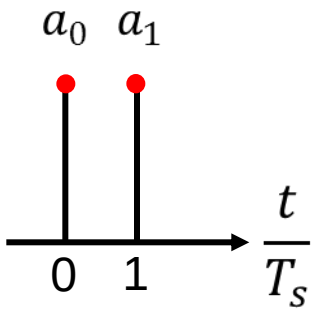




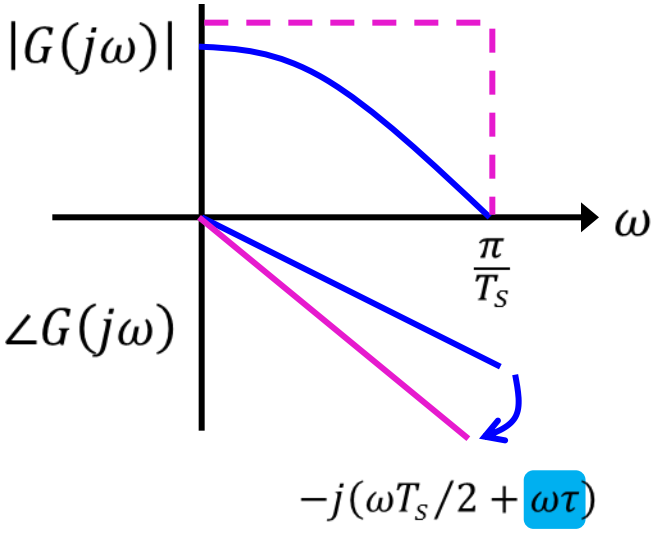
IIR

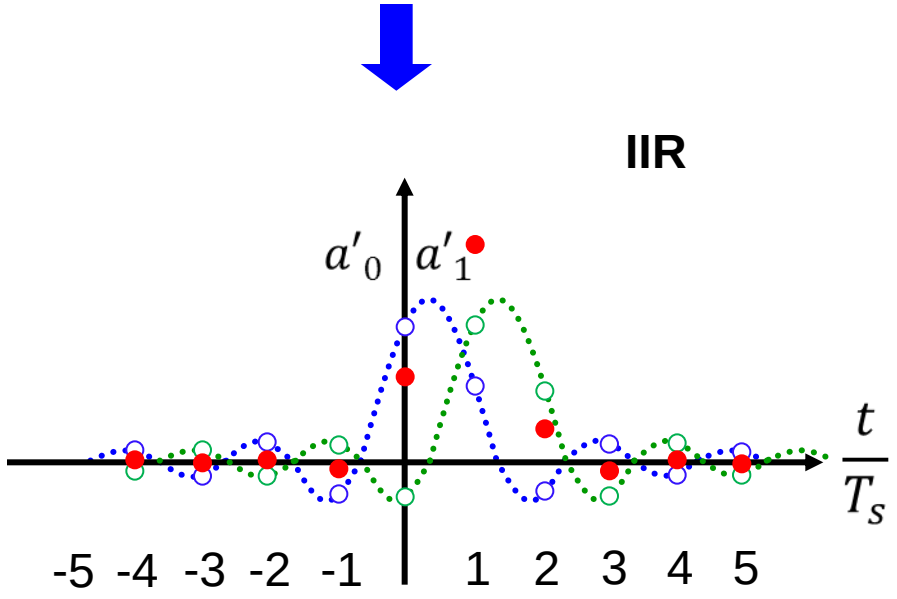
2-Tap Filter: Delay Model





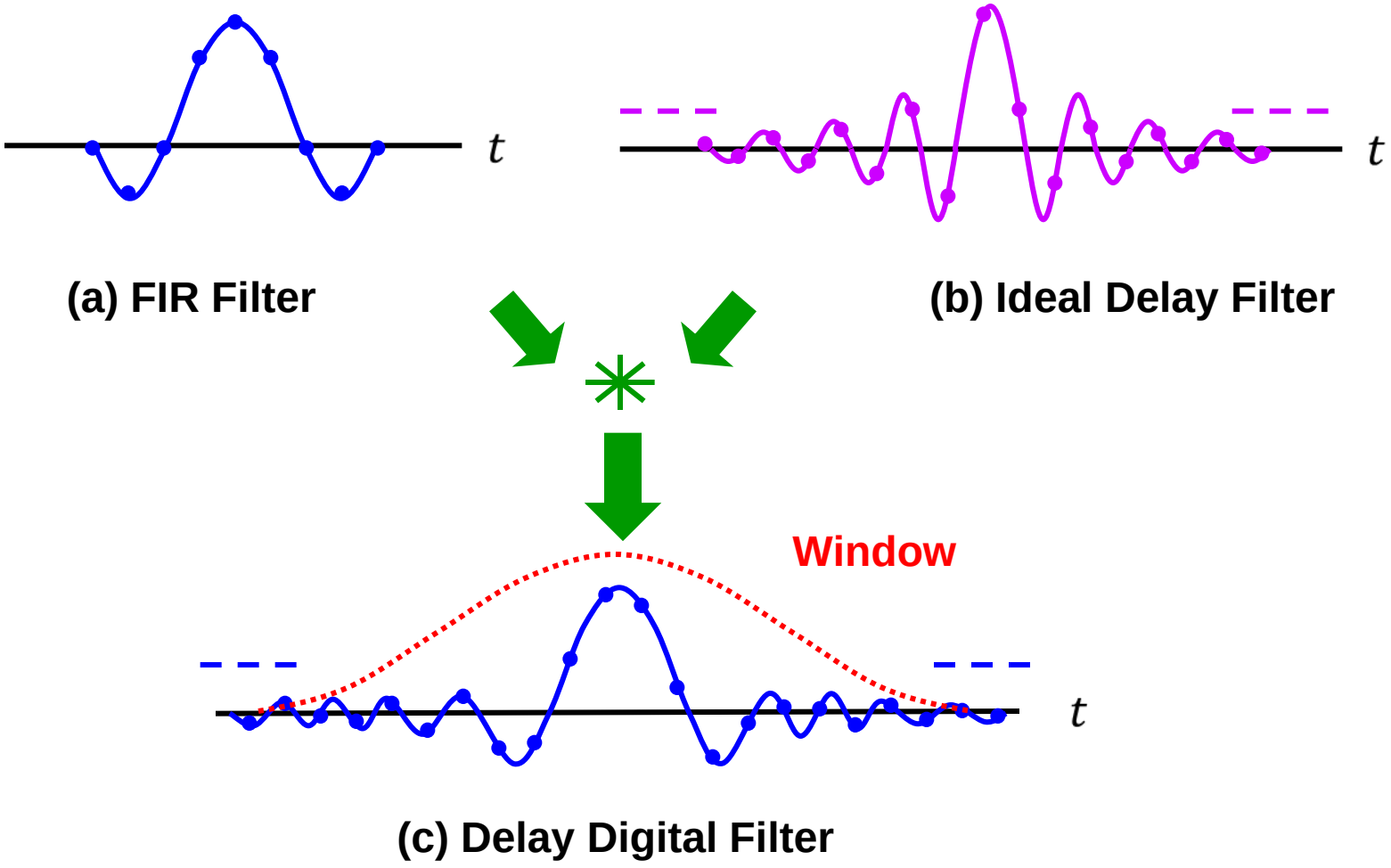
FIR





IIR

Proposed Delay Digital Filter



(a) FIR Filter

(b) Ideal Delay Filter

(c) Delay Digital Filter

Frequency Characteristics of Proposed Delay Digital Filter

$g(nT)$	$G(e^{j\omega T})$
Case 1	$e^{-j\omega(N-1)T_s/2+\omega\tau} \sum_{k=0}^{(N-1)/2} a_k \cos[\omega k T_s]$
Case 2	$e^{-j\omega(N-1)T_s/2+\omega\tau} \sum_{k=1}^{N/2} b_k \cos[\omega(k-1/2)T_s]$
Case 3	$e^{-j(\omega(N-1)T_s/2-\pi/2+\omega\tau)} \sum_{k=0}^{(N-1)/2} a_k \sin[\omega k T_s]$
Case 4	$e^{-j(\omega(N-1)T_s/2-\pi/2+\omega\tau)} \sum_{k=1}^{N/2} b_k \sin[\omega(k-1/2)T_s]$

Phase : proportional to ω（linear phase）

Group delay time resolution τ : Arbitrary small

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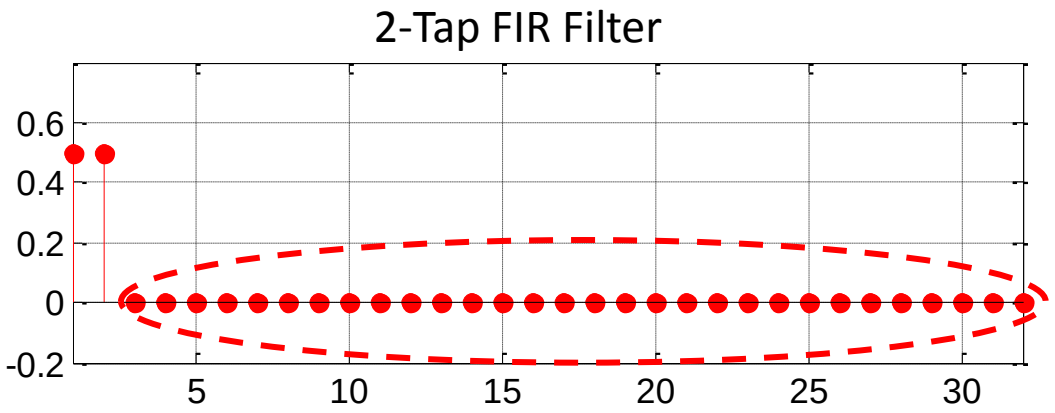
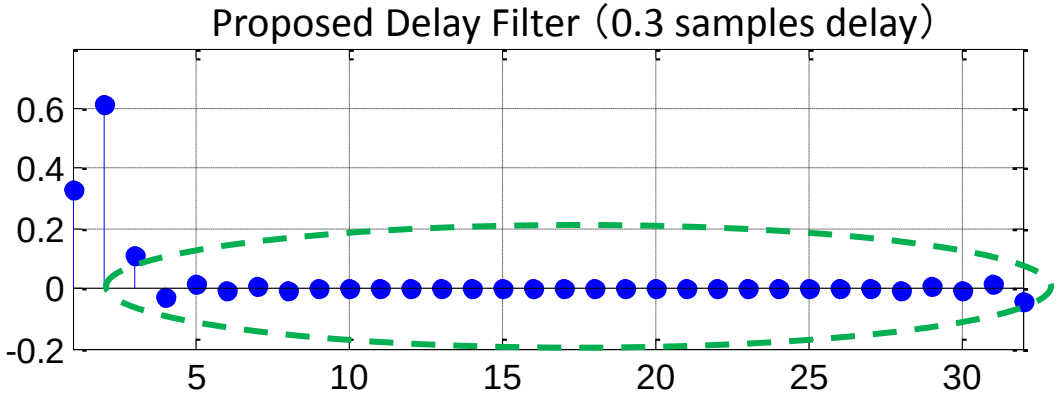
Comparison of 2-Tap Filter Impulse Responses



2-tap FIR coefficients

zero

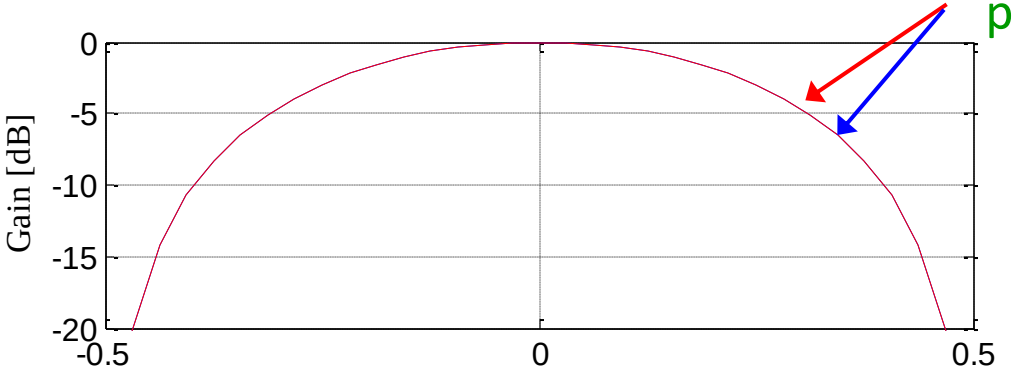


Impulse response changes.

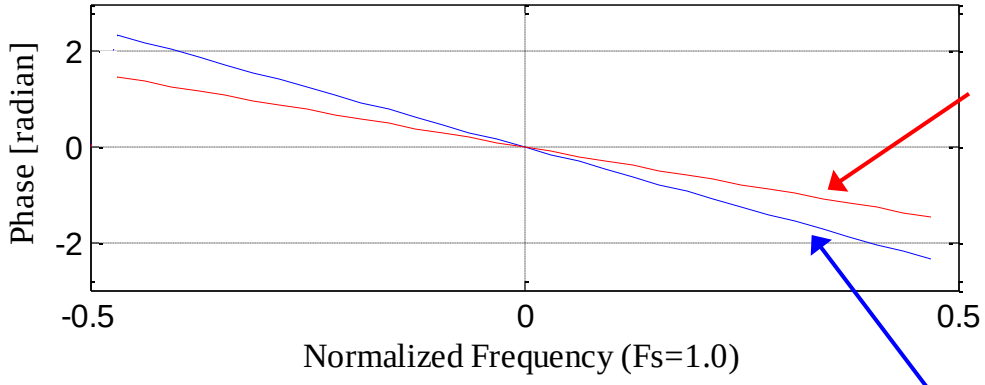
Non-zero

Comparison of 2-Tap Filter Frequency Characteristics

Original filter and proposed delay filter



No change of gain

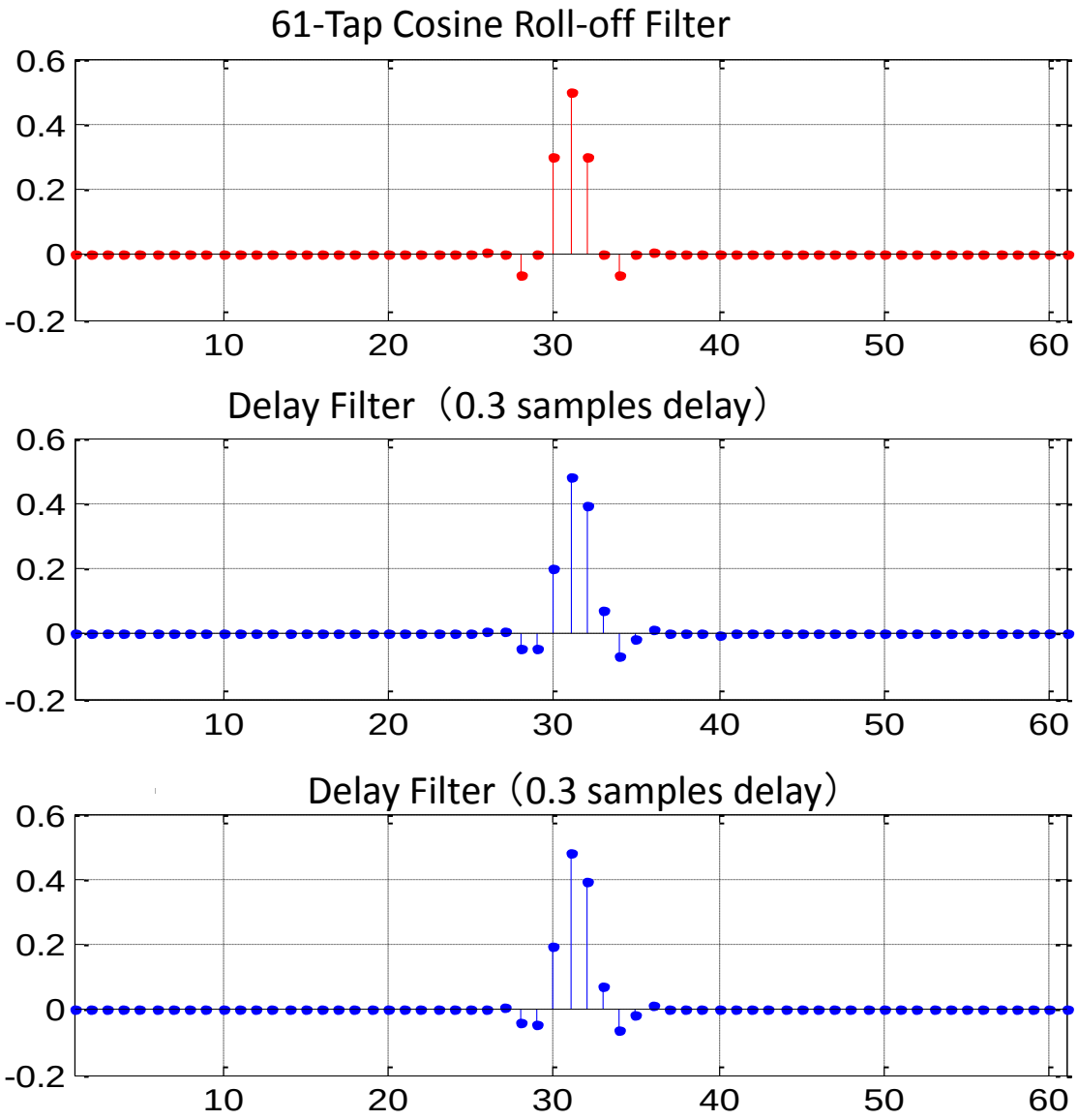


Original filter

Phase slope changes

Proposed delay filter

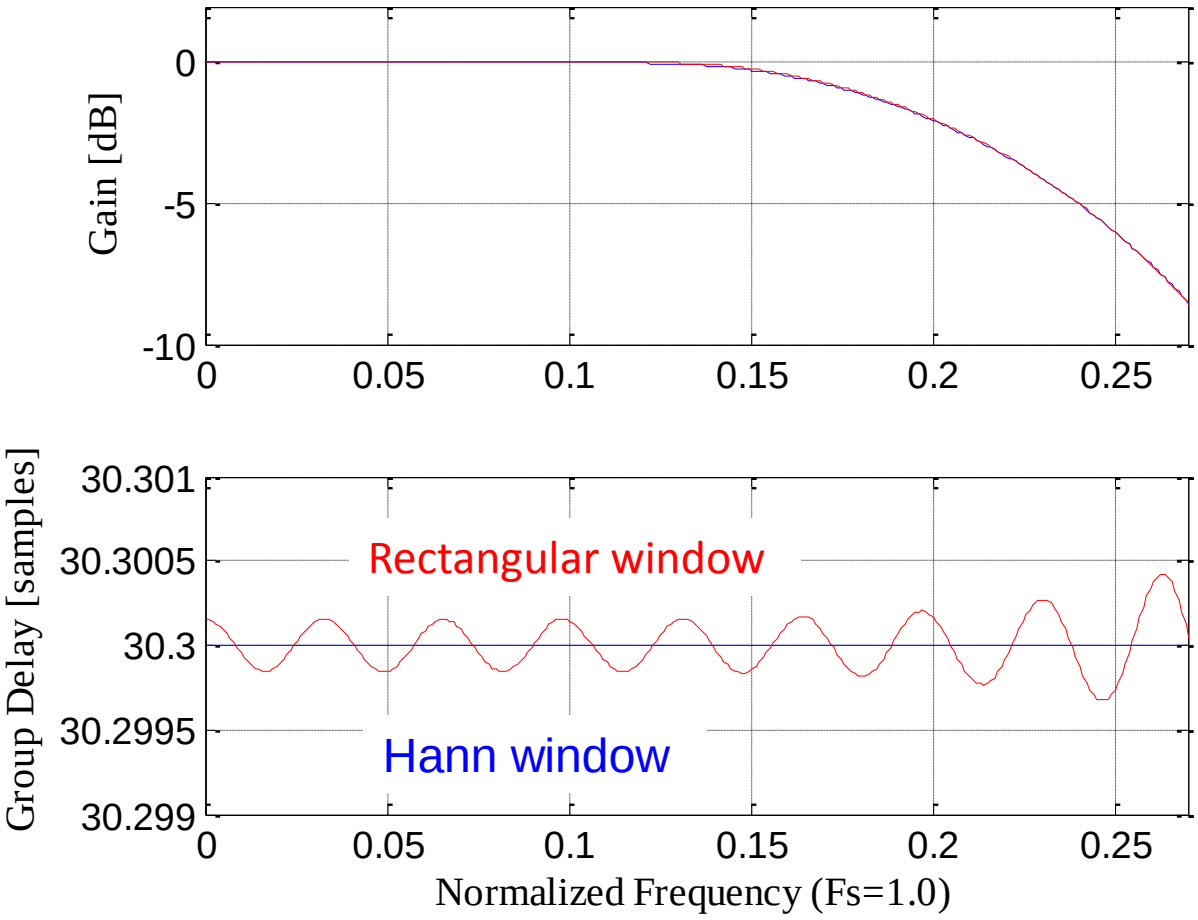
Finite Tap Truncation of Proposed Delay Filter



Rectangular window

Hann window

Effects of Window



Gibbs oscillation of group delay

Frequency characteristics of delay filter with 61-tap truncation

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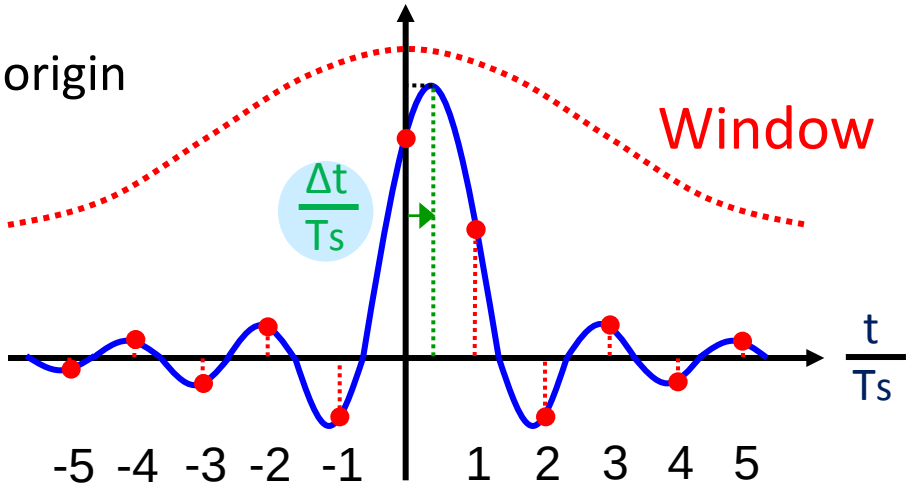
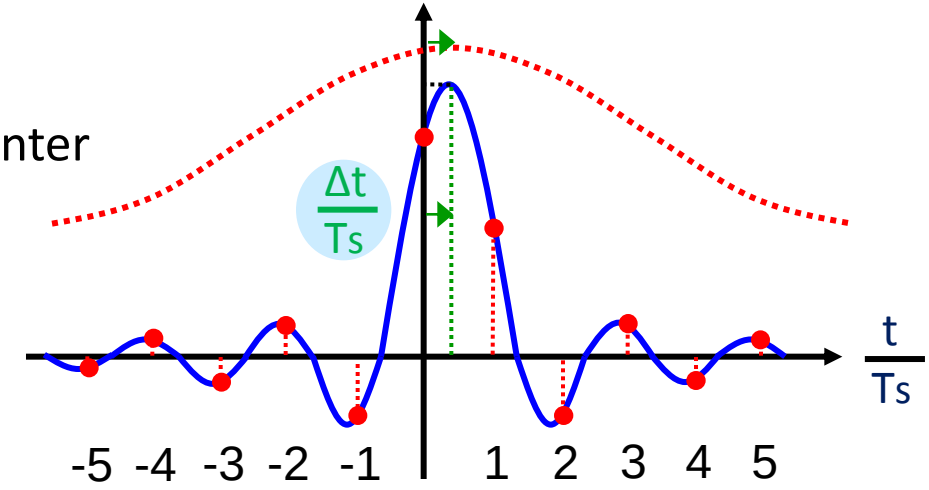
How to Apply Window

Centered at origin



Centered at
impulse response center

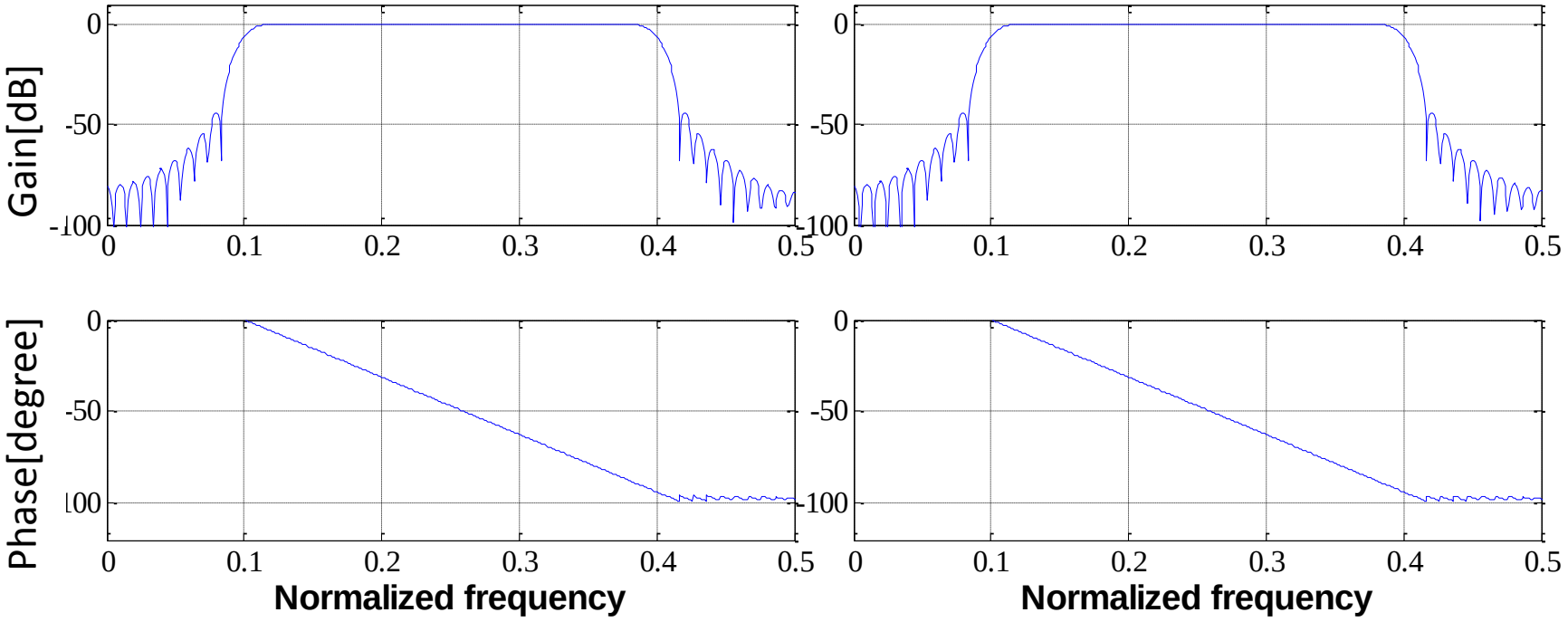


Window center shifted by Δt

Frequency Characteristics of Delay Filter after Applying Window

Window centered at origin

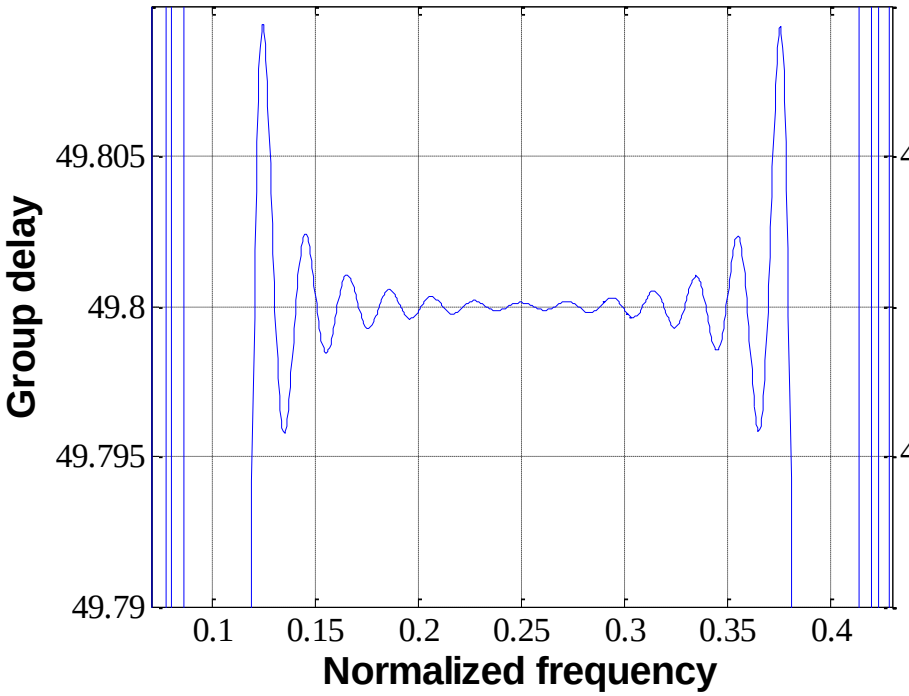
Window centered at impulse response



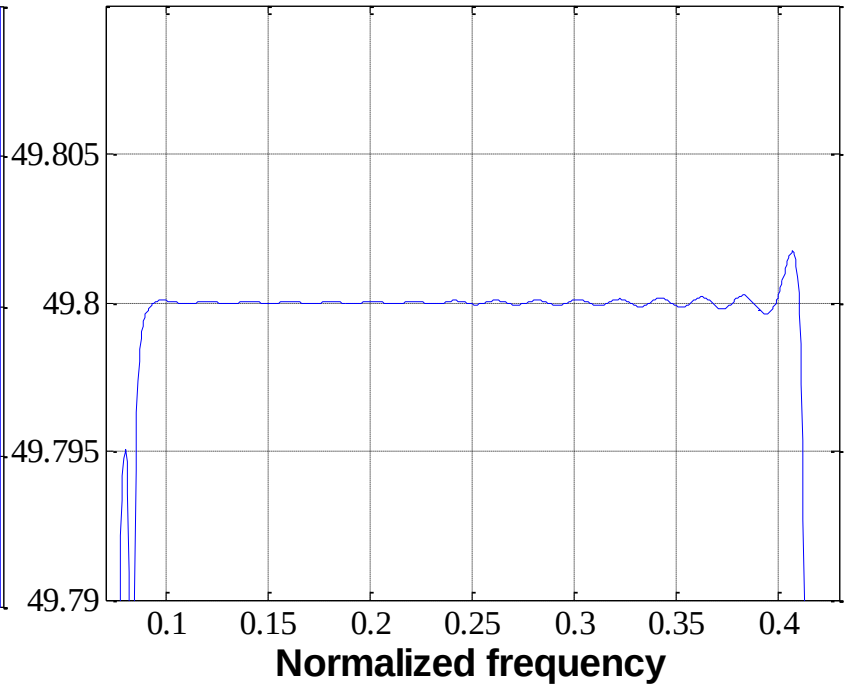
Delay	0.3 samples
Filter Tap	100 taps
Window	Han
Pass band	(0.1～0.4)・Fs
FFT points	1024 points

Group Delay Characteristics of Delay Filter after Applying Window

Window centered at origin



Window centered at impulse response

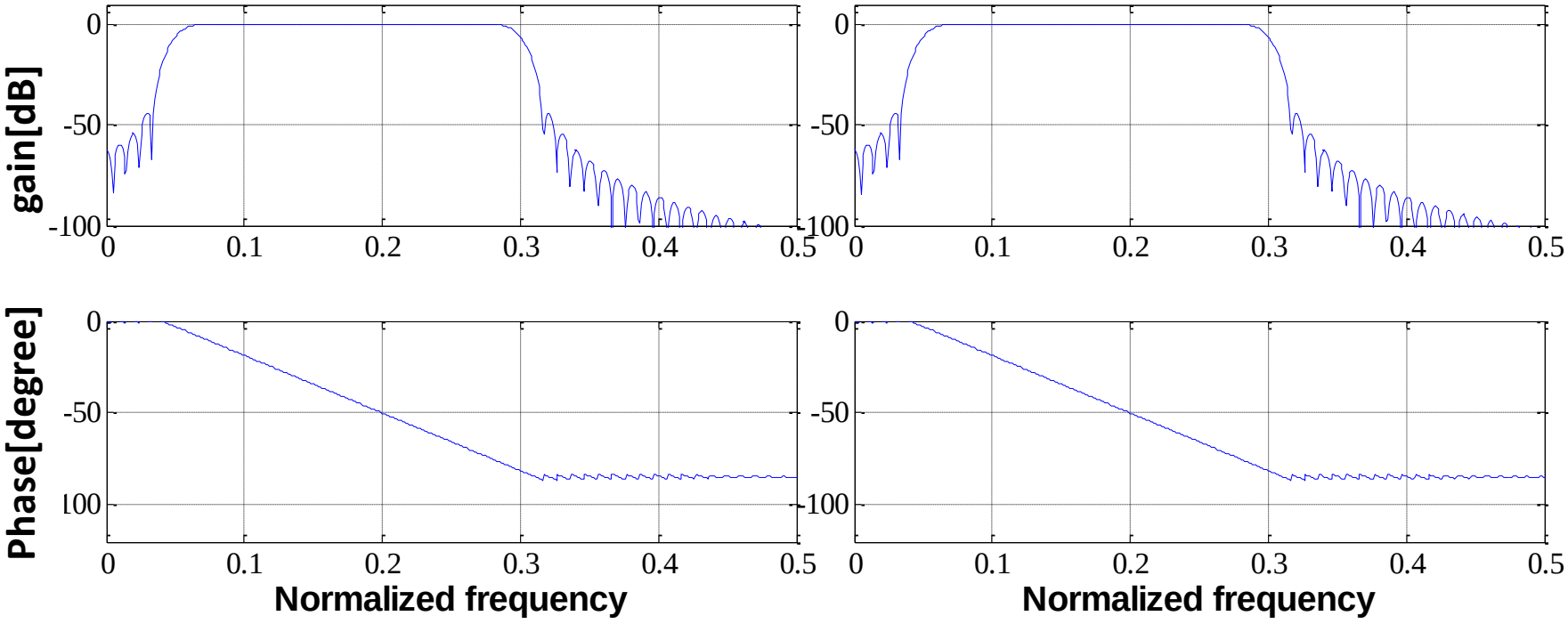


Delay	0.3 samples
Filter Tap	100 taps
Window	Han
Pass band	(0.1～0.4)･Fs
FFT points	1024 points

Frequency Characteristics of Delay Filter after Applying Window

Window centered at origin

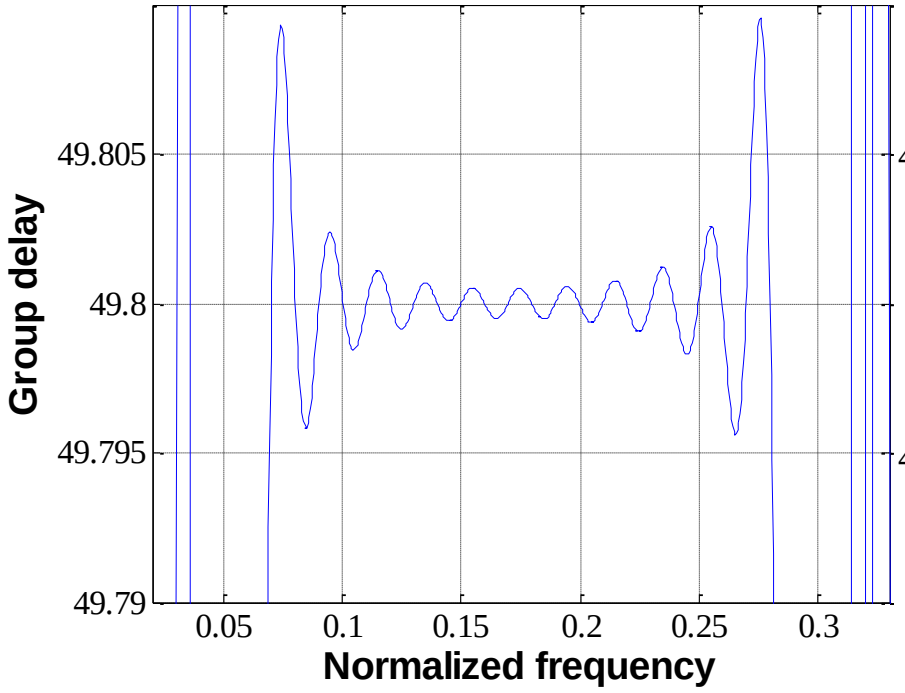
Window centered at impulse response



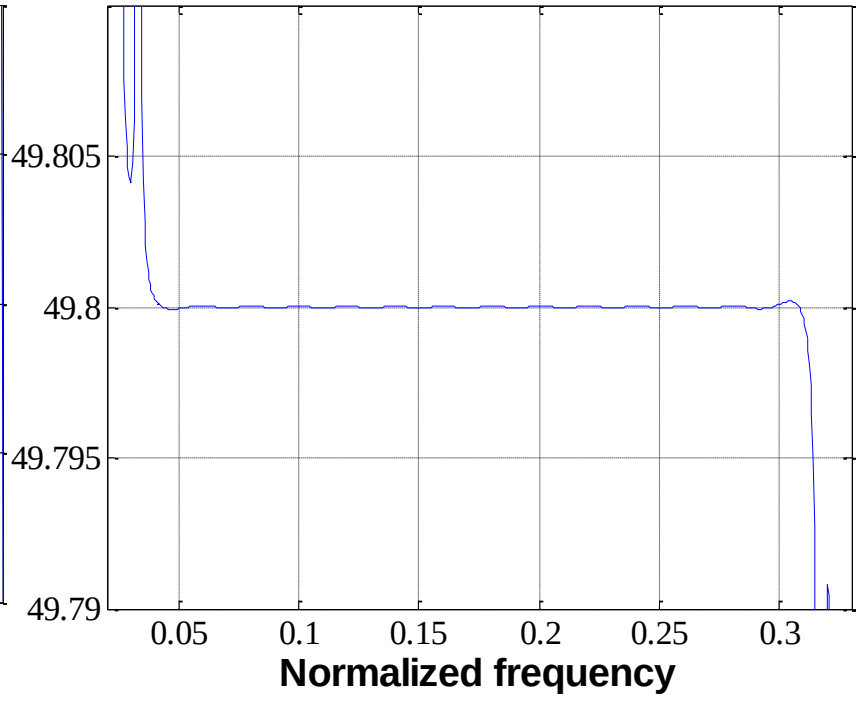
Delay	0.3 samples
Filter Tap	100 taps
Window	Han
Pass band	(0.05～0.3)・Fs
FFT points	1024 points

Group Delay Characteristics of Delay Filter after Applying Window

Window centered at origin



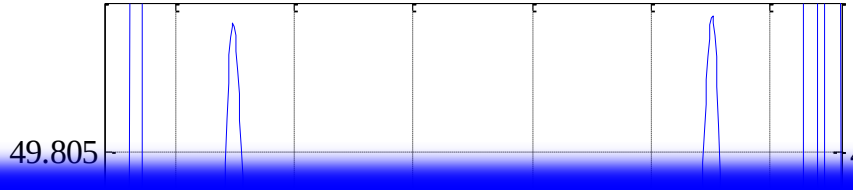
Window centered at impulse response



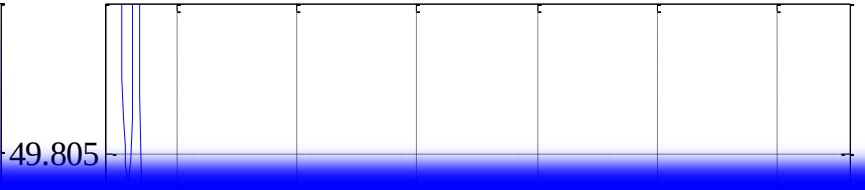
Delay	0.3 samples
Filter Tap	100 taps
Window	Han
Pass band	(0.05～0.3)・Fs
FFT points	1024 points

Group Delay Characteristics of Delay Filter after Applying Window

Window centered at origin



Window centered at impulse response



Applying window centered at impulse response

Constant group delay over entire passband

Normalized frequency

Normalized frequency

Delay	0.3 samples
Filter Tap	100 taps
Window	Han
Pass band	(0.05～0.3)・Fs
FFT points	1024 points

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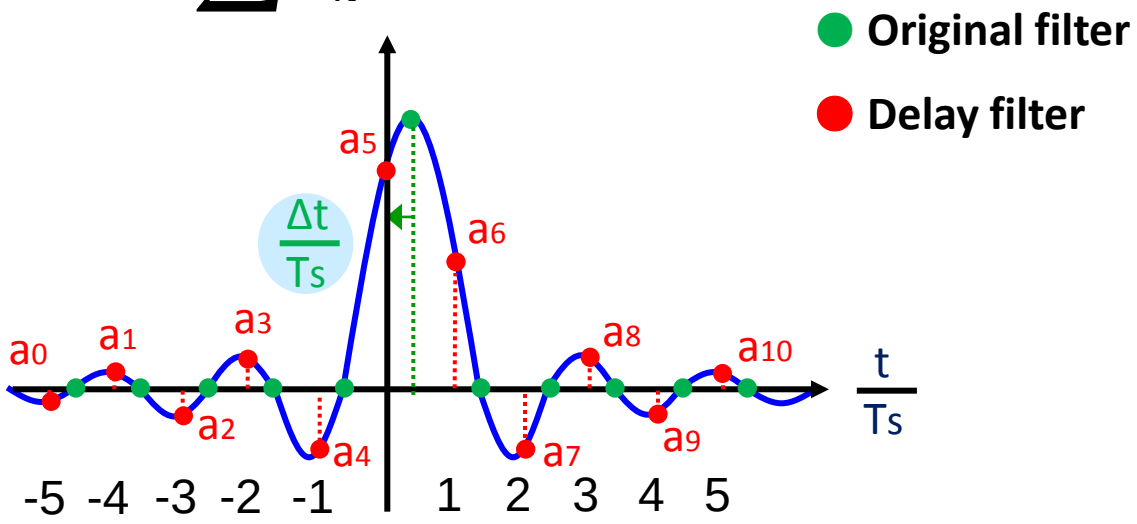
Proposed Filter DC Gain Adjustment

Digital filter DC gain : $\sum a_n$

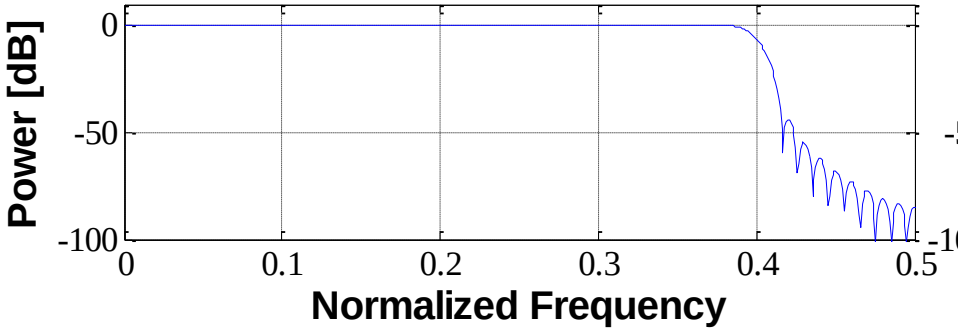


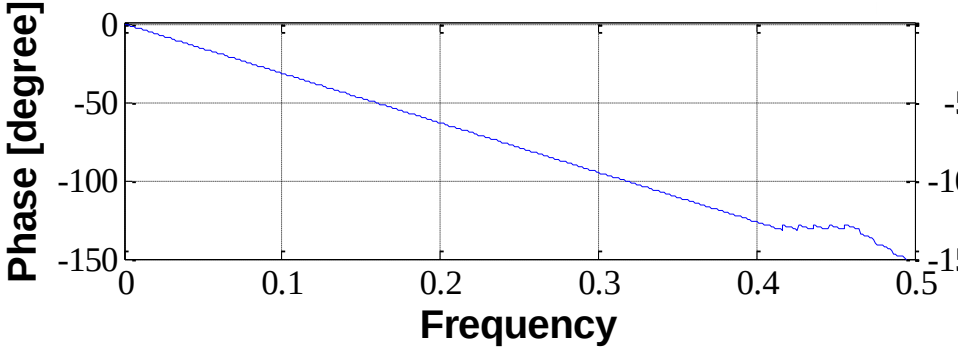
DC gain adjustment due to finite tap truncation is required

$$\sum_{n=0}^{N} a'_n = \text{DC gain of original FIR filter}$$

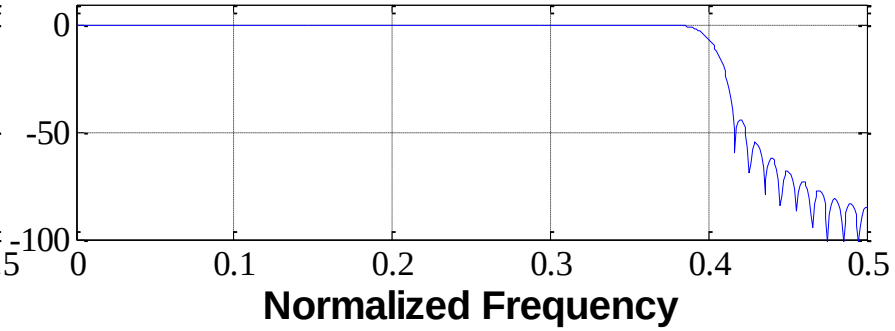
Frequency Characteristics of Proposed Delay Filter

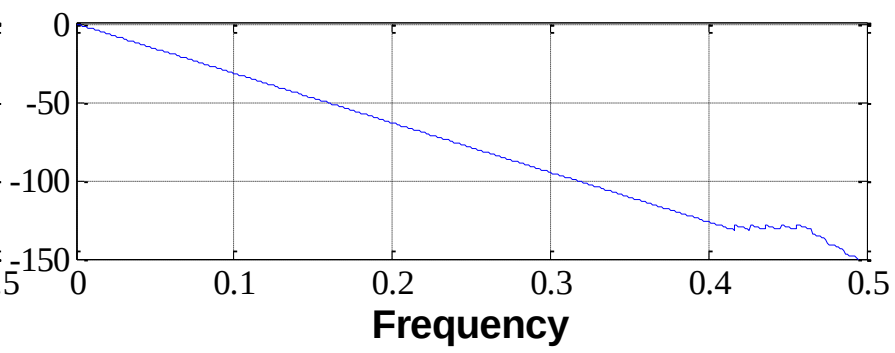
Without DC gain adjustment





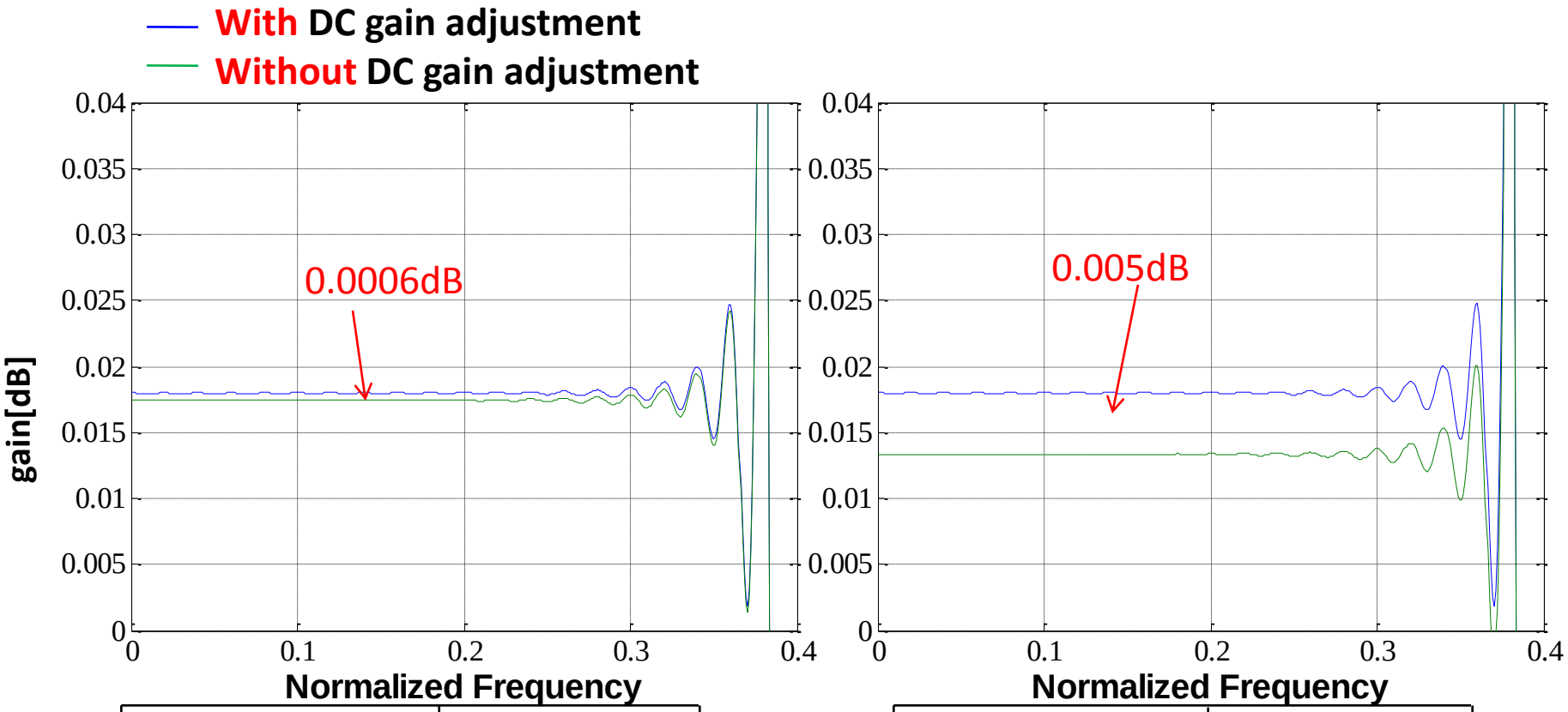
With DC gain adjustment





Delay	0.3 samples
Filter Tap	101 taps
Window	Han
Cut-off Freq.	0.4・Fs
FFTpoints	1024 points

Gain Characteristics of Proposed Delay Filter

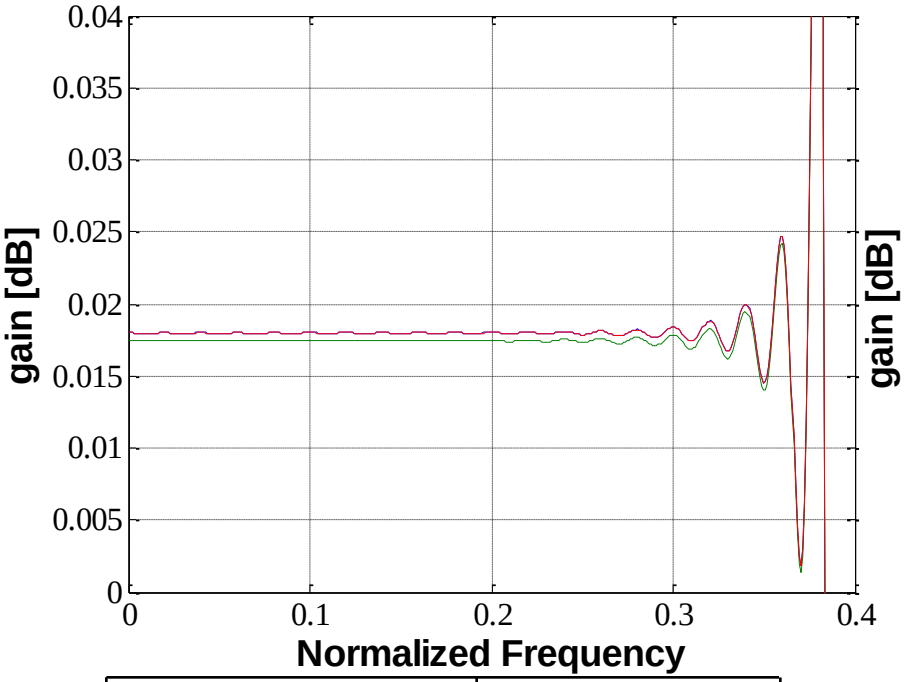


Delay	0.1samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	0.4・Fs
FFTpoints	1024 points

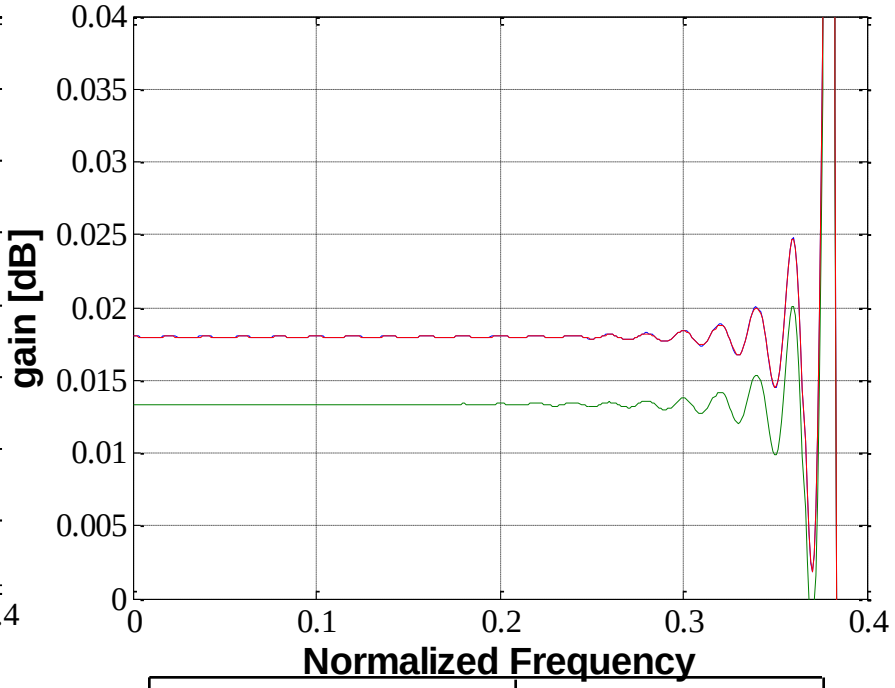
Delay	0.3 samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	0.4・Fs
FFT points	1024 points

Gain Characteristics of Proposed Delay Filter

Original FIR filter
With DC gain adjustment
Without DC gain adjustment



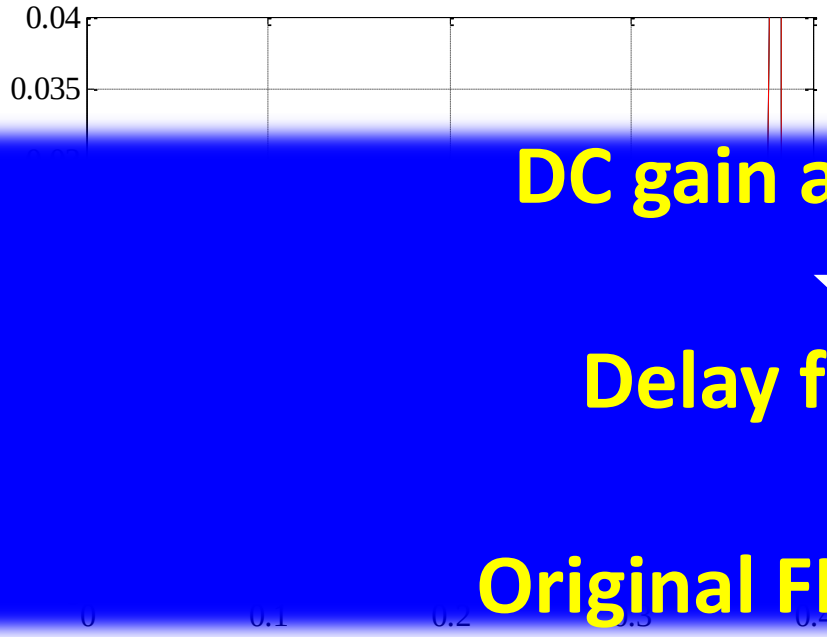
Delay	0.1samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	0.4・Fs
FFTpoints	1024 points

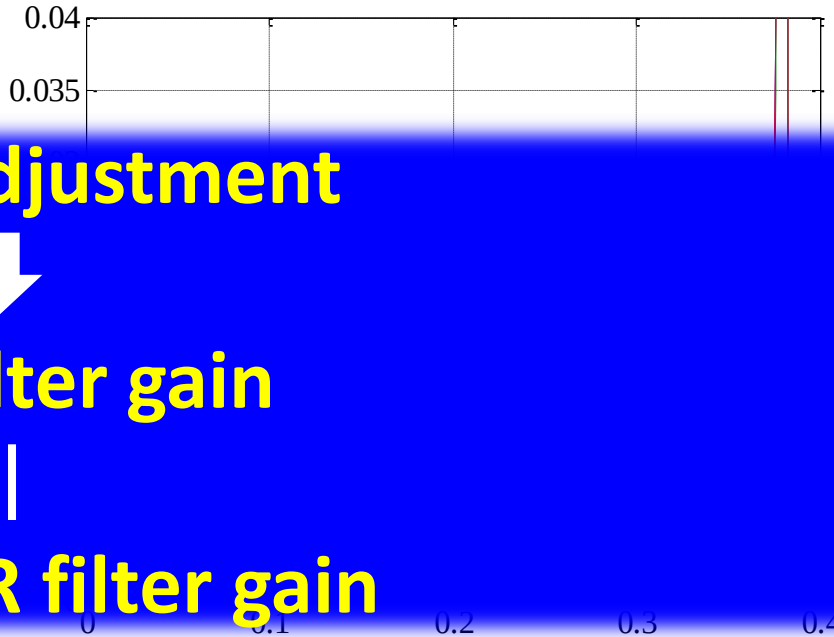


Delay	0.3 samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	0.4・Fs
FFT points	1024 points

Gain Characteristics of Proposed Delay Filter

Original FIR filter
With DC gain adjustment
Without DC gain adjustment





DC gain adjustment

↓

Delay filter gain

||

Original FIR filter gain

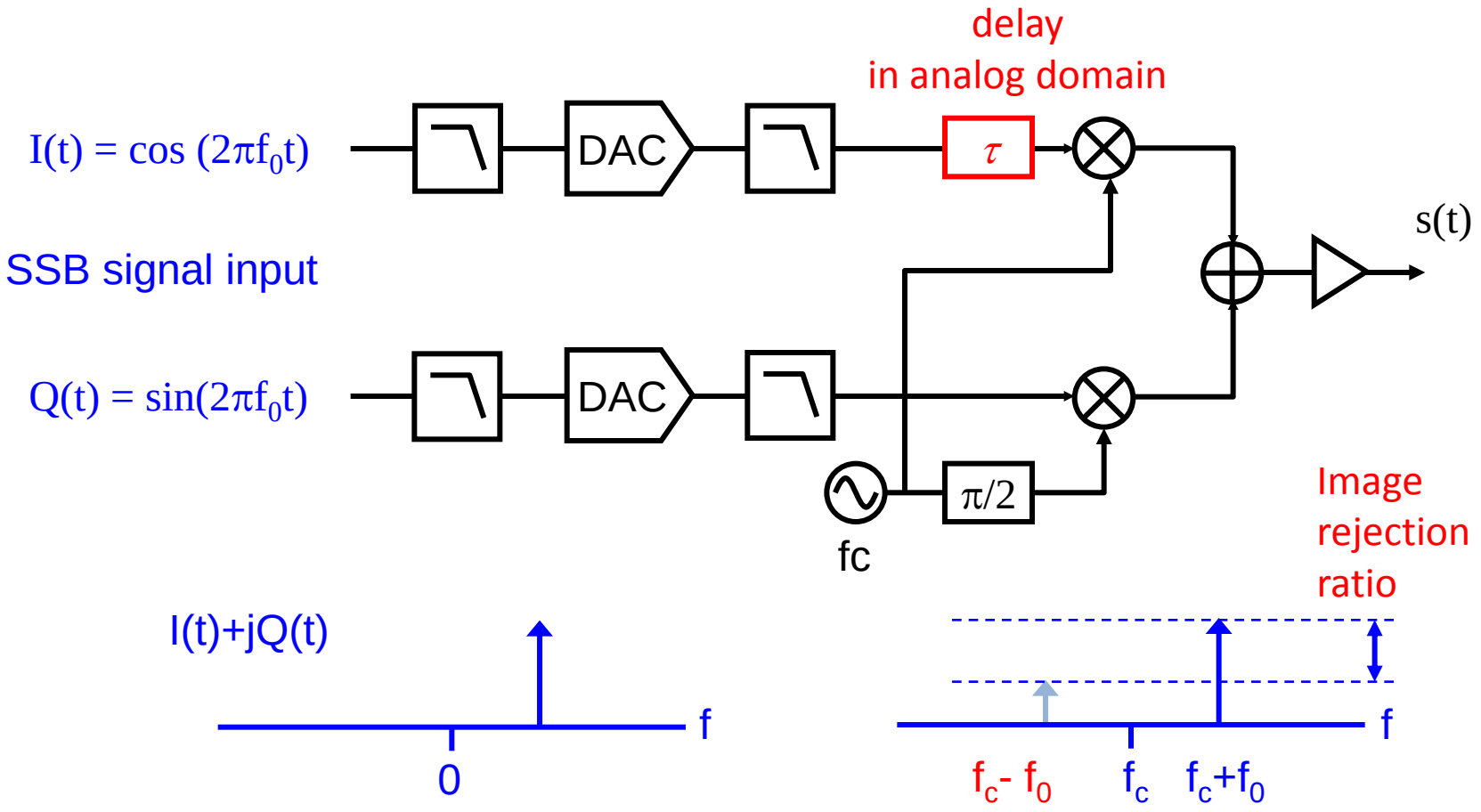
Delay	0.1samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	0.4・Fs
FFTpoints	1024 points

Delay	0.3 samples
Filter Tap	101 taps
Window	Han
Cutoff Freq.	0.4・Fs
FFT points	1024 points

Contents

- Research Purpose
- Conventional Linear Phase Digital Filter Condition
- New Linear Phase Digital Filter Condition
  - Time-Shift, Impulse Response of Ideal Filter
  - New Linear Phase Digital Filter
- MATLAB Simulation
- Design Consideration
  - Window
  - Gain Adjustment
- Application
- Conclusion

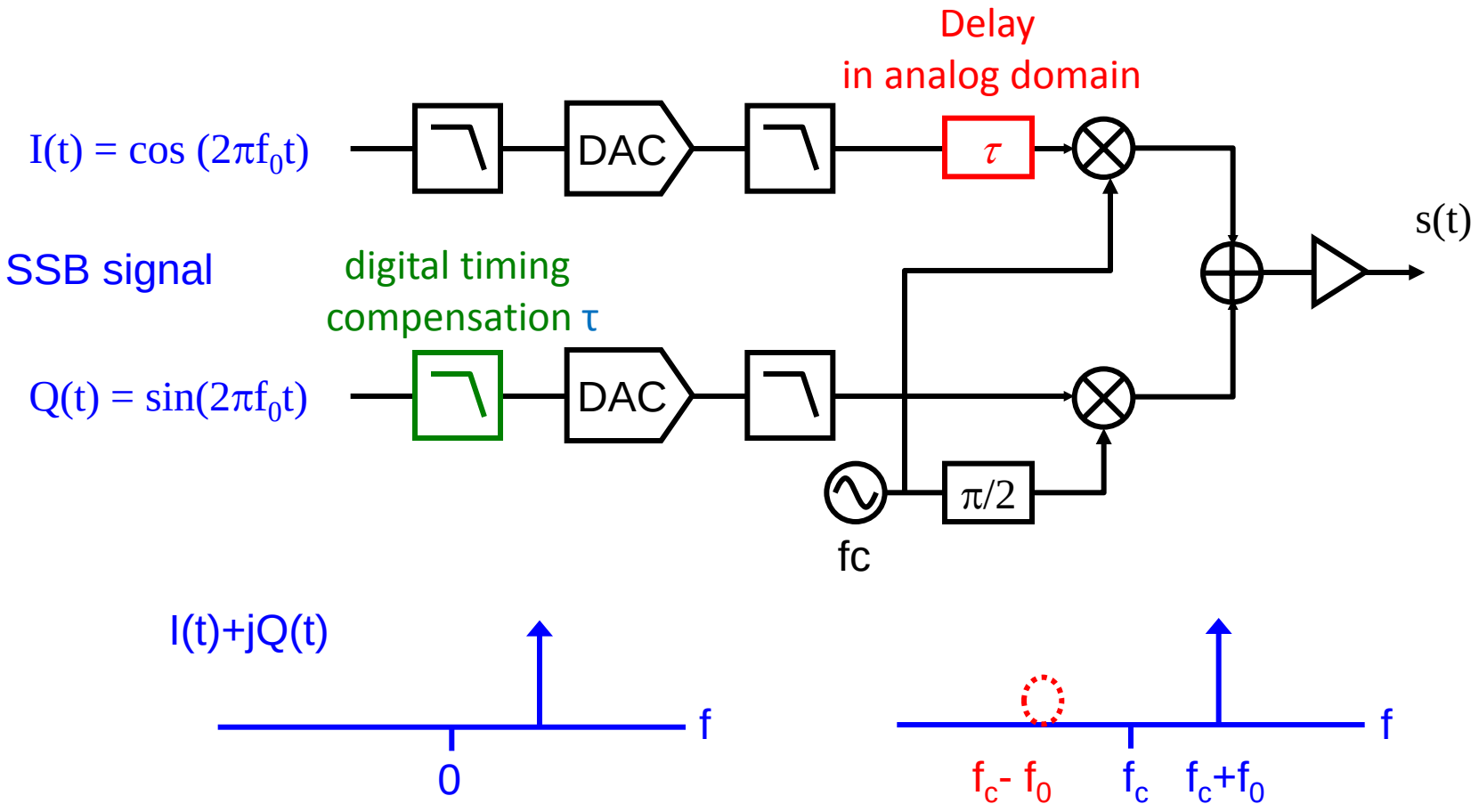
I/Q Delay Mismatch in Quadrature Modulator



SSB：single side band

DAC：digital-to-analog converter

I/Q Delay Mismatch Compensation in Quadrature Modulator

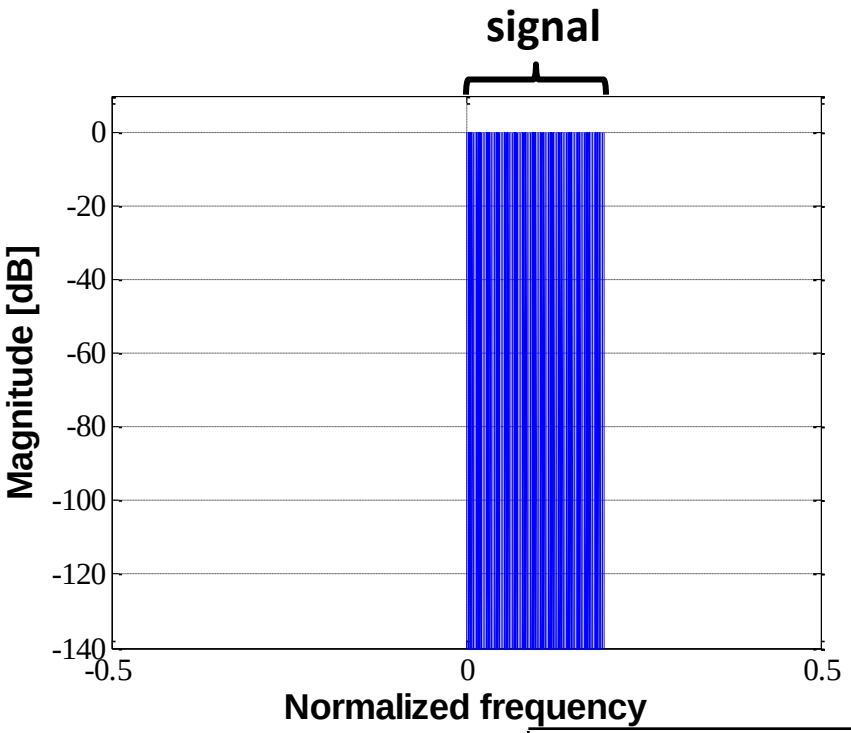


SSB：single side band

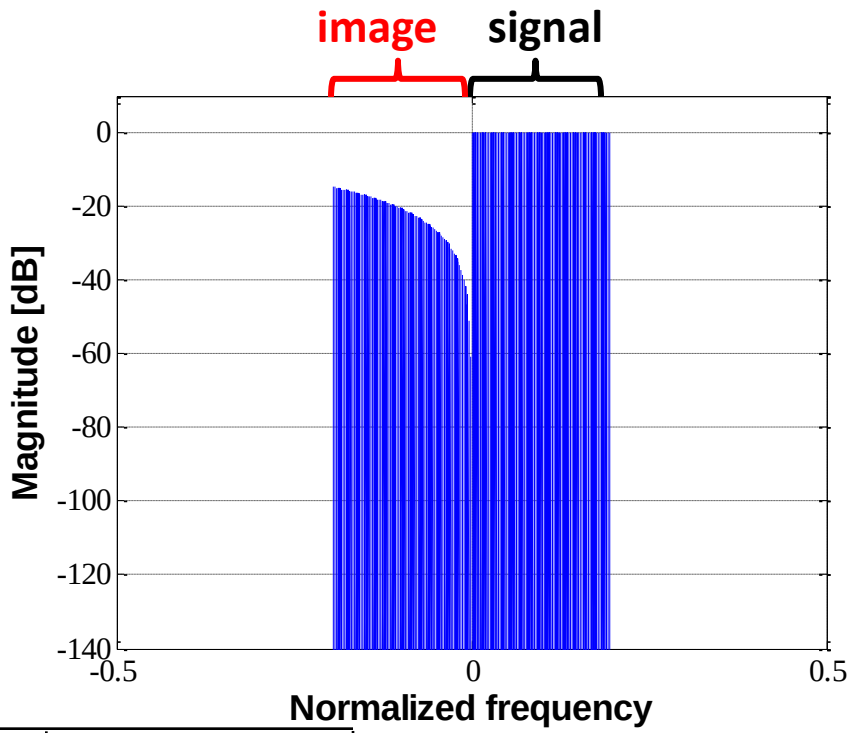
DAC：digital-to-analog converter

Matlab Simulation Results

(a) Ideal case



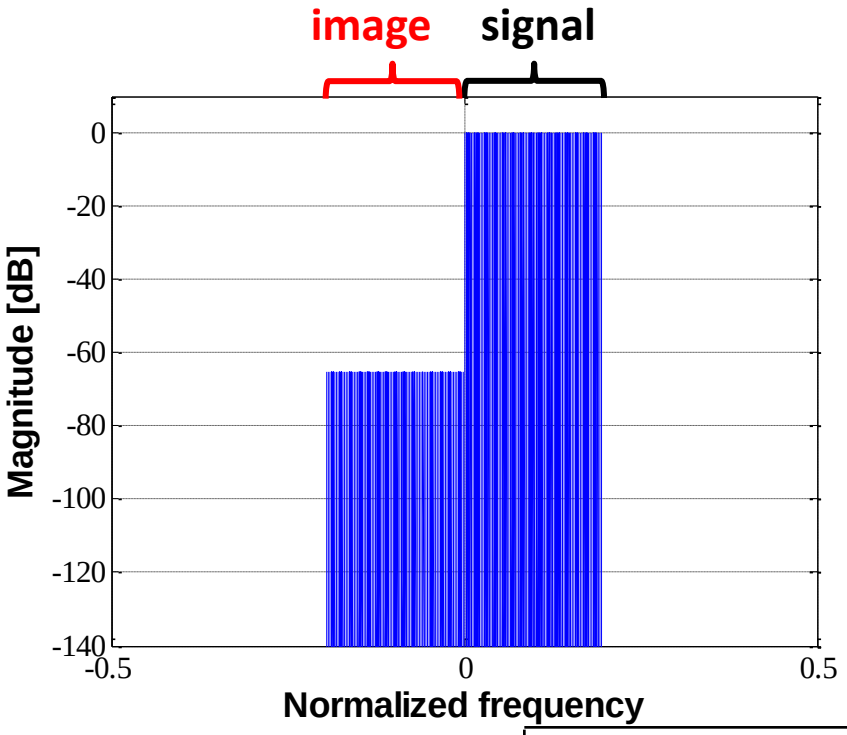
(b) Timing skew case



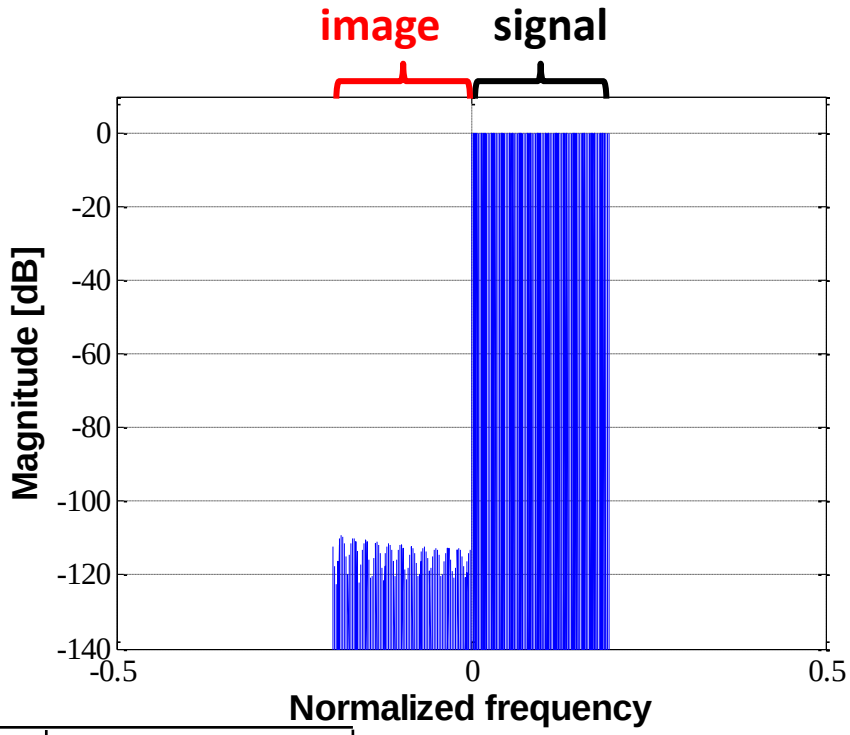
Delay	0.3 samples
Filter tap #	61 taps
Window	Han
FFT points	1024 points

Matlab Simulation Results

(c) Compensation using delay filter
Without adjustment of window, gain



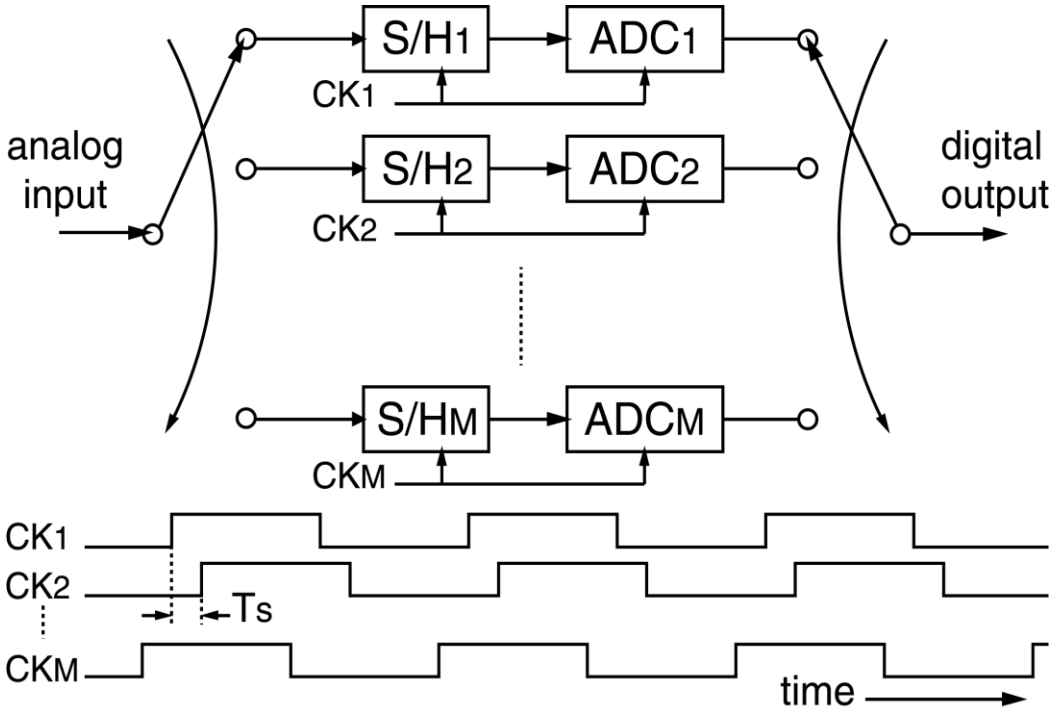
(d) Compensation using delay filter
With adjustment of window, gain



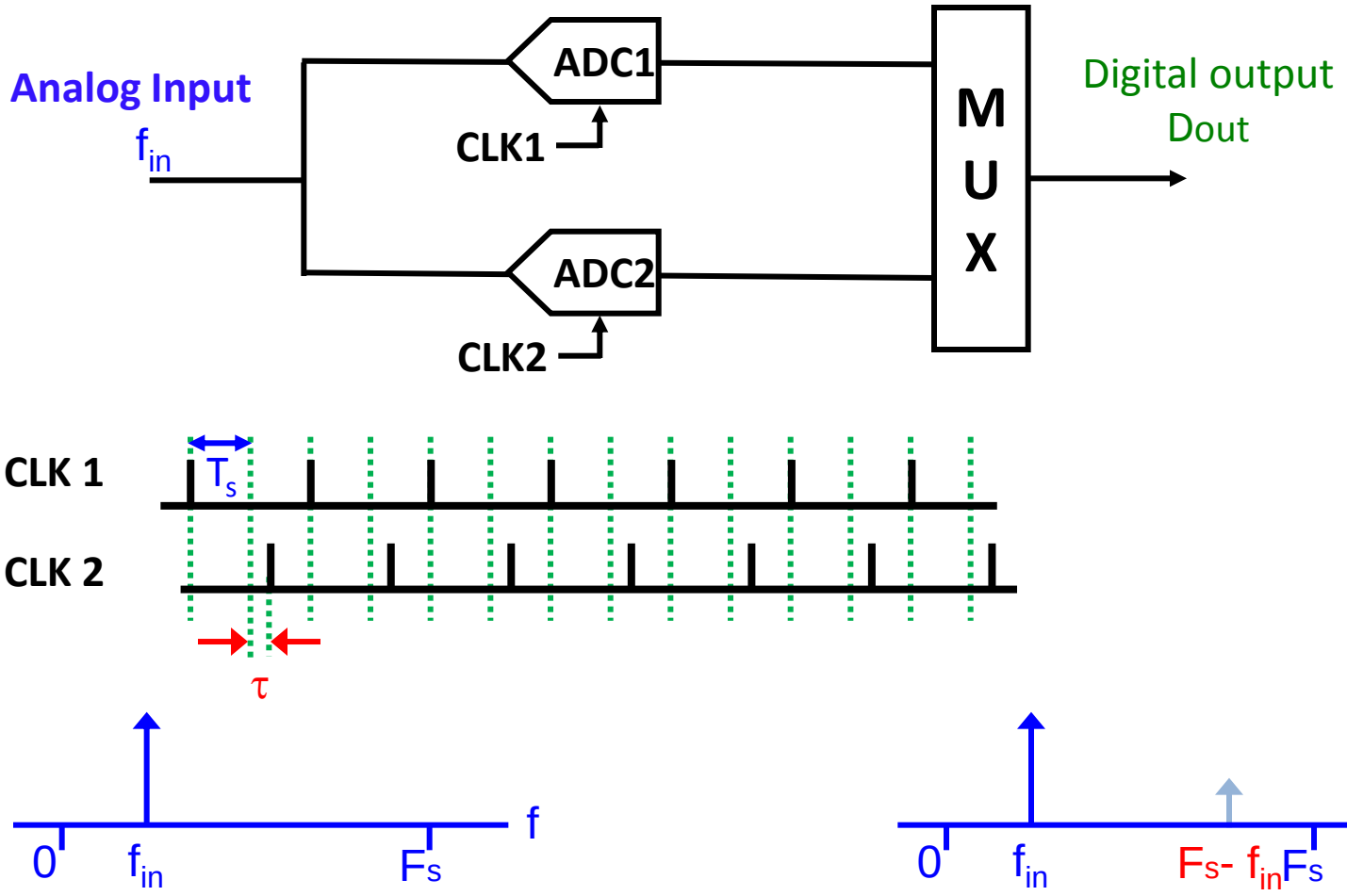
Delay	0.3 samples
Filter tap	61 taps
Window	Han
FFT points	1024 points

Interleaved ADC System

- M channel ADCs ➡ M-times sampling rate



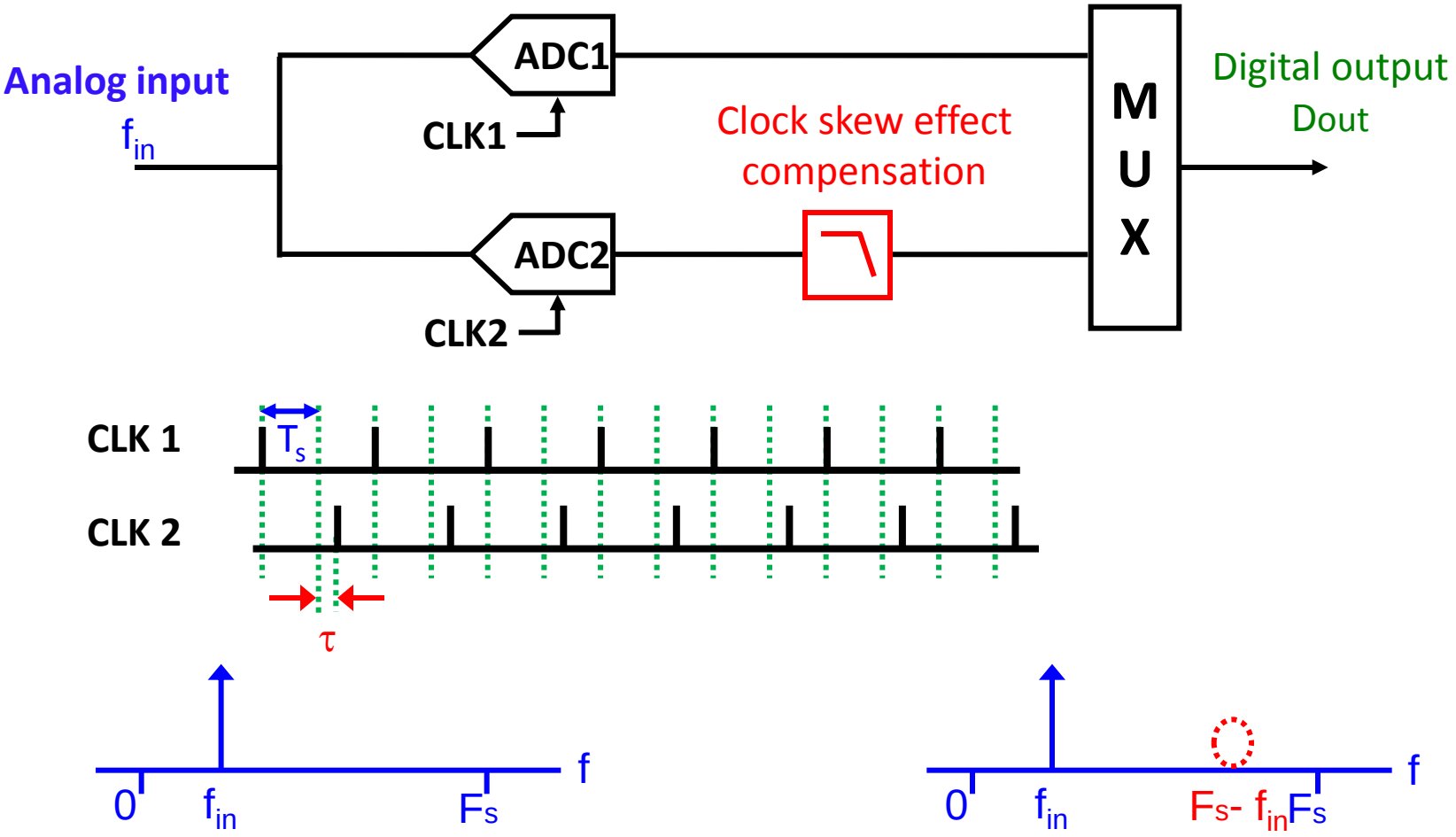
Timing Skew in Interleaved ADC System



ADC：analog-to-digital converter

$F_s=1/T_s$

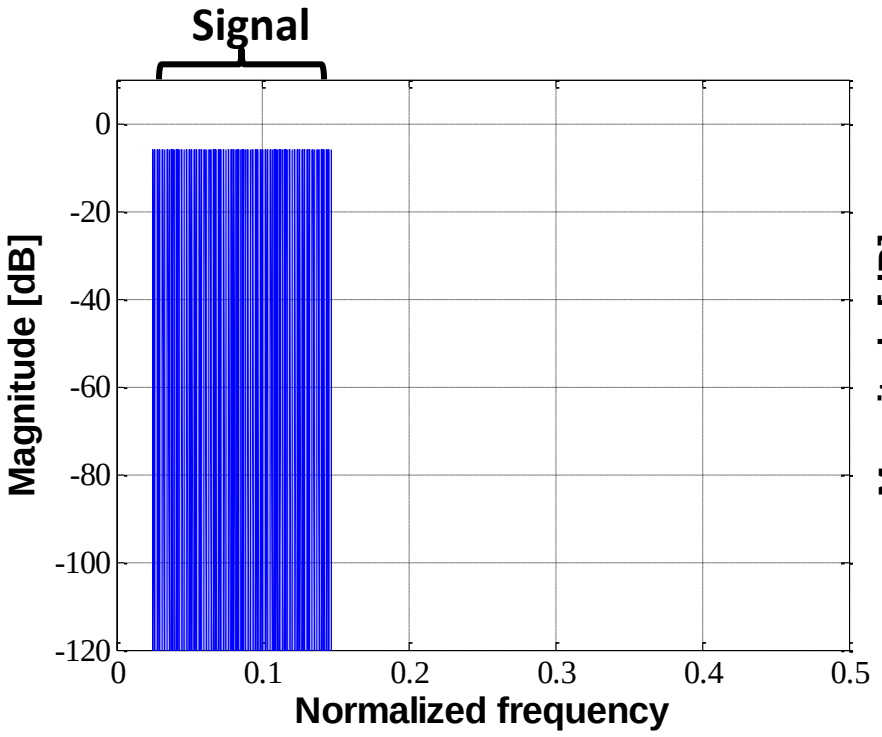
Timing Skew Compensation in Interleaved ADC System



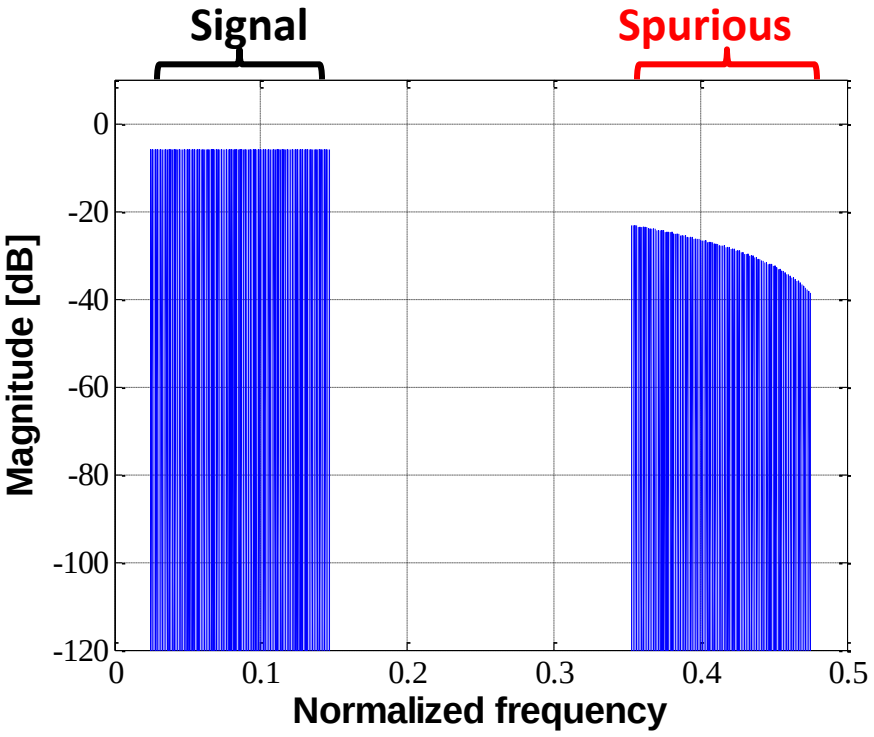
ADC：analog-to-digital converter

Matlab Simulation Results

(a) Ideal case



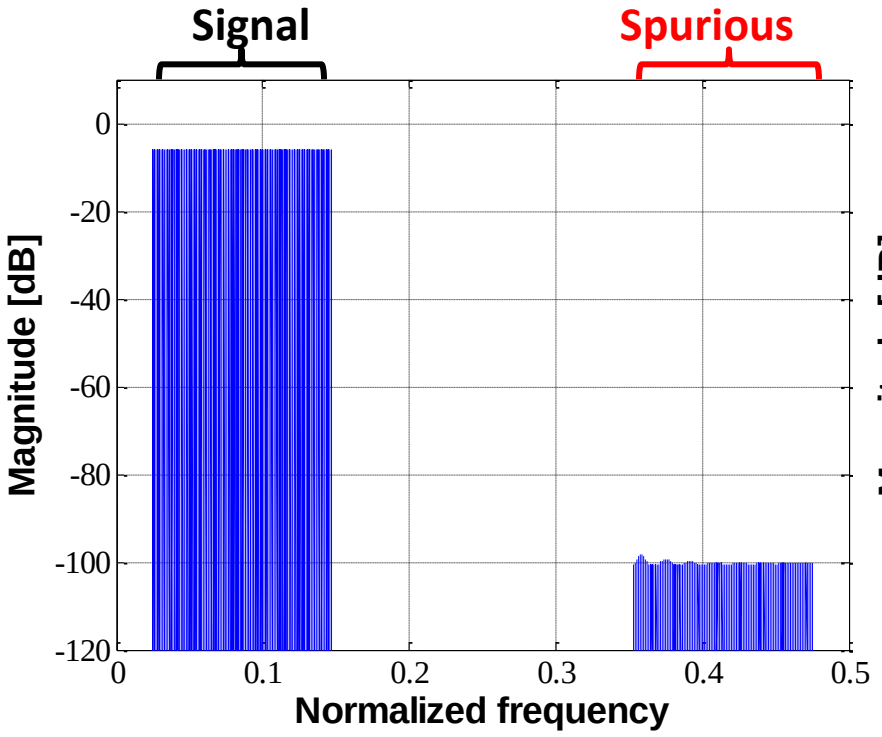
(b) Timing skew case



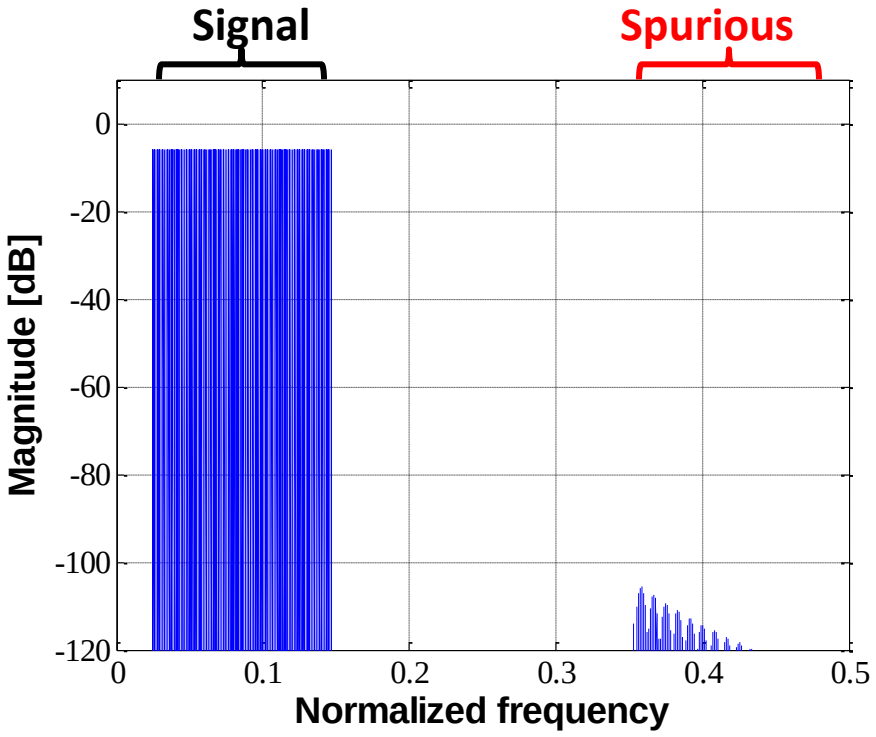
Delay	0.3 samples
Filter tap	61 taps
Window	Han
FFTpoints	1024 points

Matlab Simulation Results

(c) Compensation using delay filter
Without adjustment of window, gain



(d) Compensation using delay filter
With adjustment of window, gain



Delay	0.3 samples
Filter tap	61 taps
Window	Han
FFTpoints	1024 points

Conclusion

- Linear phase digital filter
  with fine time resolution of group delay
- Design consideration
  - How to apply window
  - DC gain adjustment
- Application Examples
  - I/Q delay mismatch compensation
    in quadrature modulator
  - Timing skew compensation in interleaved ADC system

On-going work

- Implementation issues
  - Finite word length, finite tap effects
  - LSI implementation