

P55 Consideration of Uncertainty Principle in Sampling Circuit

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Introduction

Research Objective

Background

Waveform acquisition → Sampling
Sampling circuit → Many non-idealities

Aperture time
Serious for high frequency signal

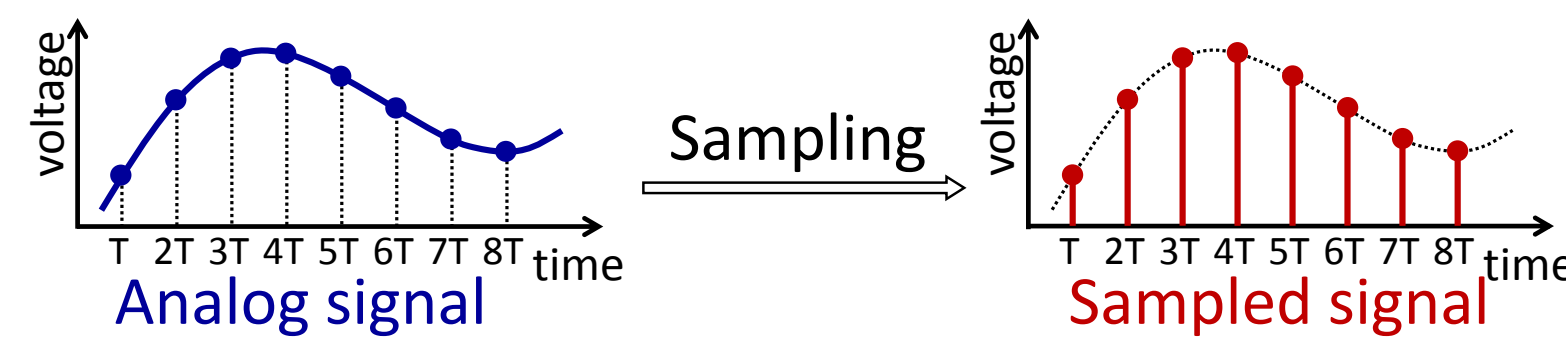
Objective

- Sampling circuit with aperture time
- Uncertainty principle

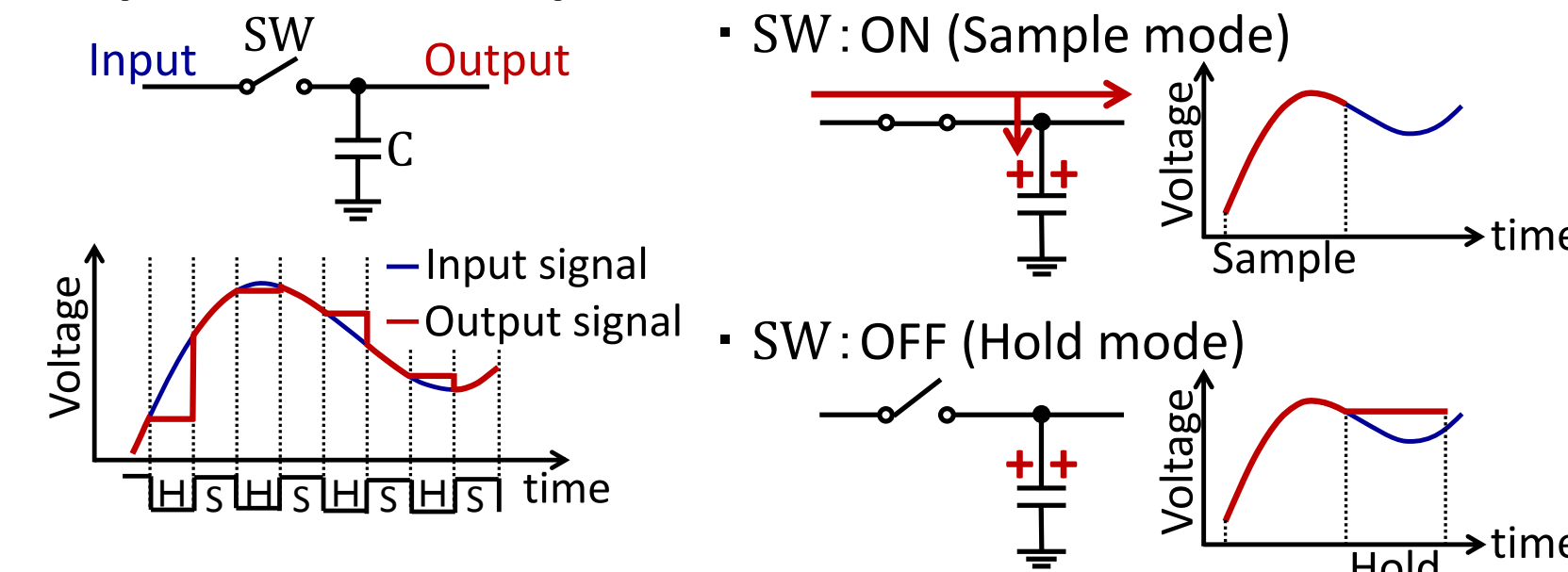
Theoretical foundation for signal acquisition

Sampling

Sampling



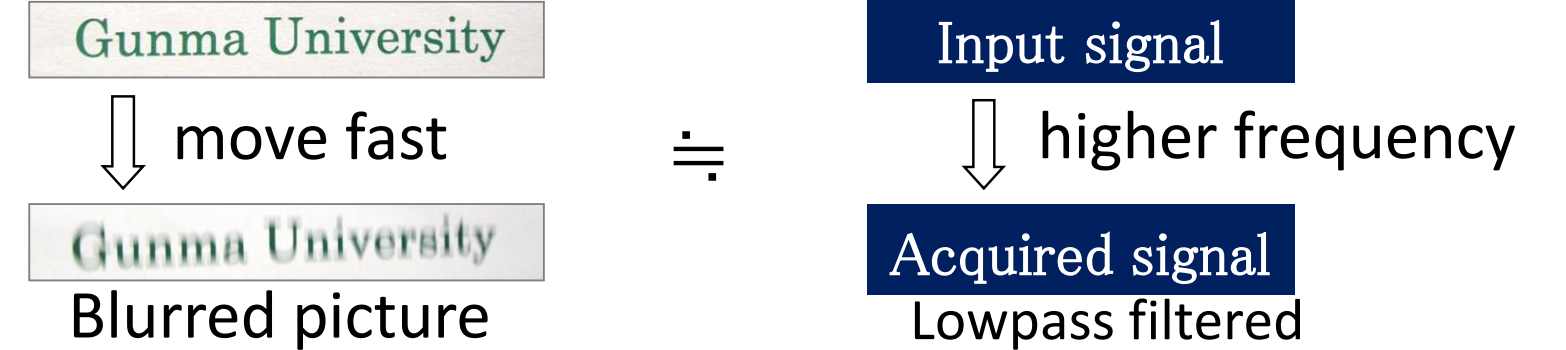
Operation of Sample/Hold circuit



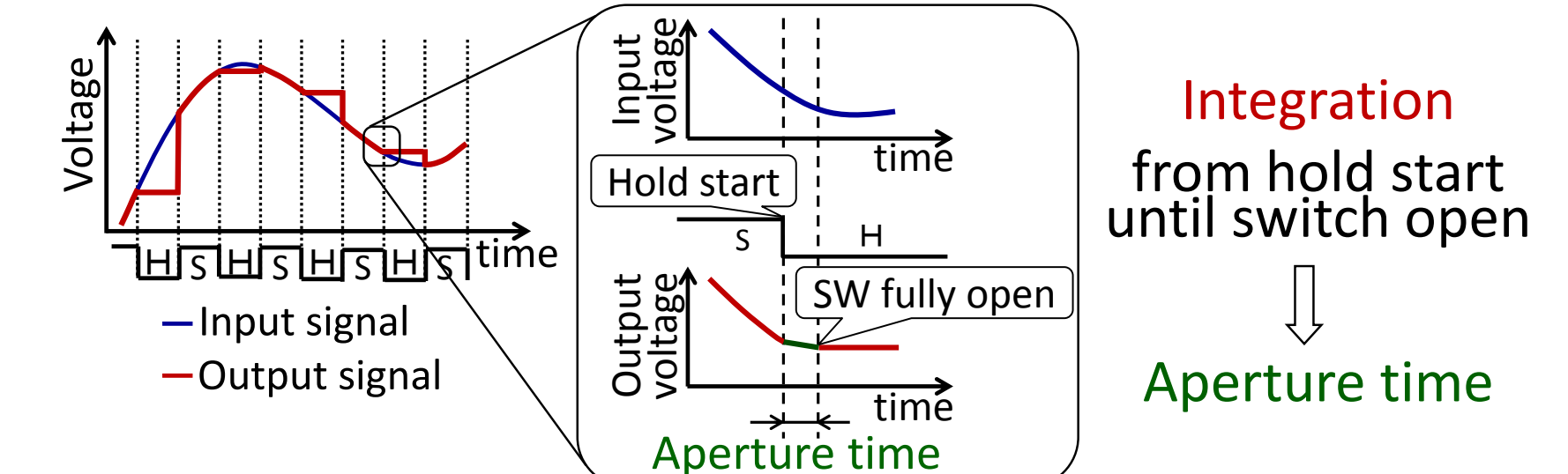
Aperture Time

Visual understanding

ex) Shutter speed of camera



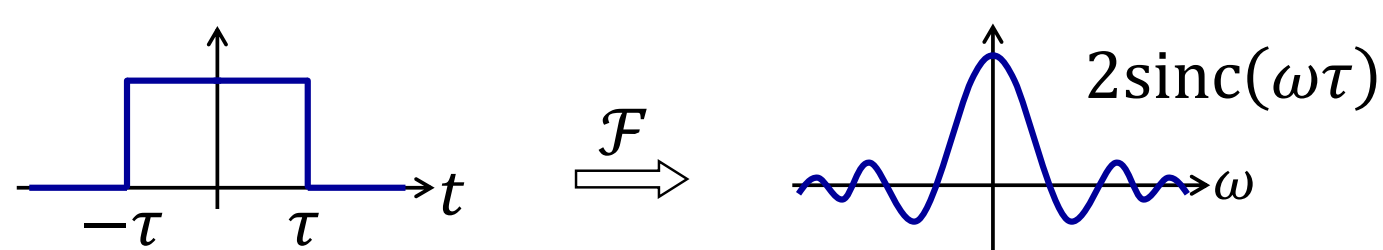
Definition



Uncertainty Principle

Uncertainty Principle

Fourier transform of rectangular pulse

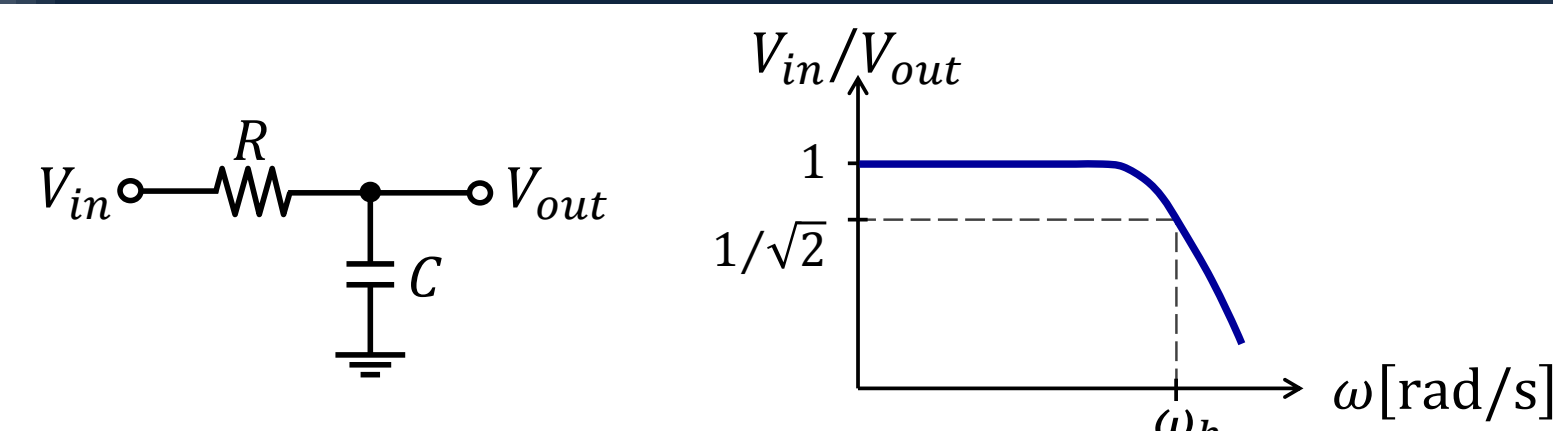


Pulse width : σ_t → Spectrum width : σ_ω
Narrow → Wide
Wide → Narrow

$$\sigma_t \sigma_\omega \geq \frac{1}{2} \quad \text{Uncertainty Principle}$$

RC Circuit and Transfer function

RC Circuit



$$\omega_h = \frac{1}{\tau_1}$$

τ_1 : Time constant RC
 ω_h : Cutoff frequency

Time constant : τ_1 → Pass band : ω_h
Small → Wide
Large → Narrow

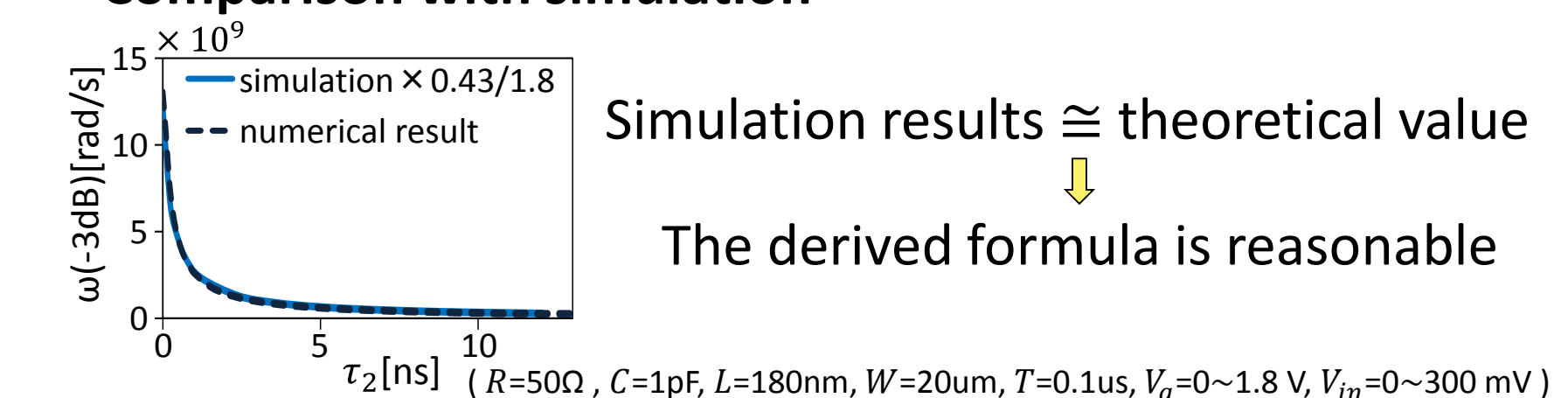
Transfer Function for RC Sampling Circuit with Aperture Time

Formula considering aperture time

$$\frac{V_C}{V_{in}} = \frac{\text{sinc}(\omega\tau_2)}{\text{sinc}(\omega\tau_2) + j\omega\tau_1}$$

$\tau_1 = RC$
 $\tau_2 = \text{Aperture time}$

Comparison with simulation



Uncertainty Principle and Sampling Circuit

Derivation Transfer Function of Gain=-3[dB]

$$20 \log \left| \frac{V_C}{V_{in}} \right| = 20 \log \sqrt{\frac{1}{2}}$$

$$\left| \frac{V_C}{V_{in}} \right| = \left| \frac{\text{sinc}(\omega\tau_2)}{\text{sinc}(\omega\tau_2) + j\omega RC} \right|$$

$$= \sqrt{\frac{\text{sinc}^2(\omega\tau_2)}{\text{sinc}^2(\omega\tau_2) + (\omega RC)^2}} = \sqrt{\frac{1}{2}}$$

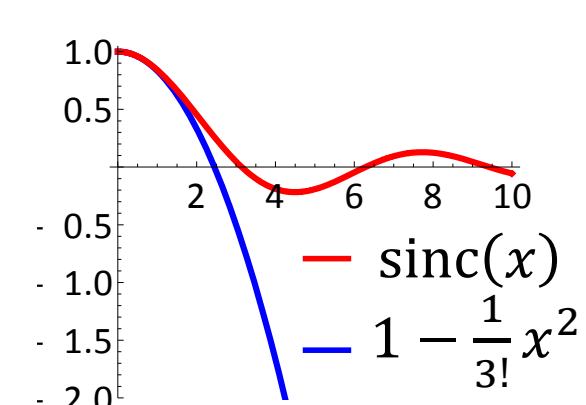
↓

$$\text{sinc}(\omega\tau_2) = \omega\tau_1$$

Application of Taylor Expansion

Taylor expansion of sinc function

$$\text{sinc}(x) = \sum_{n=0}^{\infty} \frac{(-1)^n}{(2n+1)!} x^{2n} \cong 1 - \frac{1}{3!} x^2$$



$$\text{sinc}(x) \geq 1 - \frac{1}{3!} x^2$$

$$\therefore \omega\tau_1 \geq 1 - \frac{1}{3!} (\omega\tau_2)^2$$

To rewrite the standard derivation

$$\omega \rightarrow \sigma_\omega$$

$$\tau_1 \rightarrow \sigma_{\tau_1}$$

$$\sigma_\omega \sigma_{\tau_1} + \frac{1}{6} (\sigma_\omega \tau_2)^2 \geq 1$$

Consideration of $\sigma_\tau \sigma_\omega \geq 1/2$

LPF Design

$$\sigma_\omega \sigma_\tau \geq 1/2$$

σ_ω : bandwidth
 σ_τ : time constant

Narrow bandwidth
||
Time constant increase
Large R, C chip area

High frequency signal sampling

$$\sigma_\omega \sigma_\tau = 1/2$$

σ_ω : bandwidth
 σ_τ : aperture time

Wideband
||
Short aperture time

Uncertainty Relationship with Aperture Time

$$\sigma_\omega \sigma_{\tau_1} + \frac{1}{6} (\sigma_\omega \tau_2)^2 \geq 1$$

σ_ω : bandwidth
 σ_{τ_1} : time constant
 τ_2 : aperture time

High frequency signal sampling

$$\sigma_\omega \sigma_{\tau_1} + \frac{1}{6} (\sigma_\omega \tau_2)^2 = 1$$

RC time constant $\sigma_{\tau_1} > 0$
Obtain the same bandwidth σ_ω
↓
Shorter aperture time τ_2

summary

Conclusion

We have clarified design trade-off among RC time constant, aperture time & bandwidth in sampling circuit with uncertainty principle.

$$\sigma_\omega \sigma_{\tau_1} + \frac{1}{6} (\sigma_\omega \tau_2)^2 = 1$$

σ_ω : bandwidth
 σ_{τ_1} : time constant
 τ_2 : aperture time

Publications & References

Conference presentation

- [1] A.A.Abidi, M. Arai, et. al., "Finite Aperture Time and Sampling Bandwidth", IEICE General Conference (Mar. 2011)
- [2] A.A.Abidi, M. Arai, et. al., "Finite Aperture Time Effects in Sampling Circuit", 24th IEICE Workshop on Circuits and Systems (Aug. 2011)
- [3] M. Arai, "Analysis of Non-ideal Factors of T/H Circuit", 51th System LSI Joint Seminar (Jun. 2012)

References

- [1] T.Tuduki, *The Uncertainty Principle - Challenge to Fate*, Kodansha (Sep. 2002)
- [2] L. Cohen, *Time-Frequency Analysis*, Prentice Hall (1995).

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