

A Study of a Complex Multi-Band Pass $\Delta\Sigma$ D/A Modulator for I,Q signal generation

Shaiful Nizam Bin Mohyar*

Masahiro Murakami, Haruo Kobayashi, Tatsuji Matsuura (Gunma Univ.)

Osamu Kobayashi (STARC)

Kobayashi Laboratory
Depart. of Electronic Eng.,
Graduate Sch. of Eng.
Gunma University

Outline

- Background & motivation
- Approach
- Complex bandpass $\Delta\Sigma$ modulator
- Complex multi-bandpass $\Delta\Sigma$ modulator
- Simulation result
- Conclusion

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- **Background & motivation**
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- Conclusion & Future work

Background & Motivation

Communication devices

- cellular, wireless LAN, blue-tooth, low IF transmitter/receiver (use I,Q signal)



I,Q signal generator is desired

- communication IC testing (receiver)

Important!

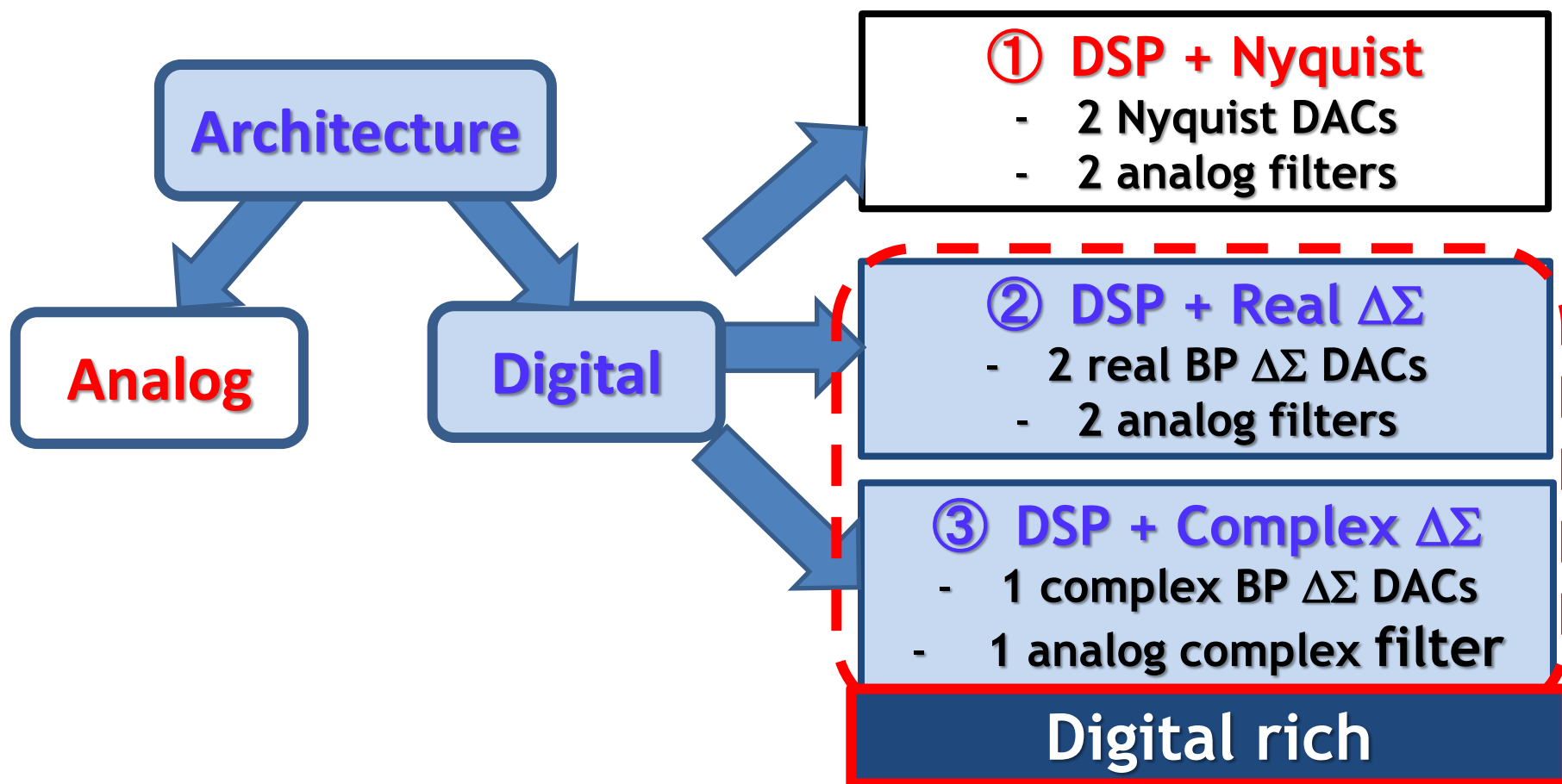


High quality + low cost
I,Q Signal Generator

Outline

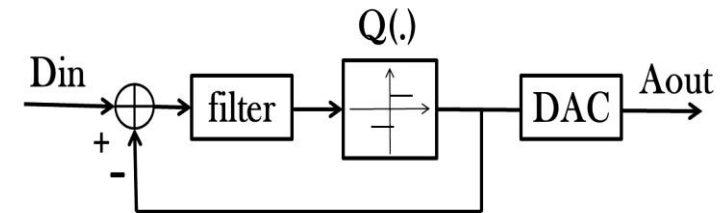
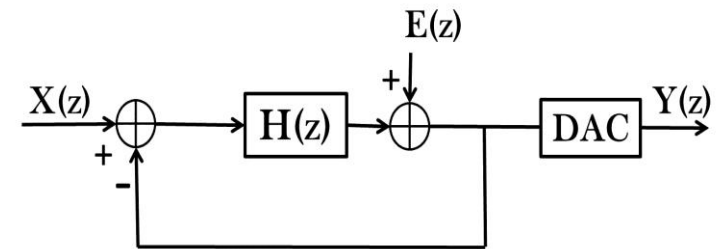
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Approach: I,Q Signal Generation Architecture



Why Delta-Sigma ($\Delta\Sigma$) Modulator?

- Only simple analog circuit
- Easy to design in digital
 - High speed
 - Low power consumption
 - Low cost
- Easy to configure
 - programmable

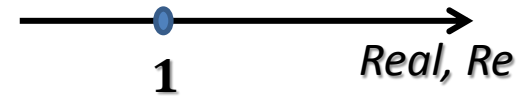


Outline

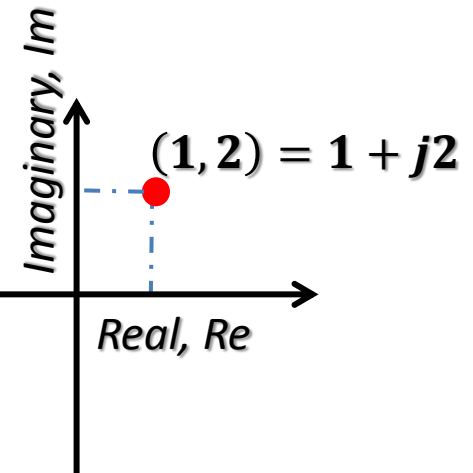
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What is complex number?

Real number



$$- I_{in}(n), Q_{in}(n) \rightarrow I_{out}(n), Q_{out}(n)$$

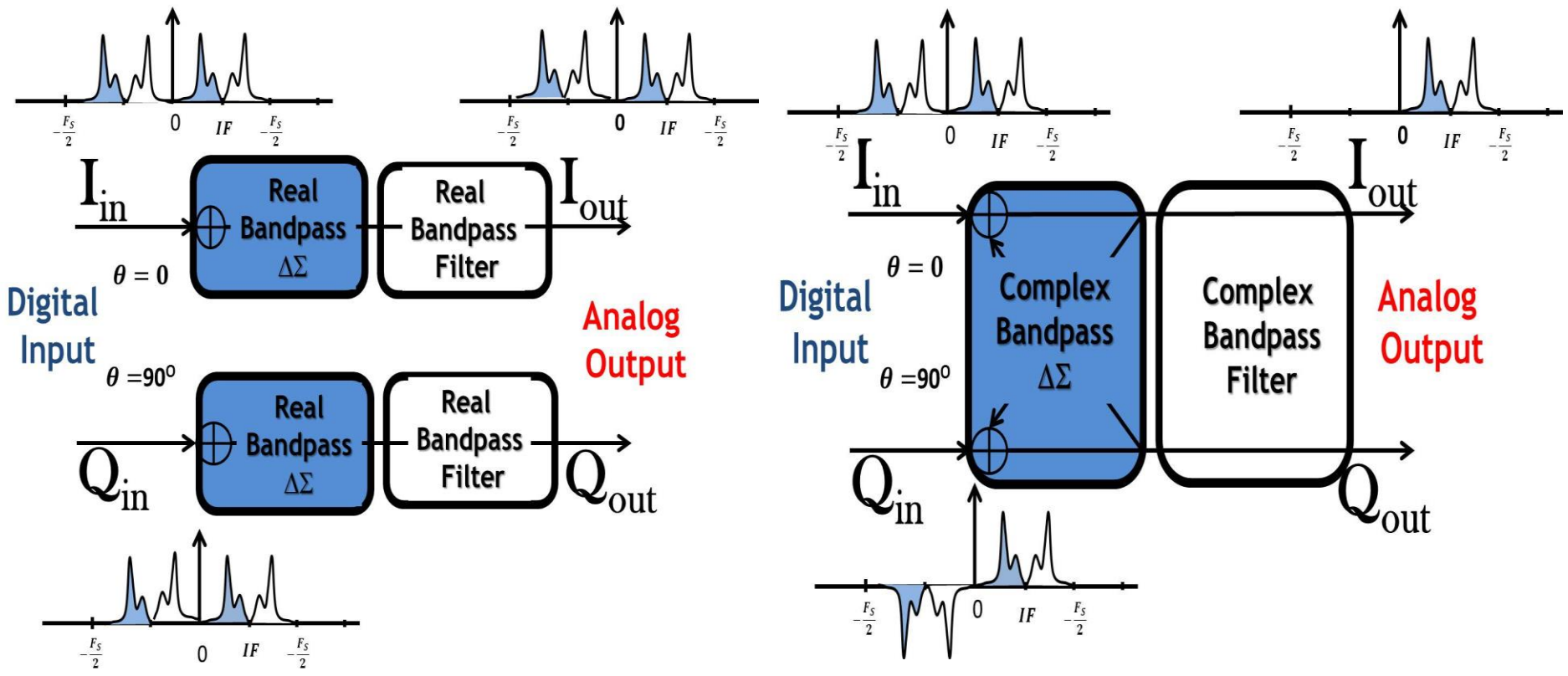


Complex number

$$- I_{in}(n) + jQ_{in}(n) \rightarrow I_{out}(n) + jQ_{out}(n)$$

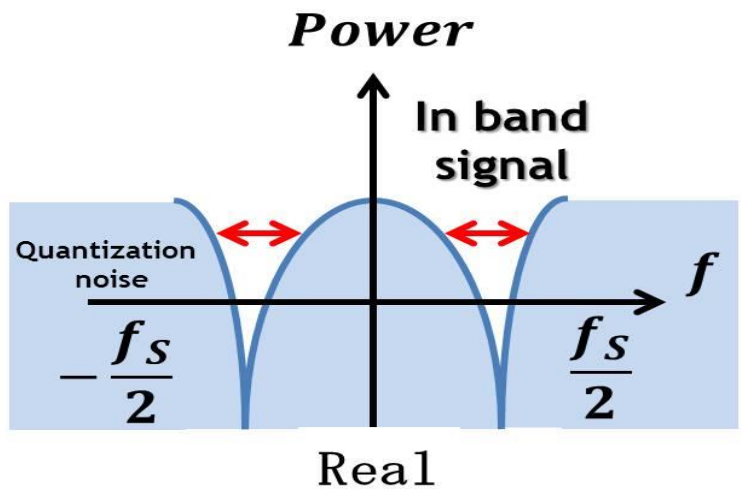
I - In phase (real) , Q - In Quadrature (imaginary)

Filtering

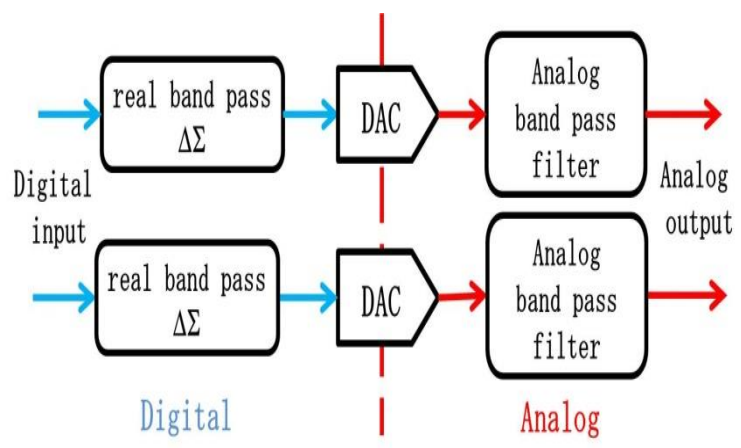
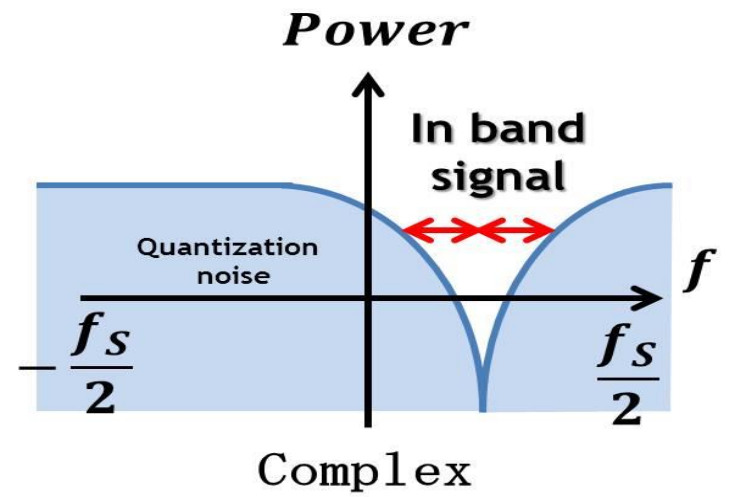


more filtering stage \leftrightarrow **less filtering stage**

Real & Complex

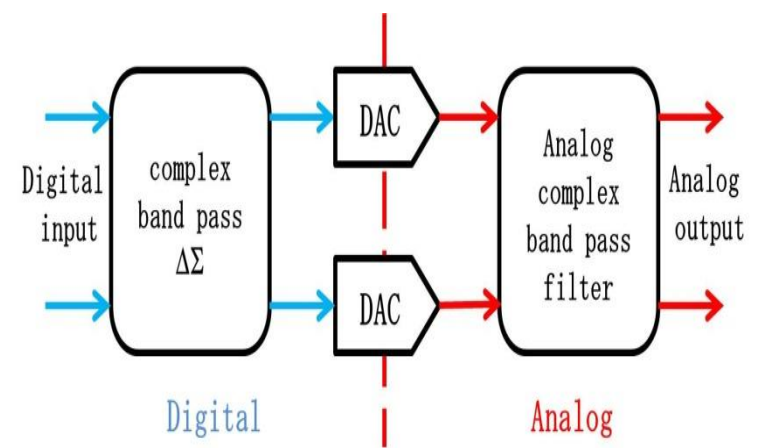


bandwidth

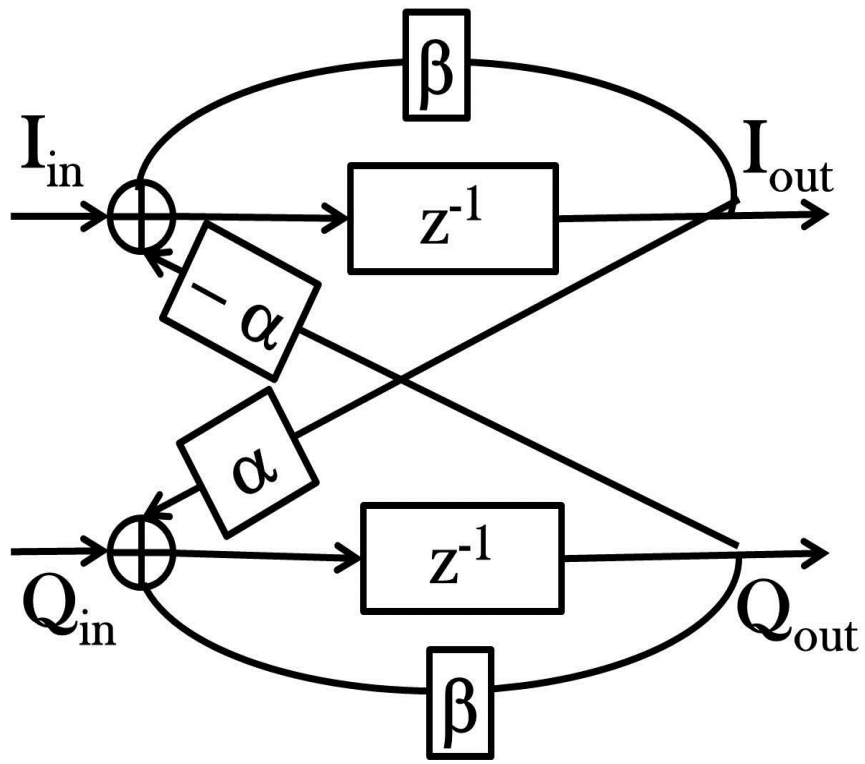


Power consumption

Cost



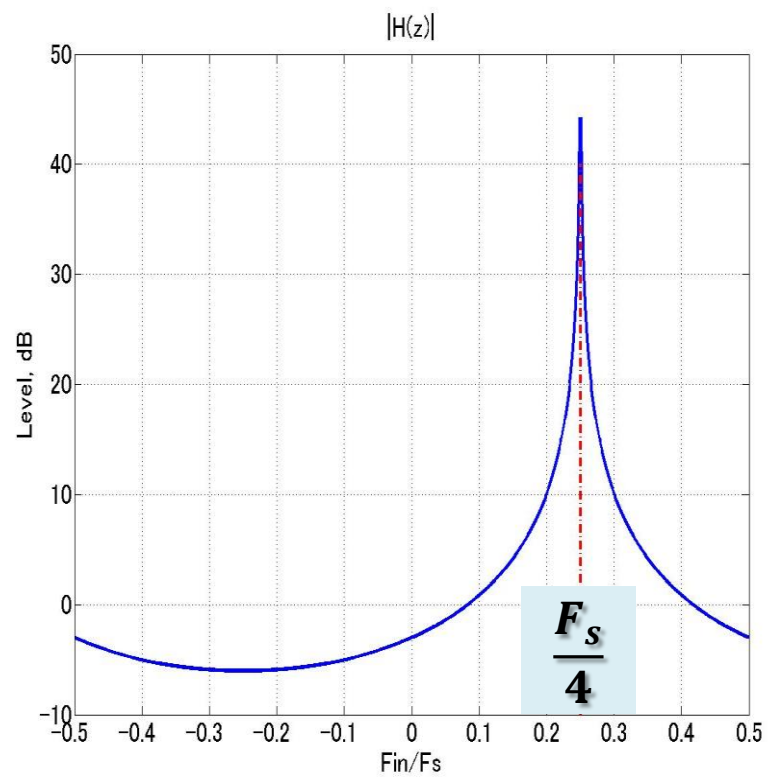
Complex Band pass Filter



$$H(z) = \frac{1}{z - (\beta + j\alpha)}$$

Transfer function

Frequency response



Frequency Response

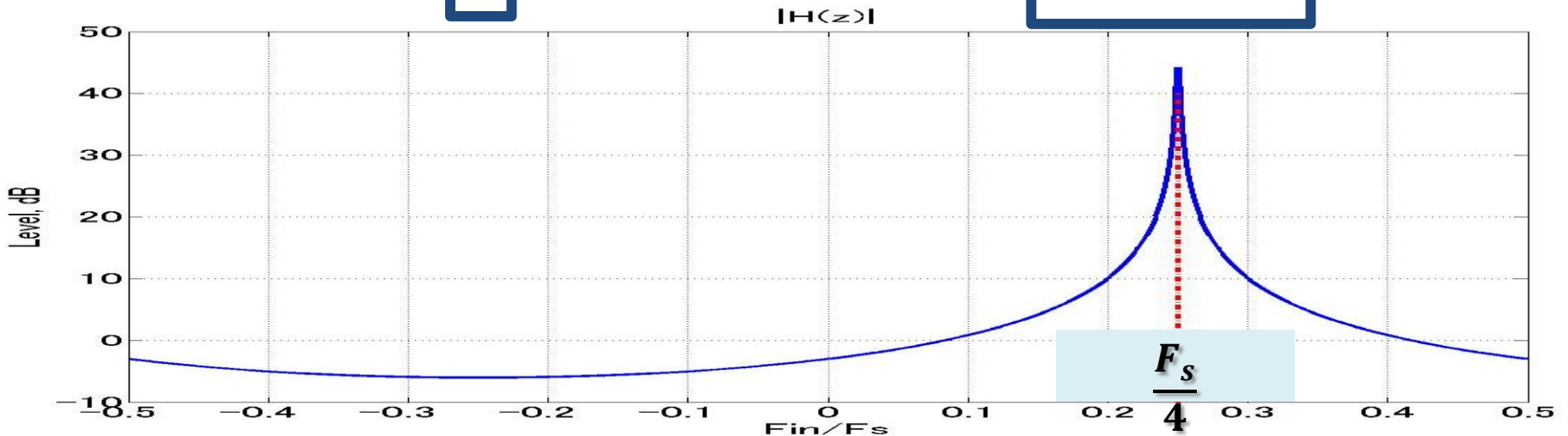
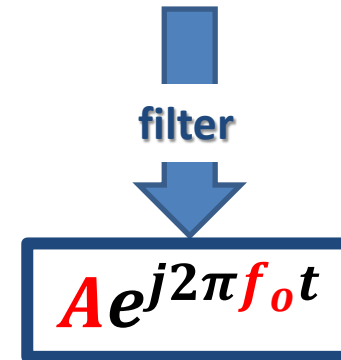
Case: negative

$$\begin{cases} I_{in}(t) = \cos(2\pi f_o t) \\ Q_{in}(t) = -\sin(2\pi f_o t) \end{cases} \rightarrow e^{-j2\pi f_o t}$$

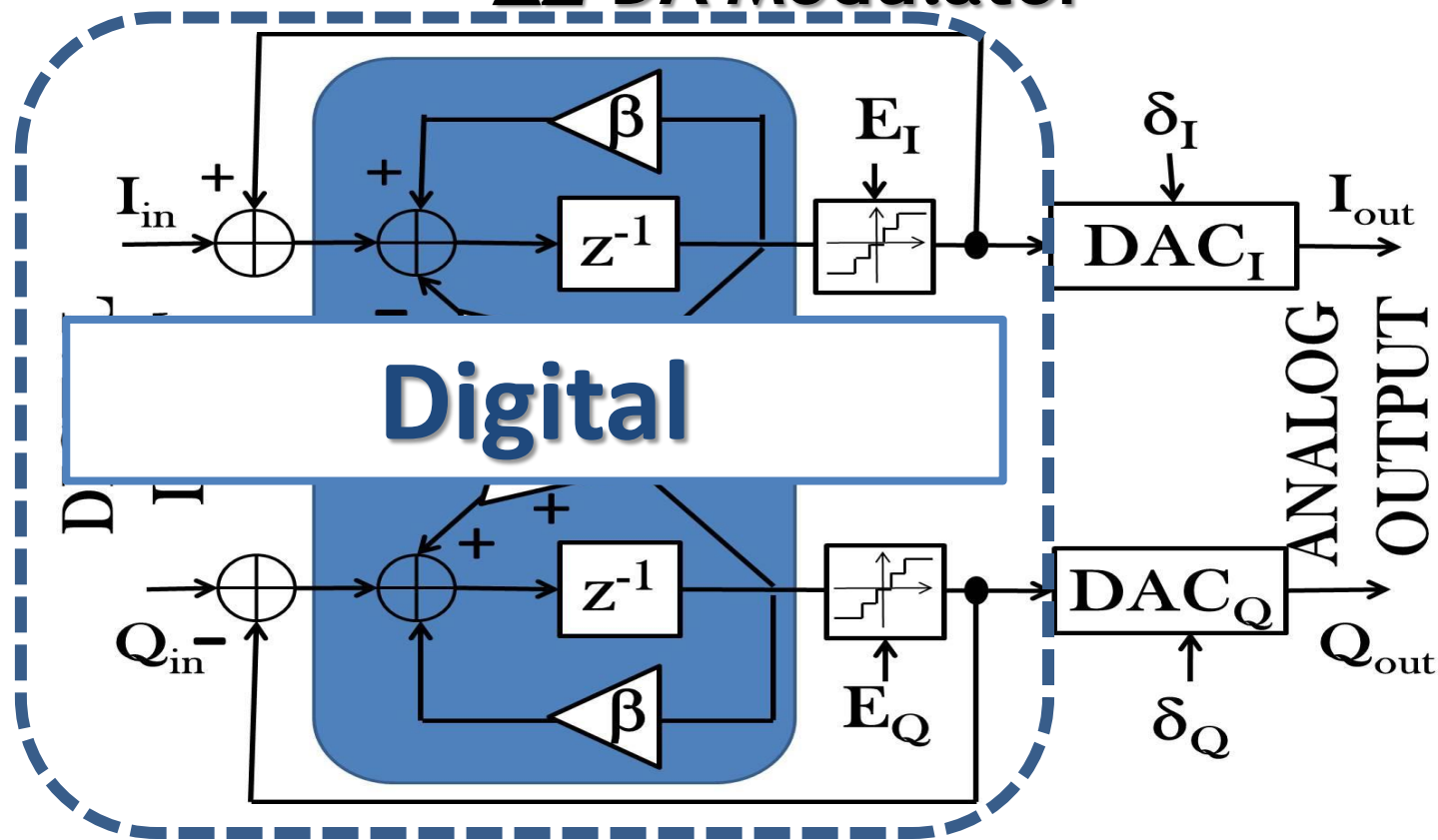


Case: positive

$$\begin{cases} I_{in}(t) = \cos(2\pi f_o t) \\ Q_{in}(t) = \sin(2\pi f_o t) \end{cases} \rightarrow e^{j2\pi f_o t}$$



1storder Complex Band Pass $\Delta\Sigma$ DA Modulator

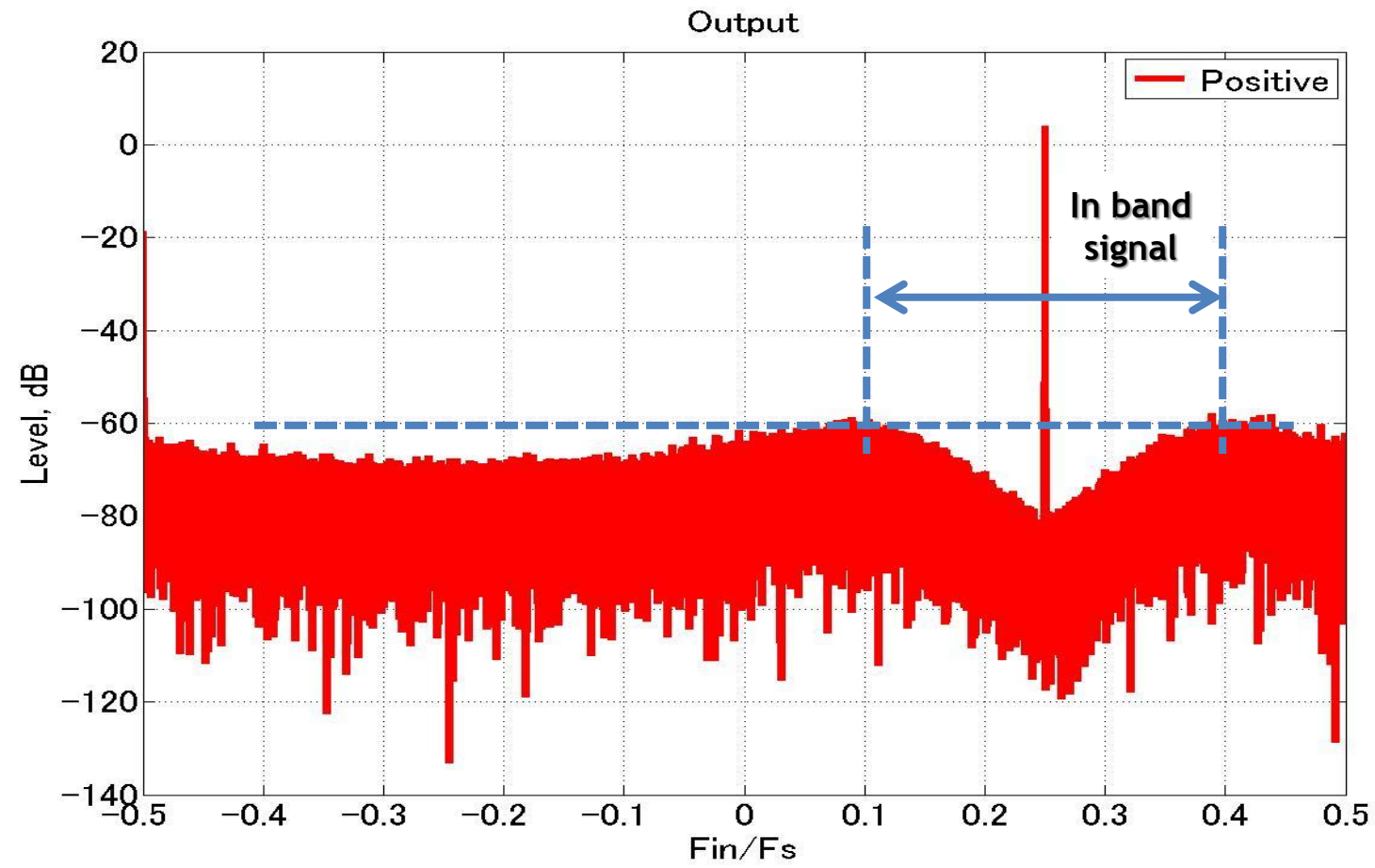


- ① Oversampling
- ② Noise-shaping

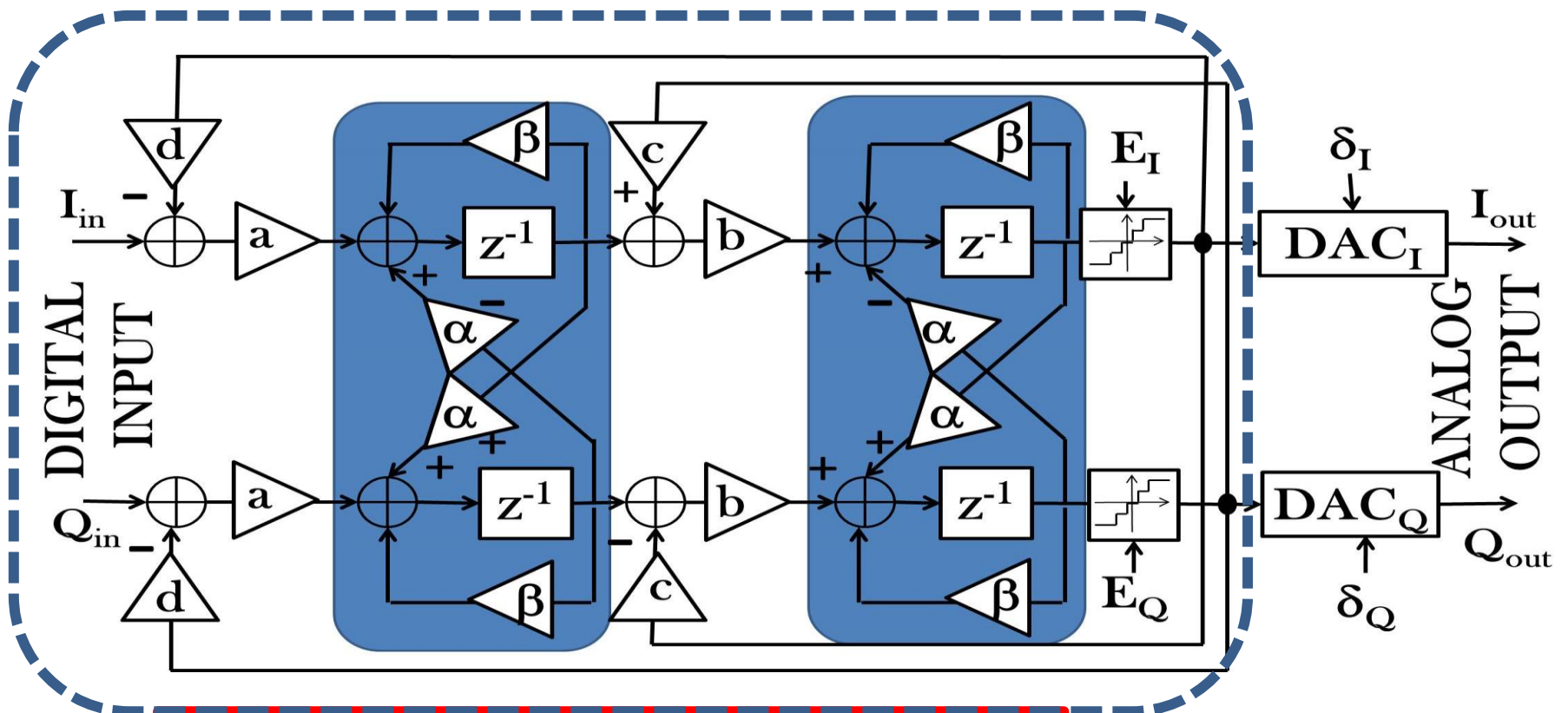
$$Y(z) = \frac{H(z)}{1 + H(z)} X(z) + \frac{1}{1 + H(z)} E(z)$$

$H(z) \rightarrow \infty, STF = 1$ $H(z) \rightarrow \infty, NTF = 0$

1st order Complex Bandpass $\Delta\Sigma$ Output Spectrum

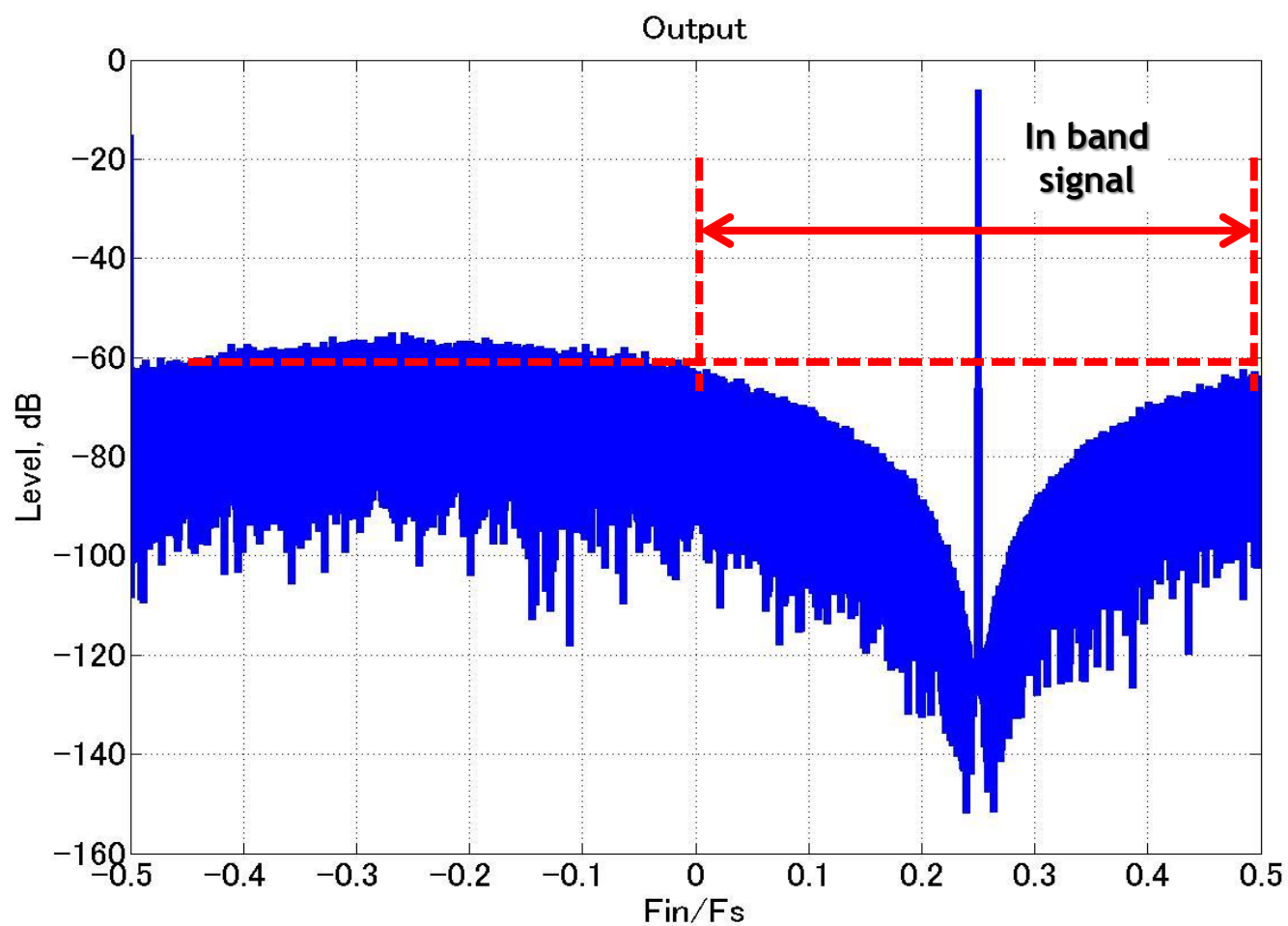


2nd order Complex Band Pass $\Delta\Sigma$ DA Modulator



No I,Q mismatch in modulation

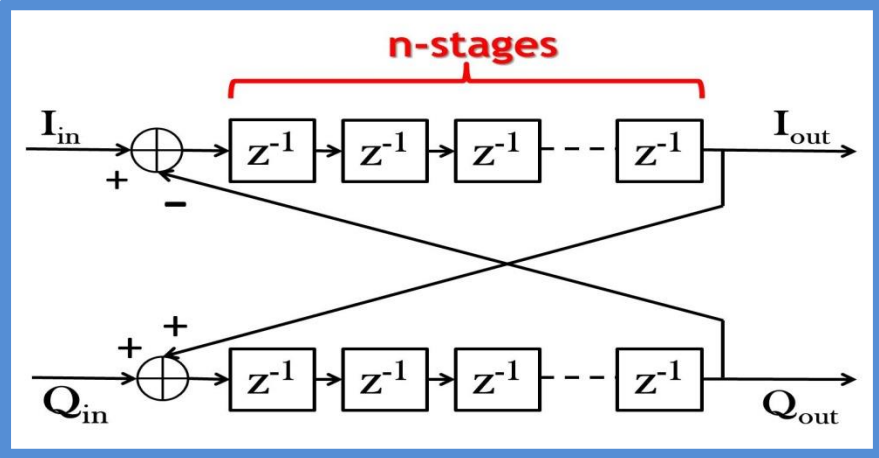
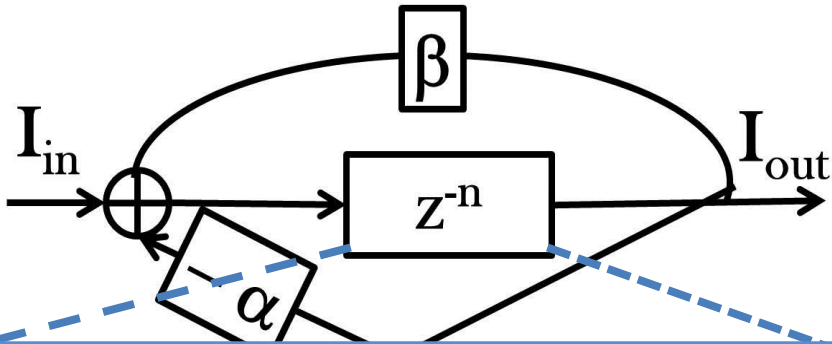
2nd order Complex Bandpass $\Delta\Sigma$ Output Spectrum



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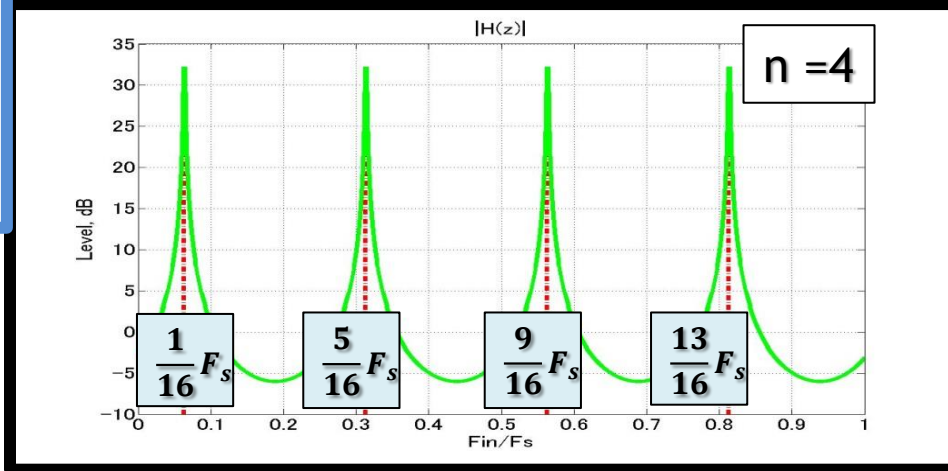
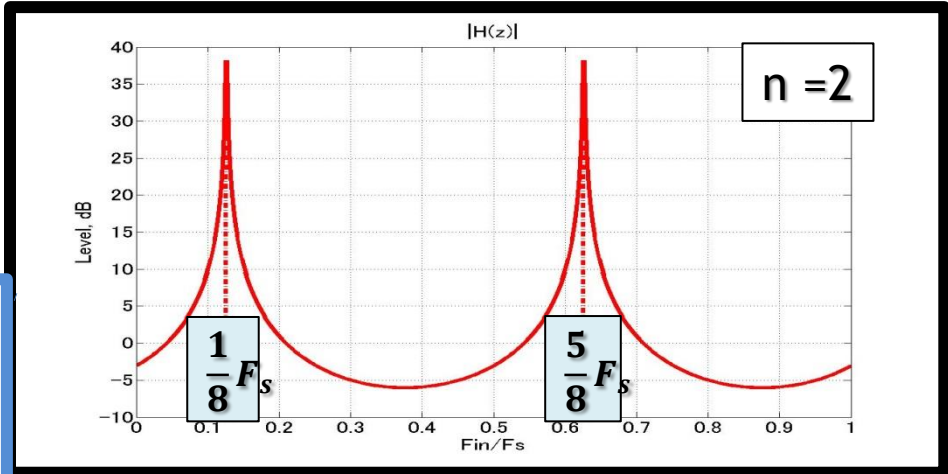
Complex Multi-Band pass Filter



$$H(z) = \frac{1}{z^n - (\beta + j\alpha)}$$

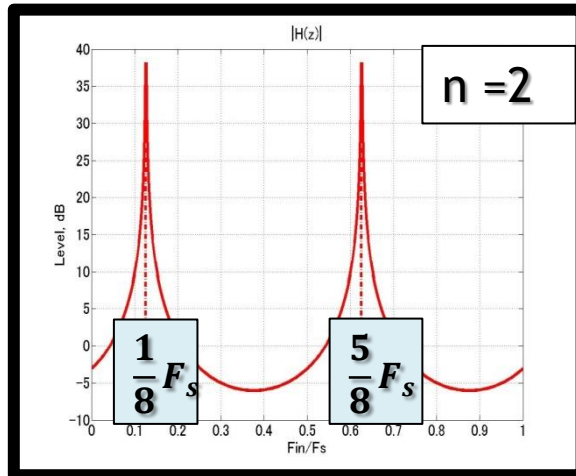
Transfer function

Frequency response



Frequency Response

Frequency response

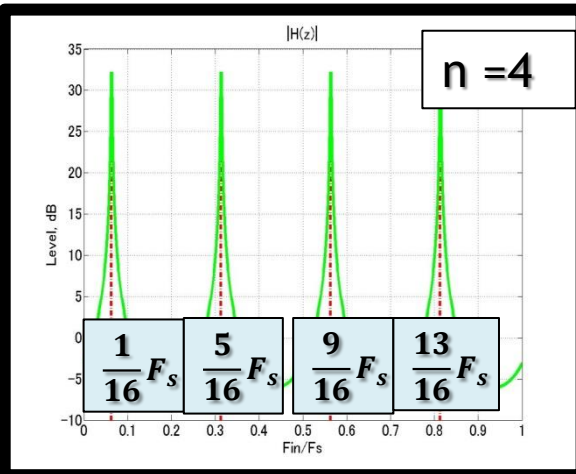


Case: $n=2$

$$\begin{cases} I_{in}(t) = \cos(2\pi f_0 t) + \cos(2\pi f_1 t) \\ Q_{in}(t) = \sin(2\pi f_0 t) + \sin(2\pi f_1 t) \end{cases}$$

filter

$$A_0 e^{j2\pi f_0 t} + A_1 e^{j2\pi f_1 t}$$



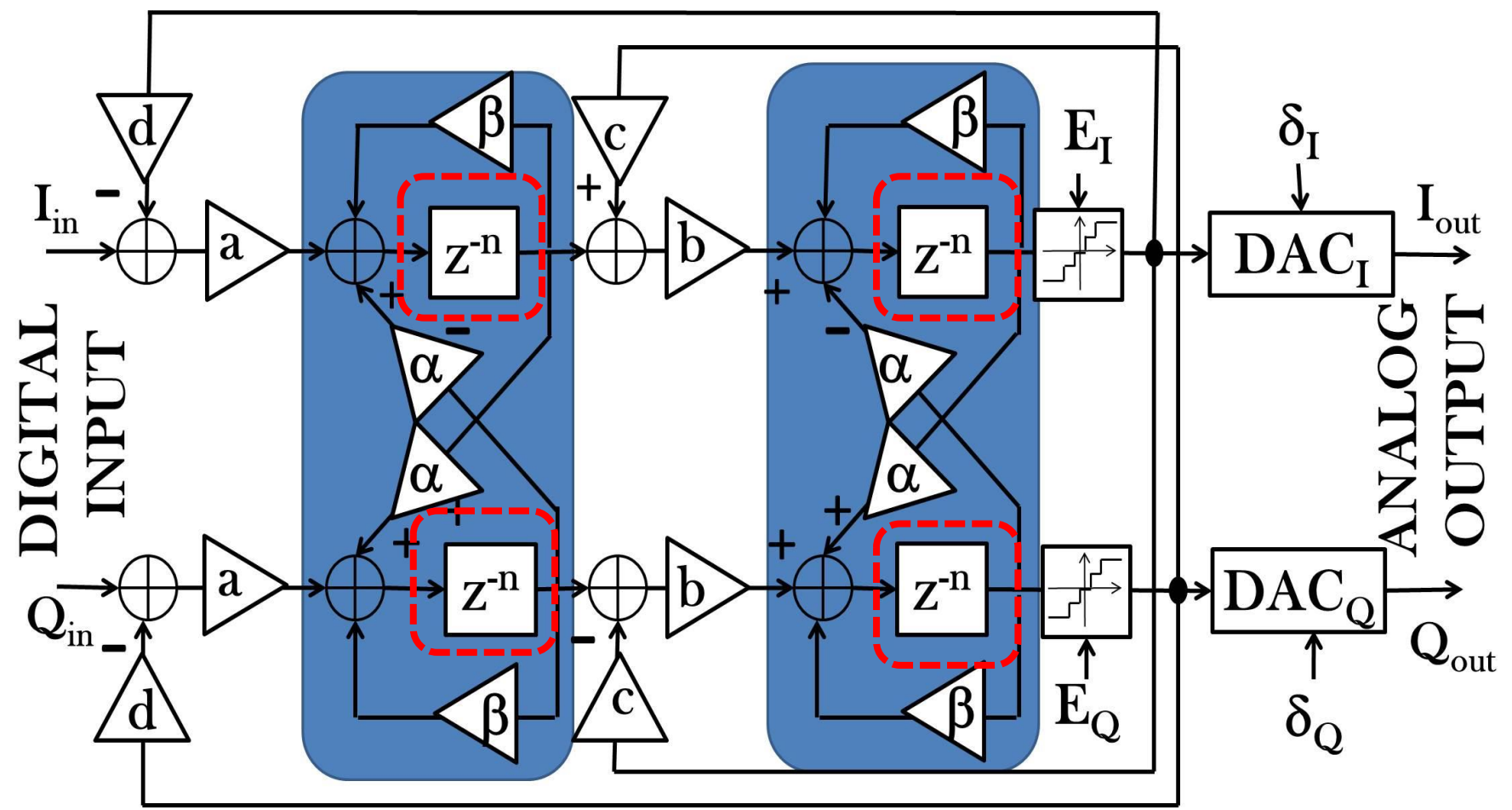
n -stages

$$\begin{cases} I_{in}(t) = \cos(2\pi f_0 t) + \cos(2\pi f_1 t) + \cos(2\pi f_2 t) + \dots \cos(2\pi f_n t) \\ Q_{in}(t) = \sin(2\pi f_0 t) + \sin(2\pi f_1 t) + \sin(2\pi f_2 t) + \dots \sin(2\pi f_n t) \end{cases}$$

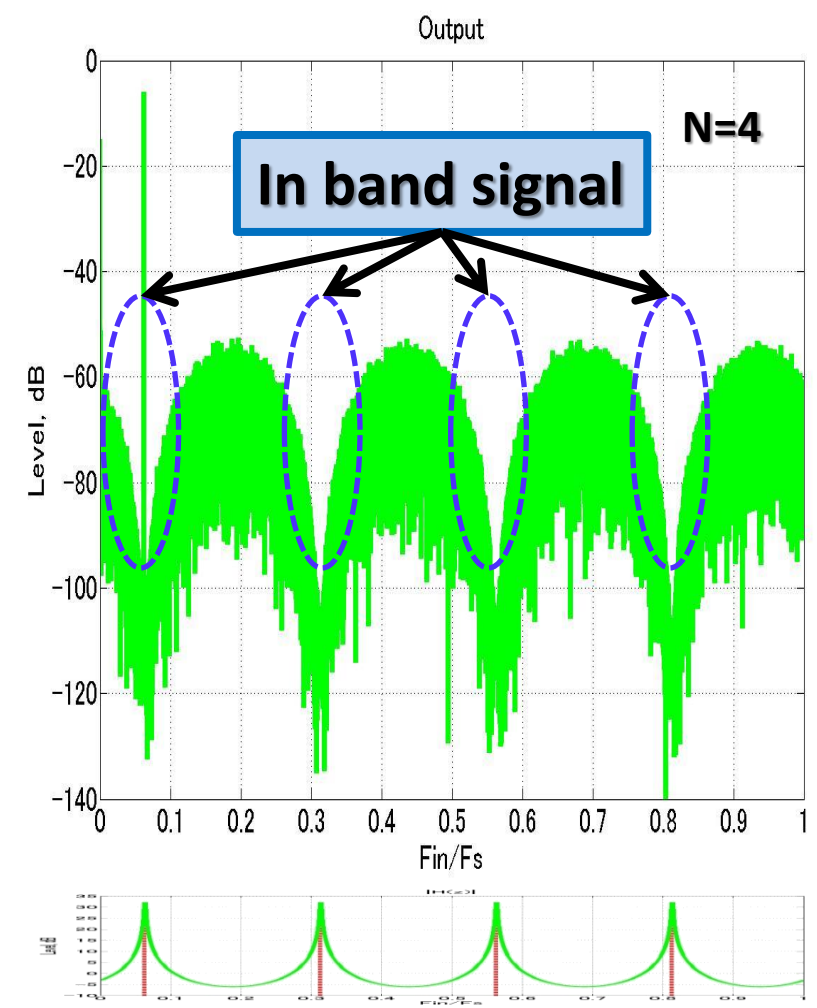
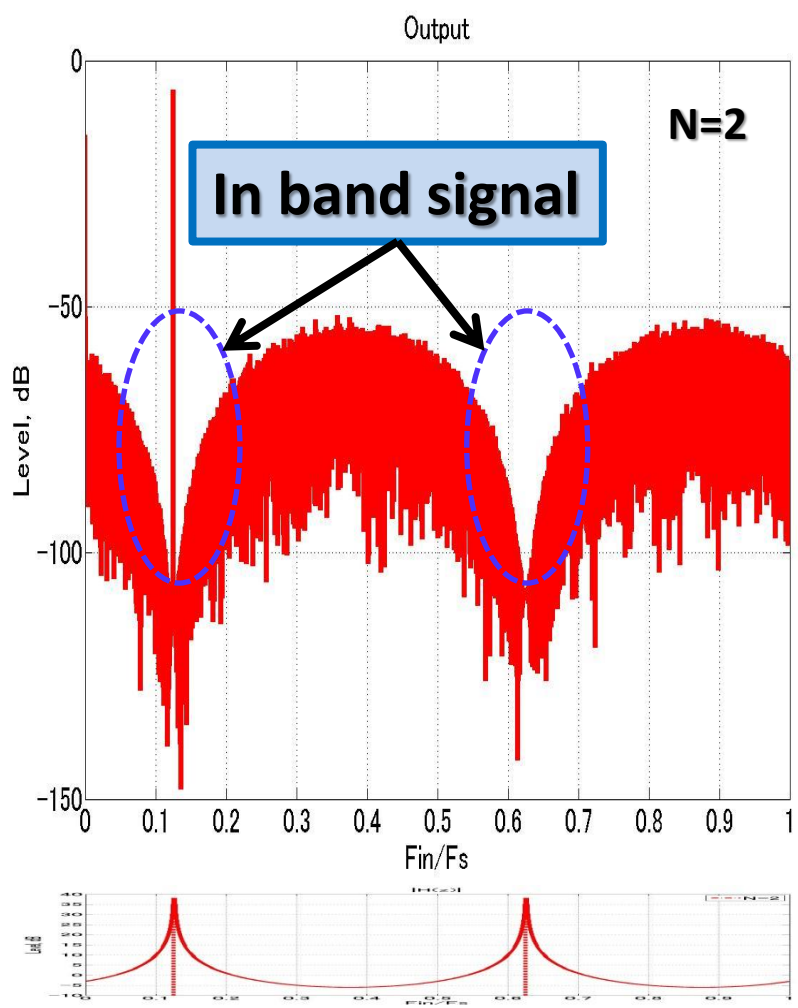
filter

$$A_0 e^{j2\pi f_0 t} + A_1 e^{j2\pi f_1 t} + A_2 e^{j2\pi f_2 t} + \dots + A_n e^{j2\pi f_n t}$$

2ndorder Complex Multi-Band Pass $\Delta\Sigma$ DA Modulator



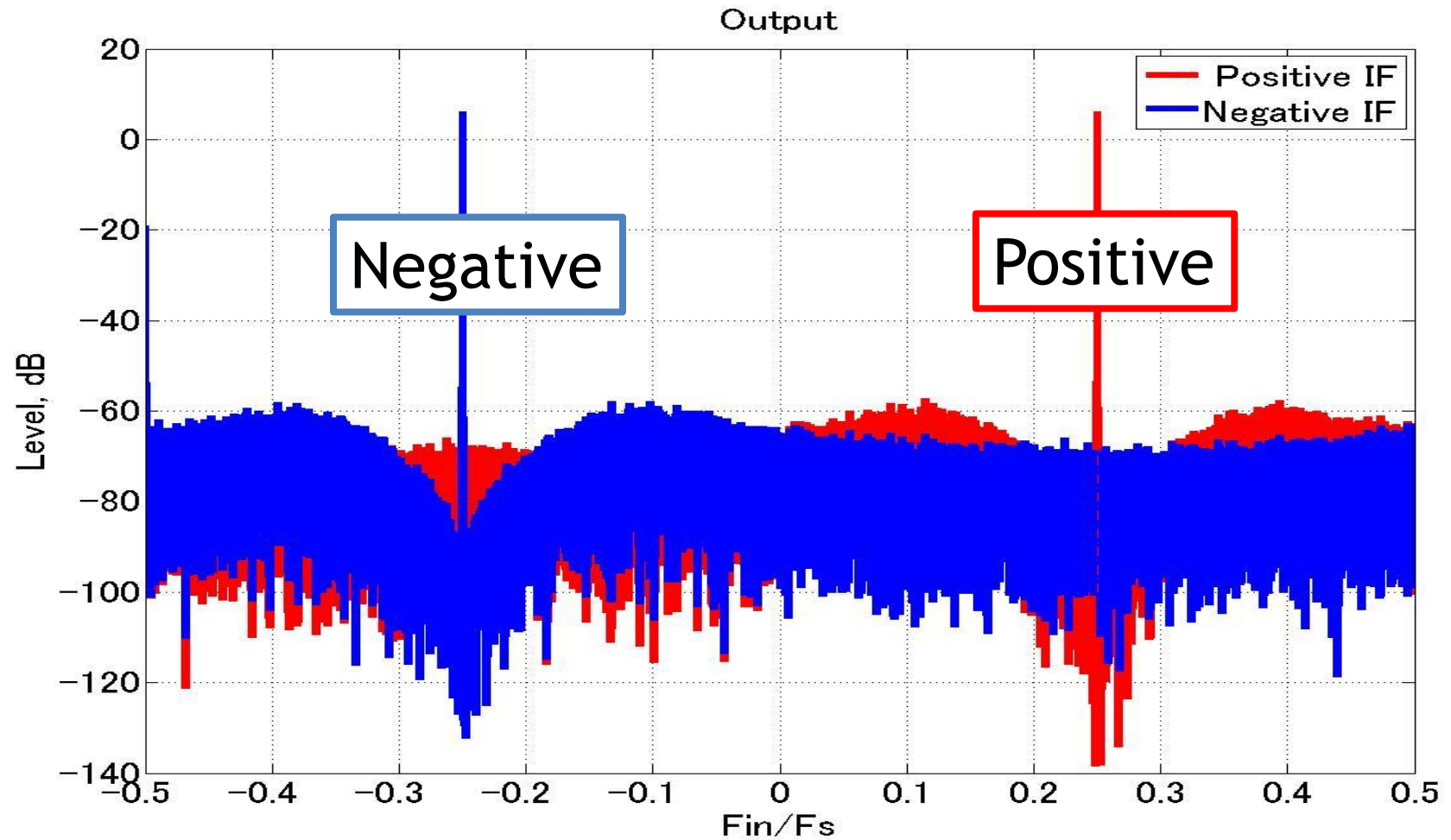
Complex Multi-Bandpass(1)



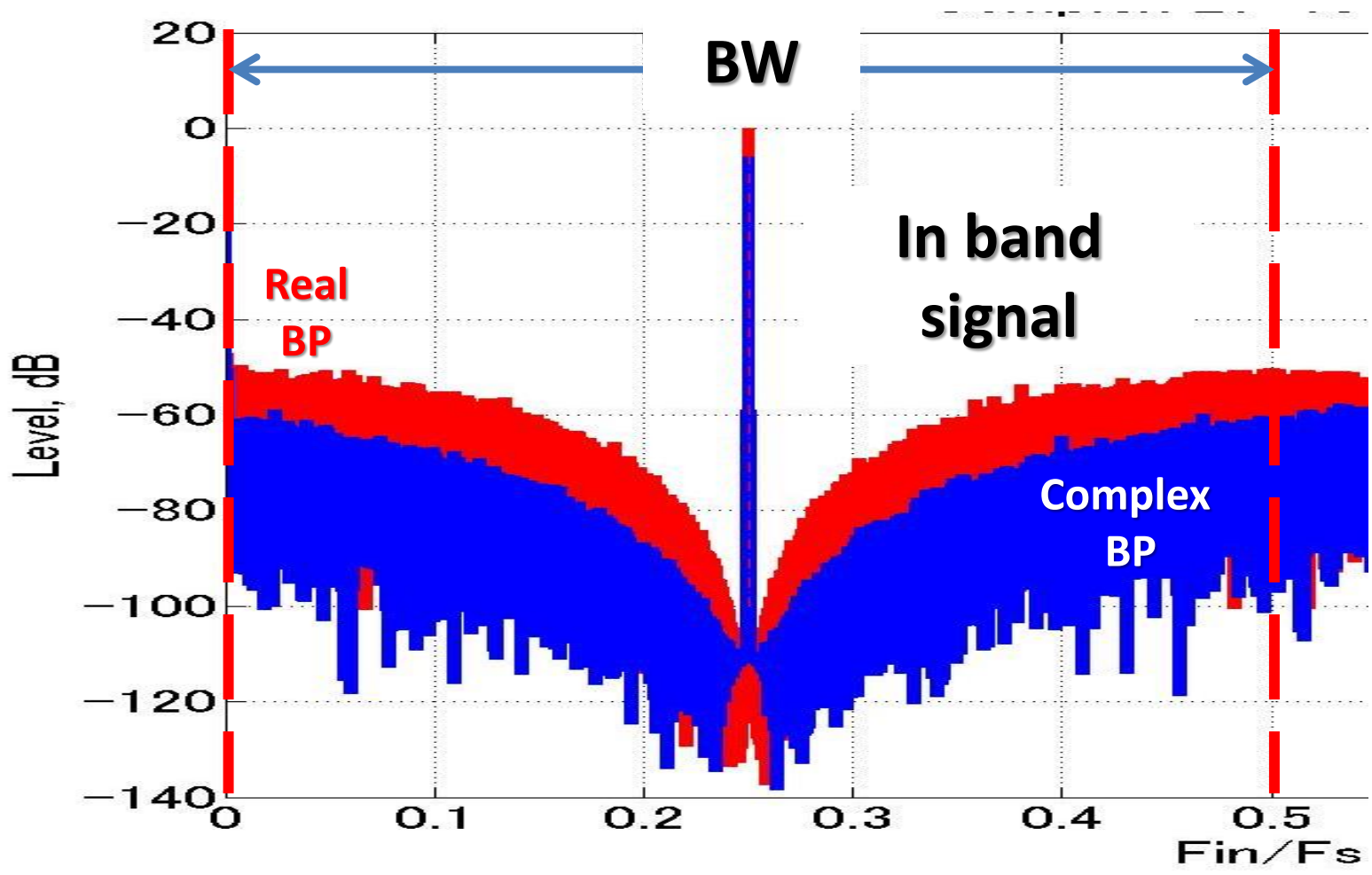
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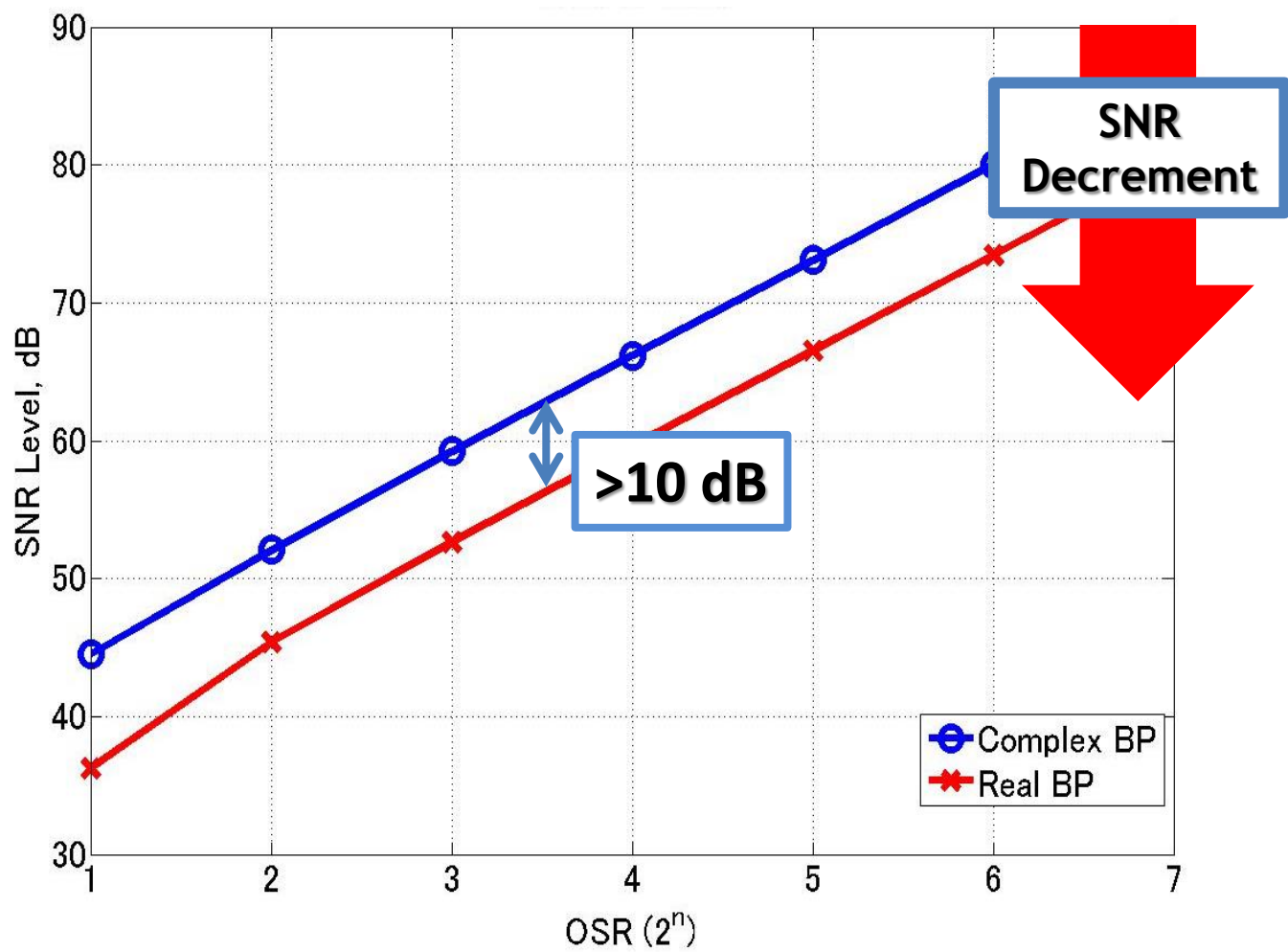
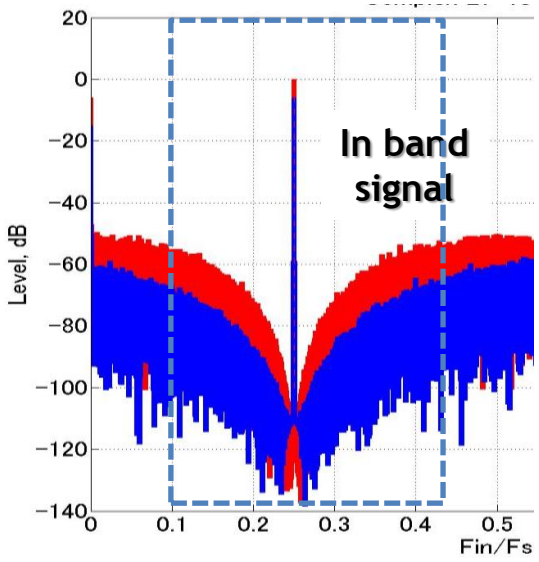
Positive & Negative



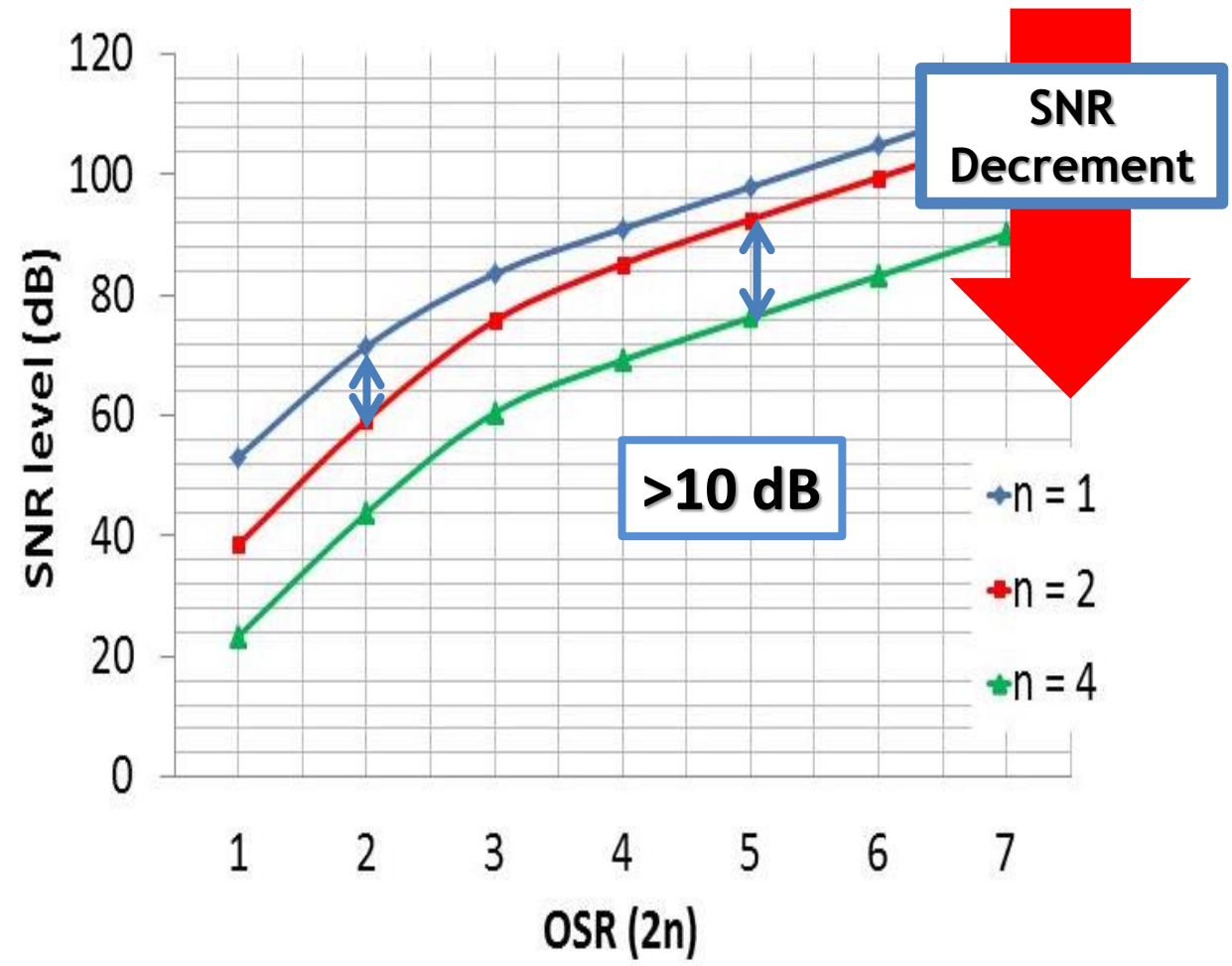
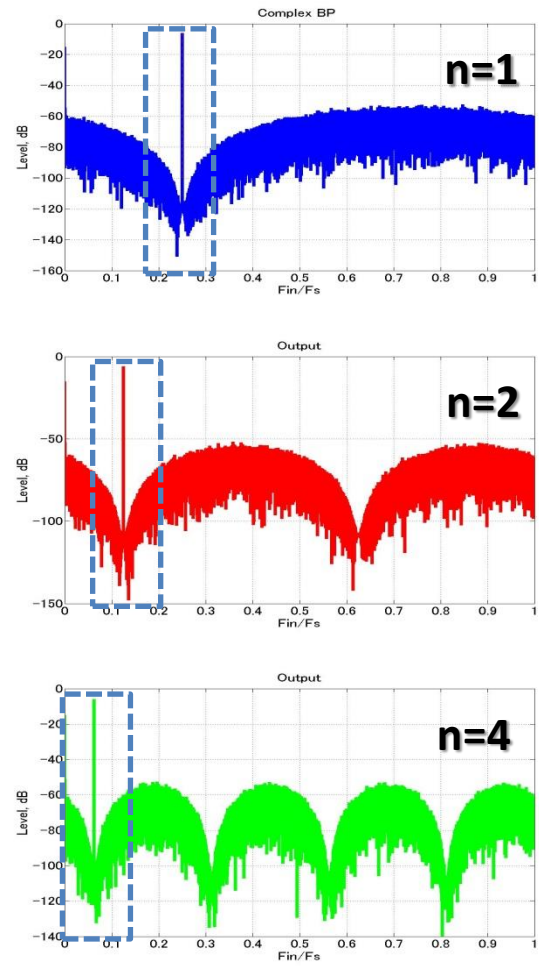
Real versus Complex



Real versus Complex(2)



Complex Multi-Bandpass (2)



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Conclusion

- **Real & complex $\Delta\Sigma$ modulator**
 - Digital rich
 - SNR \rightarrow Complex BP $\Delta\Sigma$ > Real BP $\Delta\Sigma$

Suitable for a high quality, low cost for I,Q signals generation

- **Complex multi-bandpass $\Delta\Sigma$ modulator**
 - lower SNR compare to single tone

Suitable for a high quality, low cost multi-tone I,Q signals generation

Thank you very much