

Low-Distortion Signal Generation for ADC Testing

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STARC



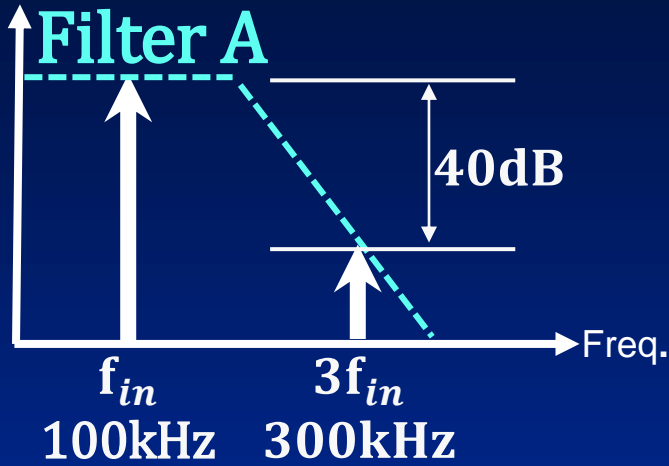
Research Objectives

- Use an Arbitrary Waveform Generator (AWG) as low-cost low-distortion signal source for ADC testing
- Validate our 3rd-harmonic-cancelling phase-switching signal generation technique
- Perform analysis, simulation, and experiments using AWGs for ADC testing

Expected Advantage of Our Solution

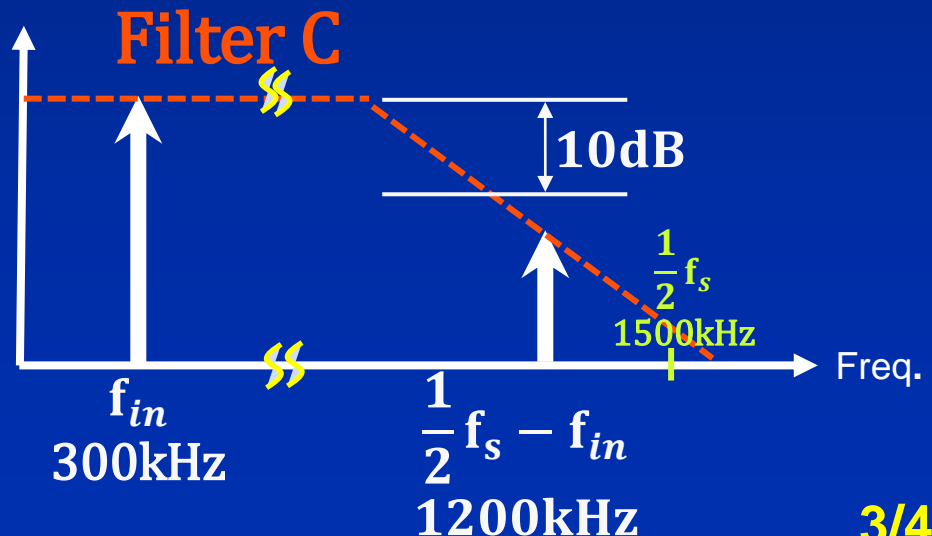
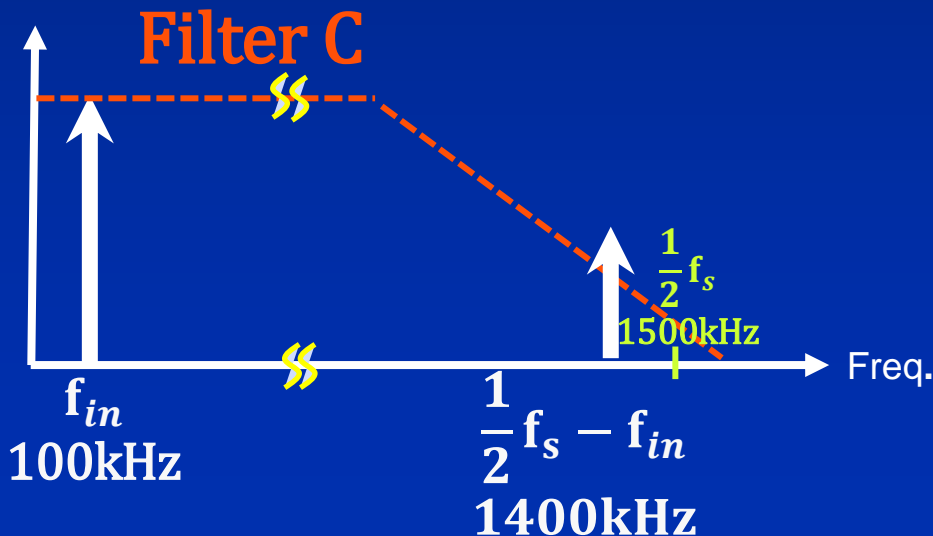
Conventional Method

Tough LPF requirements



Our Method

Relaxed LPF requirements



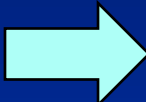
Outline

- Background to this research
- Proposed solution
- Problems with proposed solution
- Remedy 1
- Remedy 2
- Experimental results
- Conclusions

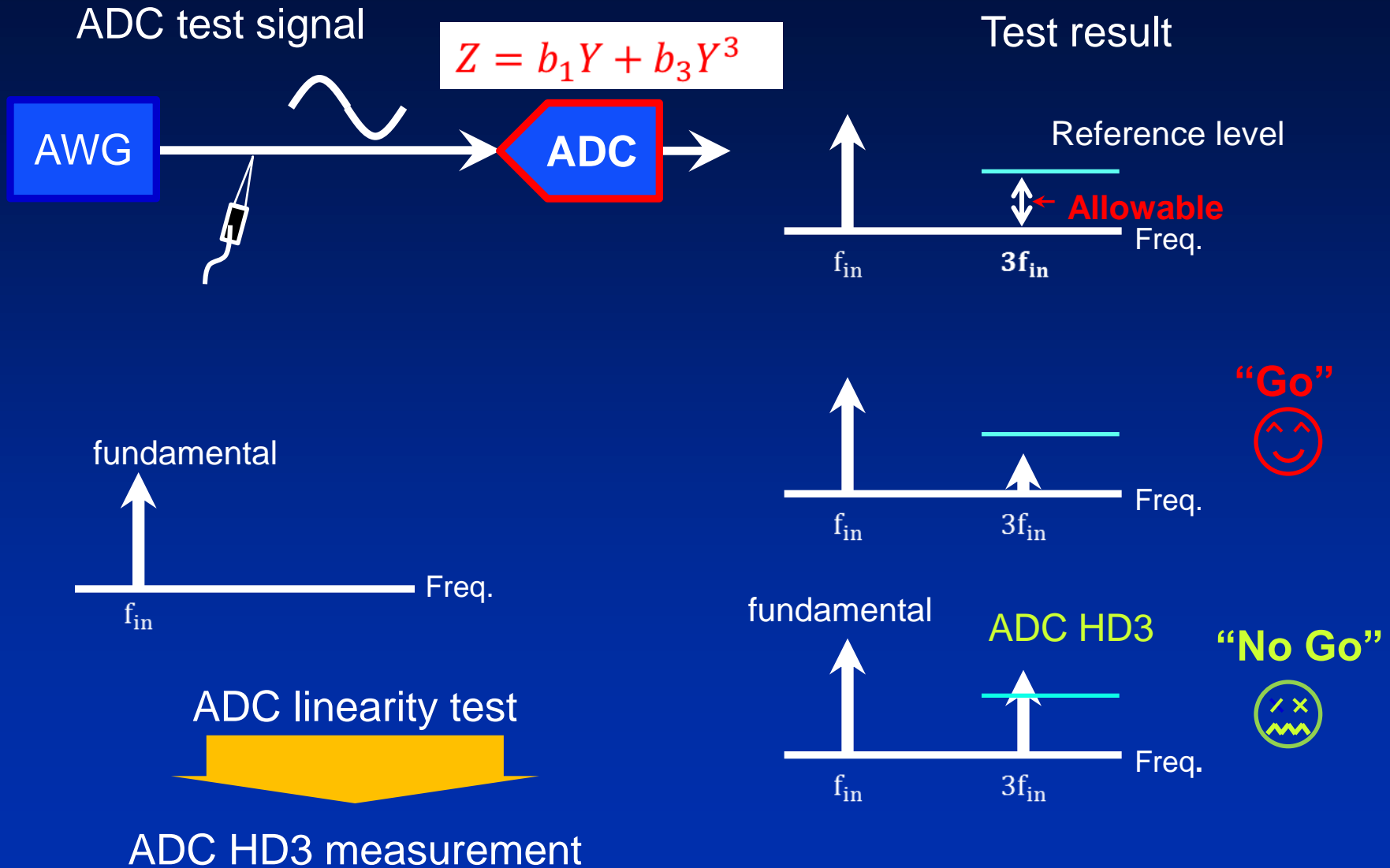
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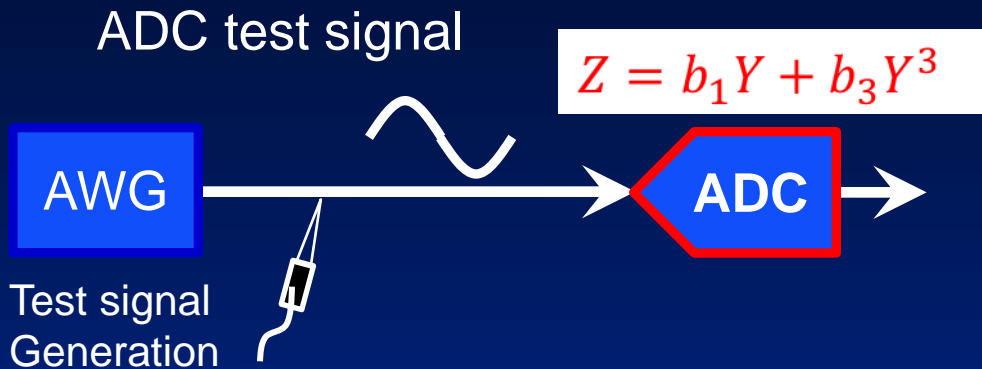
Background to this Research

- ADCs are important part of mixed-signal SOCs.
- Need a low-cost, low-distortion signal source for ADC linearity testing.
- #1: Use existing AWGs in testers  low cost
- # 2: Develop a technique that doesn't require identification of AWG nonlinearity, just a DSP program change.

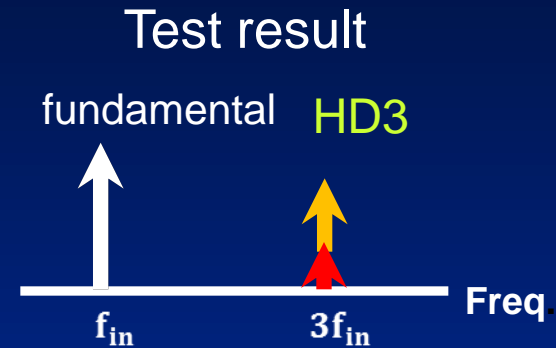
ADC Testing with Sine Signal



Problem with Conventional Method

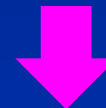


$$Y = a_1 D_{in} + a_3 D_{in}^3$$

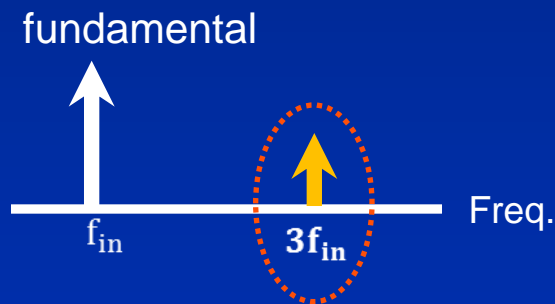


AWG HD3+ADC HD3

ADC HD3 cannot be measured accurately.

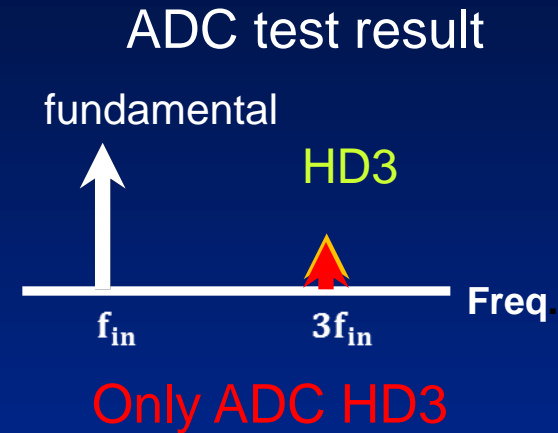
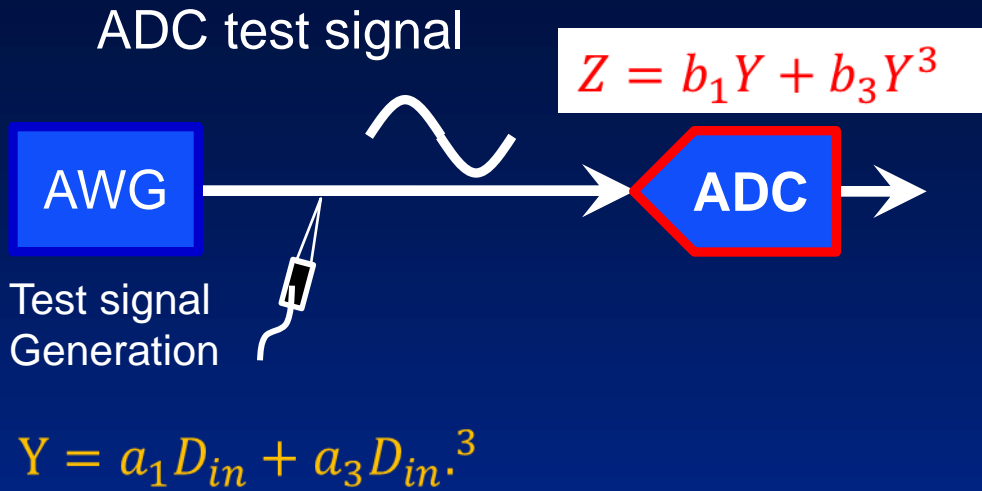


Low quality test

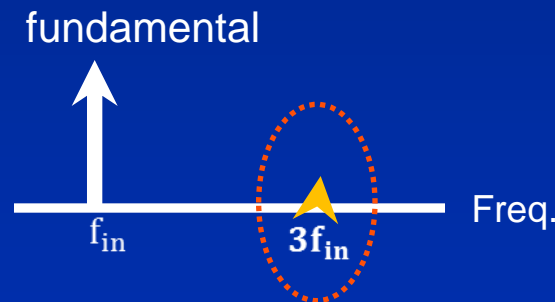


AWG HD3

Research Objective



only
program
change



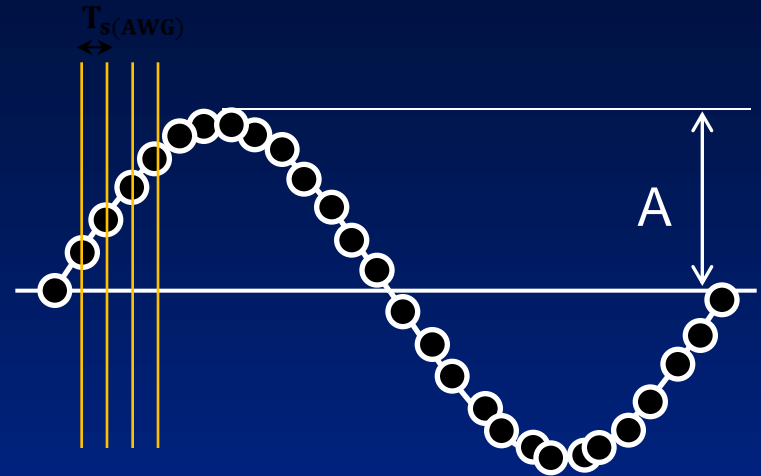
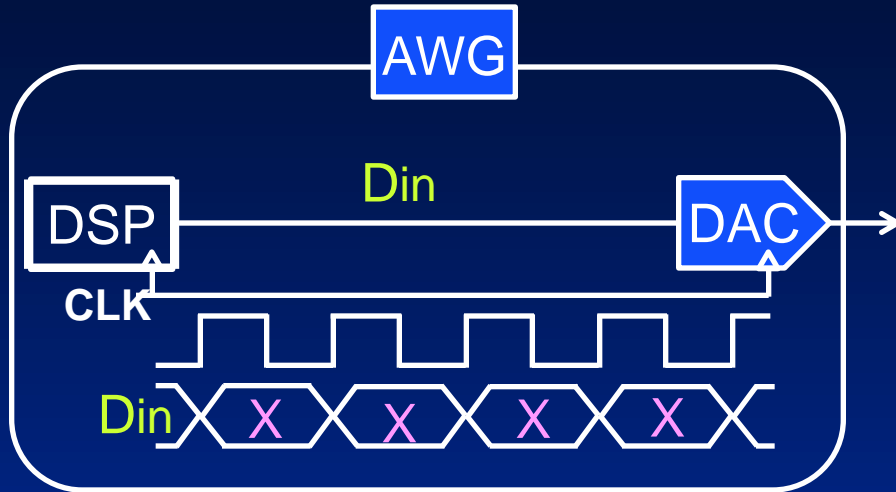
AWG HD3 reduction

High quality test

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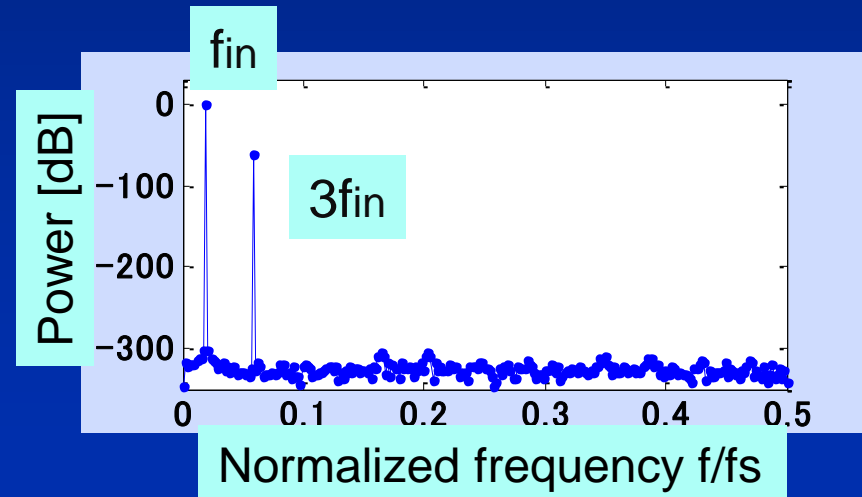
Conventional Signal Generation with AWG



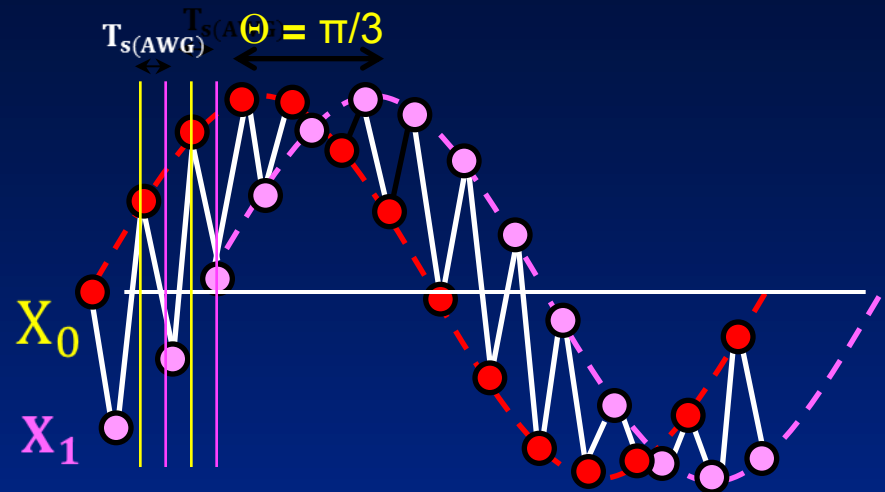
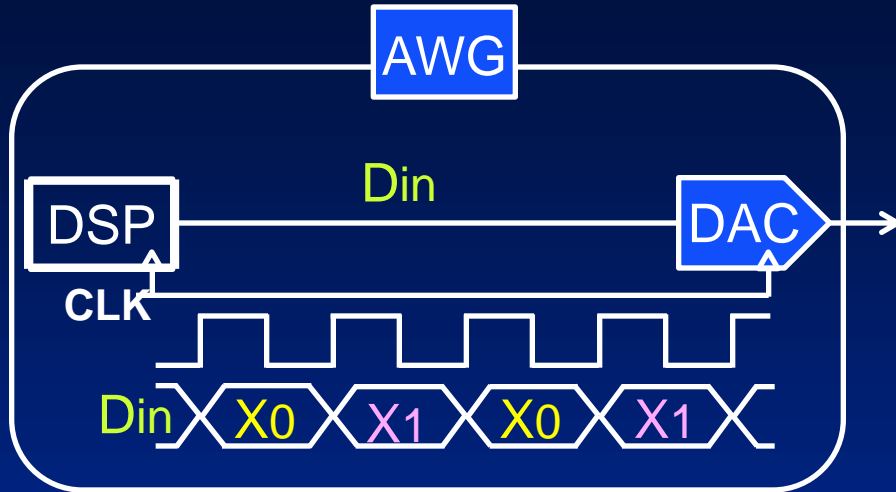
AWG sampling frequency: $f_{s(AWG)} = 1/T_s$

$$X = A \cos(2\pi f_{in} n T_s)$$

DAC has 3rd order nonlinearity



Proposed Signal Generation with AWG

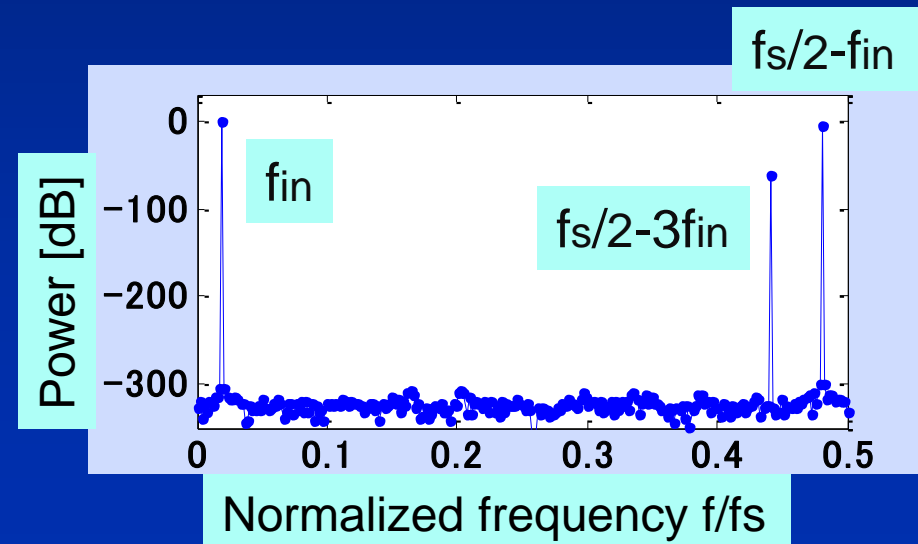


AWG sampling frequency : $f_{s(AWG)} = 1/T_{s(AWG)}$

$$X_0 = 1.15A \cos(2\pi f_{in} n T_s - \pi/6)$$

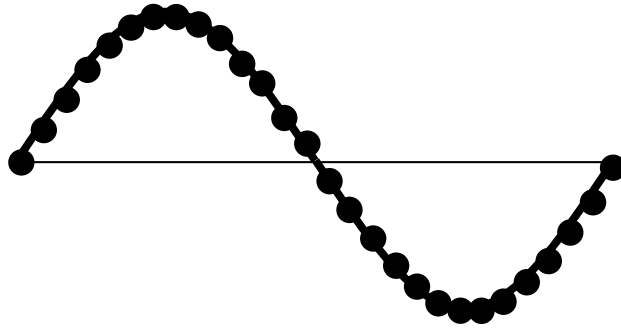
$$X_1 = 1.15A \cos(2\pi f_{in} n T_s + \pi/6)$$

HD3 is cancelled



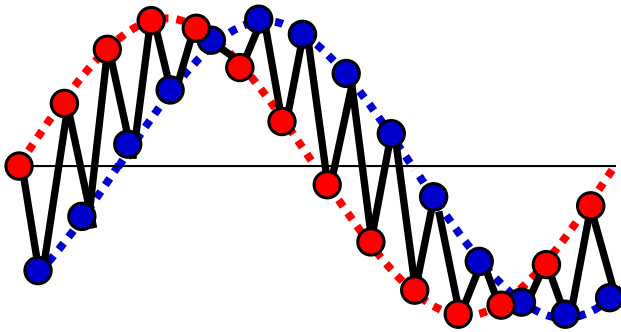
Principle of 3rd Harmonics Cancellation

Conventional



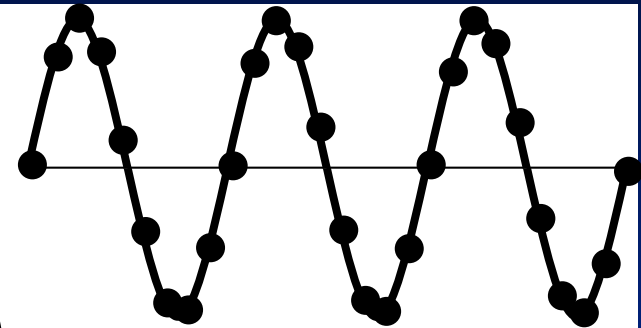
fundamental: f_{in}

Phase Switching

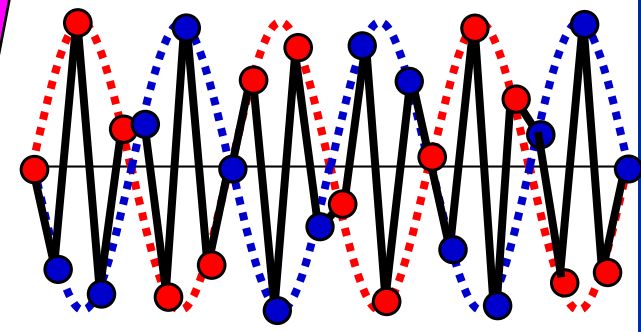


$$\Theta = \pi/3$$

3rd order non-linear system
Phase rotation by x3



3rd harmonics: $3 f_{in}$

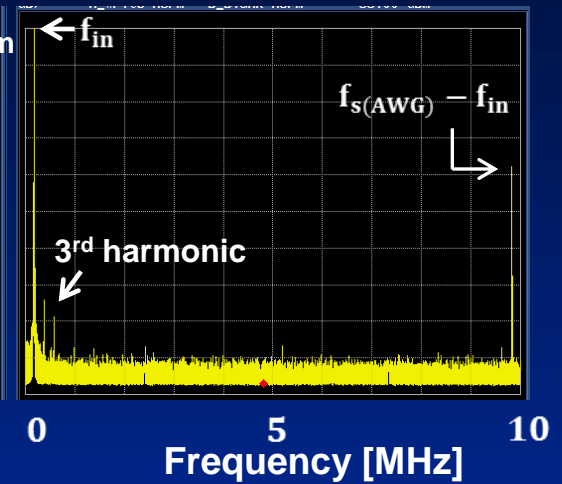
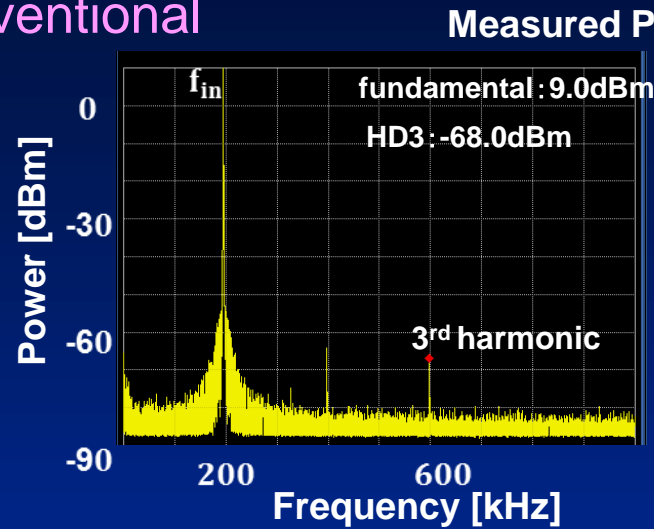
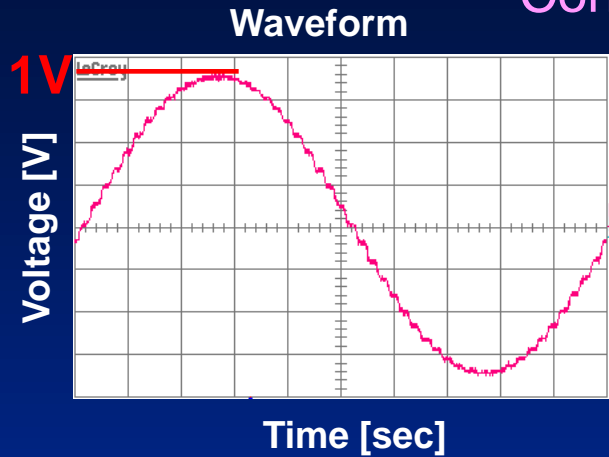


$$3\Theta = \pi$$

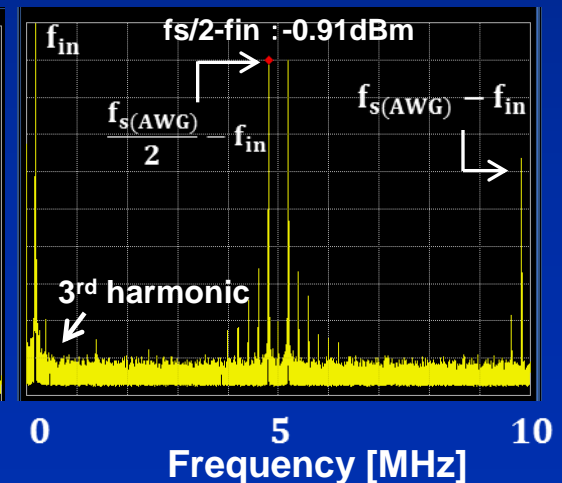
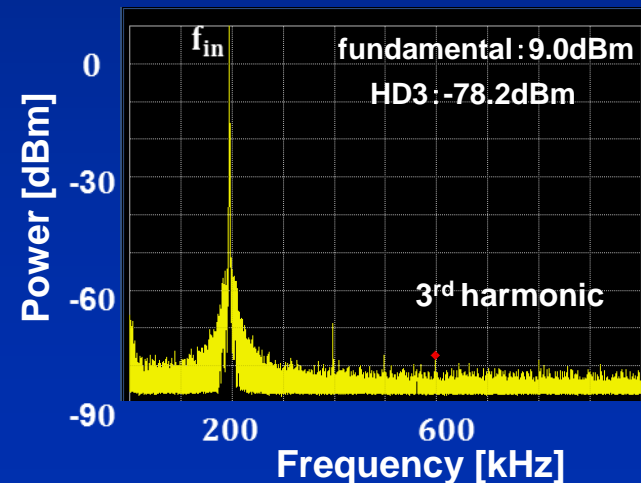
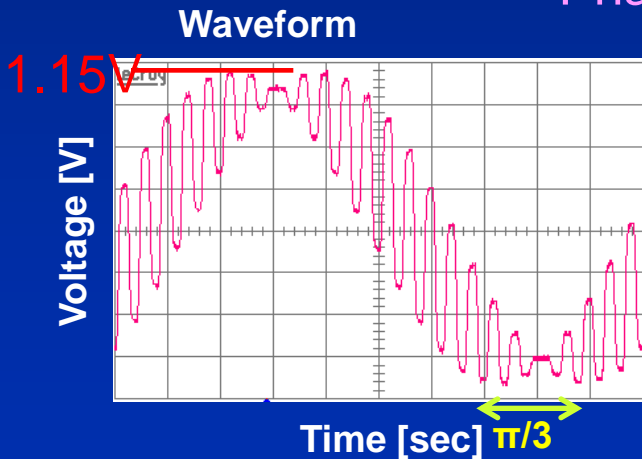
Two waves with phase difference π
are cancelled

Conventional and Phase Switching Signals

Conventional



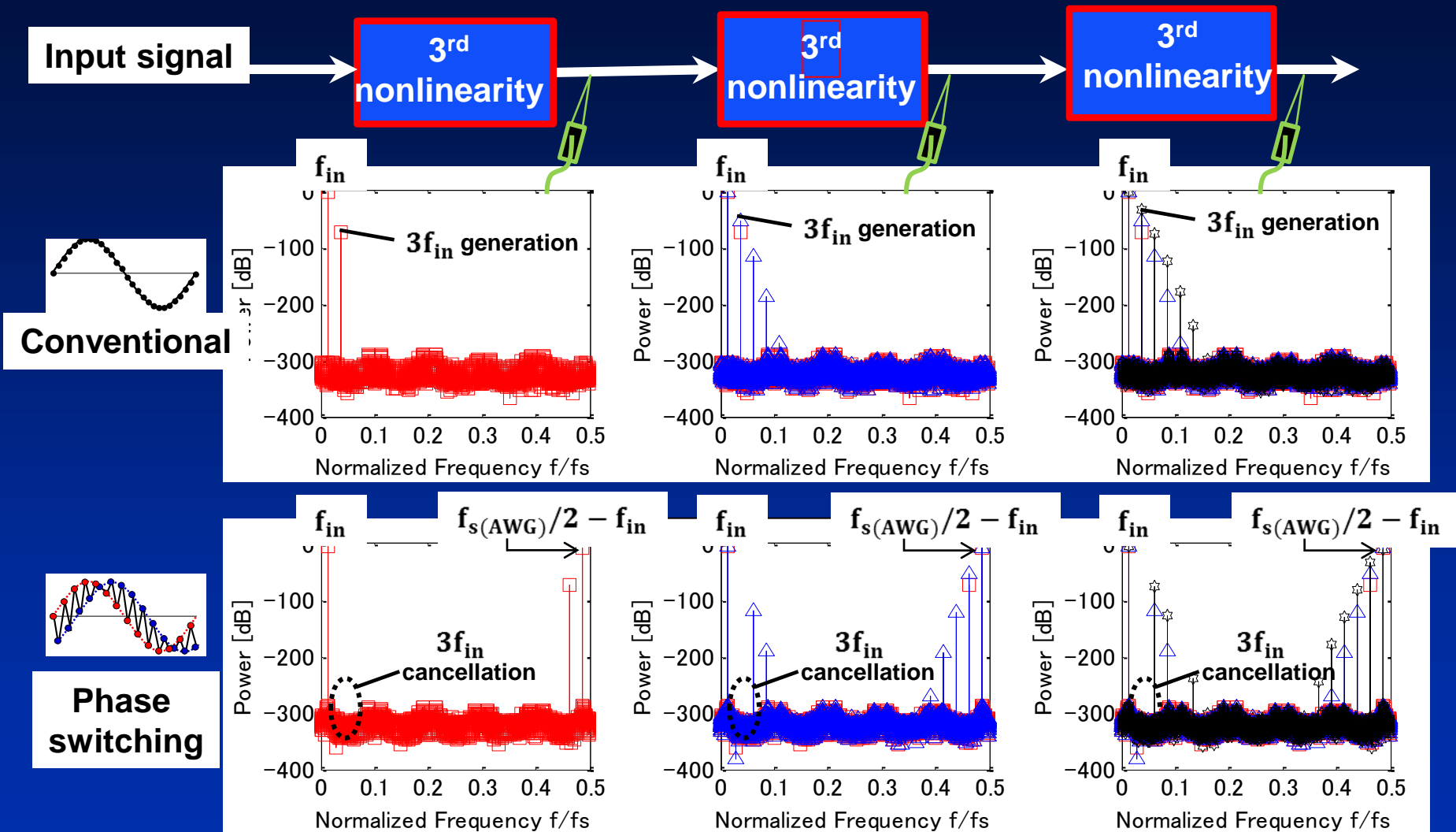
Phase Switching



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Conditions of HD3 Cancellation in Cascaded System



HD3's are not cancelled if attenuate spurious at $f_s/2 - f_{in}$.

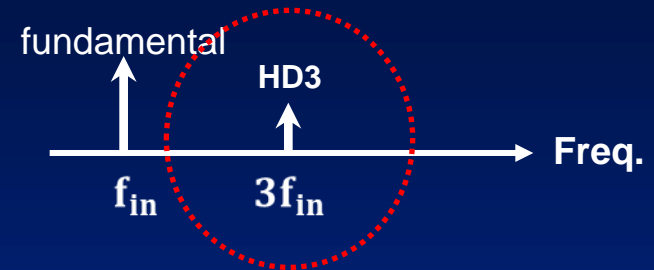
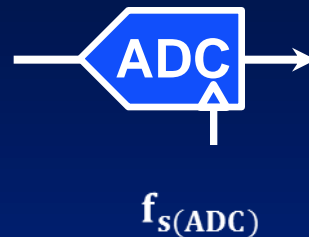
Problem of Phase Switching Signal Generation

Ideal



Nonlinearity Model

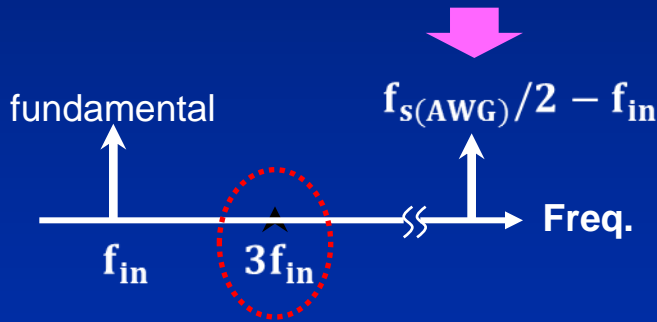
$$Y = a_1X + a_3X^3$$



HD3 due to ADC nonlinearity

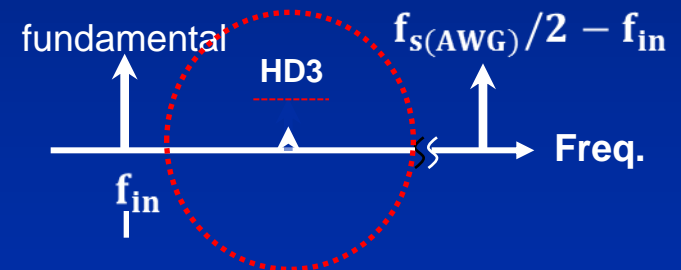
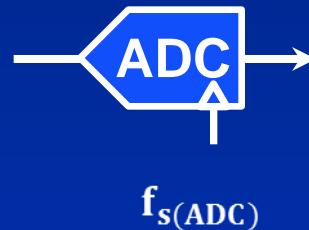
Phase switching

Spurious due to phase switching



Reduction of HD3 due to AWG nonlinearity

$$Y = a_1X + a_3X^3$$



Reduction of HD3 due to ADC nonlinearity

Big problem !

Model for Theoretical Analysis

AWG Input with Phase Switching

$$D_{in}(nT_s) = \begin{cases} A \cdot \sin(2\pi f_{in} nT_s - \frac{\pi}{6}) & n: \text{odd} \\ A \cdot \sin(2\pi f_{in} nT_s + \frac{\pi}{6}) & n: \text{even} \end{cases}$$

AWG Nonlinearity Model

$$Y(nT_s) = a_1 D_{in}(n) + a_3 \{D_{in}(n)\}^3$$

For simplicity
 $f_s(\text{AWG}) = f_s(\text{ADC})$

ADC Nonlinearity Model

$$Z(n) = b_1 Y(nT_s) + b_3 \{Y(nT_s)\}^3$$



AWG Output

$$Y(nT_s) = a_1 D_{in}(n) + a_3 \{D_{in}(n)\}^3$$

$$= P \sin(2\pi f_{in} nT_s) \quad \longleftarrow \text{signal}$$

$$+ \overset{\text{large}}{Q} \cos\left(2\pi\left(\frac{f_s}{2} - f_{in}\right)nT_s\right)$$

$$+ \underset{\text{small}}{R} \cos\left(2\pi\left(\frac{f_s}{2} - 3f_{in}\right)nT_s\right)$$

spurious

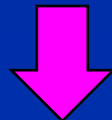
Direct Application of Phase Switching Signal

ADC output $Z(n)$ components

α_0	$\beta_1 \sin(2\pi f_{in} nT_s)$	← signal
$\alpha_{-1} \cos\left(2\pi\left(\frac{f_s}{2} - f_{in}\right)nT_s\right)$	$\beta_3 \sin(2\pi 3f_{in} nT_s)$	← HD3
$\alpha_1 \cos\left(2\pi\left(\frac{f_s}{2} + f_{in}\right)nT_s\right)$	$\beta_{-1} \sin(2\pi(f_s - f_{in})nT_s)$	← HD3
$\alpha_{-3} \cos\left(2\pi\left(\frac{f_s}{2} - 3f_{in}\right)nT_s\right)$	$\beta_{-3} \sin(2\pi(f_s - 3f_{in})nT_s)$	← HD3
\vdots	$\beta_{-5} \sin(2\pi(f_s - 5f_{in})nT_s)$	
\vdots	\vdots	

By calculation $\beta_3 = \beta_{-3}$

$3f_{in}$ and $f_s - 3f_{in}$ components are cancelled



CANNOT measure ADC HD3.

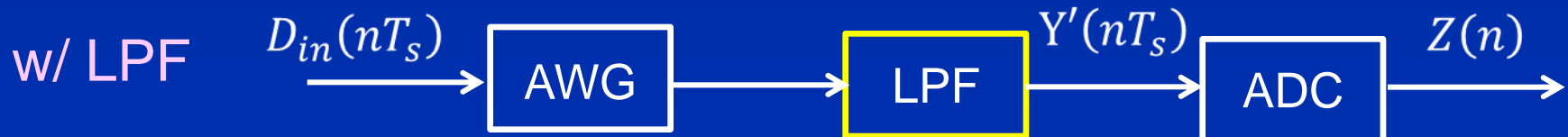
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AWG Output and Low Pass Filtering

$$\begin{aligned} Y'(nT_s) &= a_1 D_{in}(n) + a_3 \{D_{in}(n)\}^3 \\ &= P \sin(2\pi f_{in} nT_s) \\ &\quad + q Q \cos\left(2\pi \left(\frac{f_s}{2} - f_{in}\right) nT_s\right) \\ &\quad + r R \cos\left(2\pi \left(\frac{f_s}{2} - 3f_{in}\right) nT_s\right) \end{aligned}$$

$0 < q < 1$, $0 < r < 1$: spurious attenuation



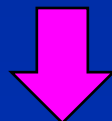
Application of Phase Switching and LPF

ADC output $Z(n)$ components

α'_0	$\beta'_1 \sin(2\pi f_{in} nT_s)$	← signal
$\alpha'_{-1} \cos\left(2\pi\left(\frac{f_s}{2} - f_{in}\right)nT_s\right)$	$\beta'_3 \sin(2\pi 3f_{in} nT_s)$	↙ HD3
$\alpha'_1 \cos\left(2\pi\left(\frac{f_s}{2} + f_{in}\right)nT_s\right)$	$\beta'_{-1} \sin(2\pi(f_s - f_{in})nT_s)$	↘ HD3
$\alpha'_{-3} \cos\left(2\pi\left(\frac{f_s}{2} - 3f_{in}\right)nT_s\right)$	$\beta'_{-3} \sin(2\pi(f_s - 3f_{in})nT_s)$	↙ HD3
\vdots	$\beta'_{-5} \sin(2\pi(f_s - 5f_{in})nT_s)$	
\vdots	\vdots	

By calculation $|\beta'_3| \gg |\beta'_{-3}|$

$3f_{in}$ and $f_s - 3f_{in}$ components are NOT cancelled



ADC HD3 can be measured.

Spurious Attenuation Effect

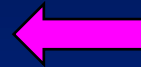
AWG output and low pass filtering

$$Y(nT_s) = P \sin(2\pi f_{in} nT_s)$$

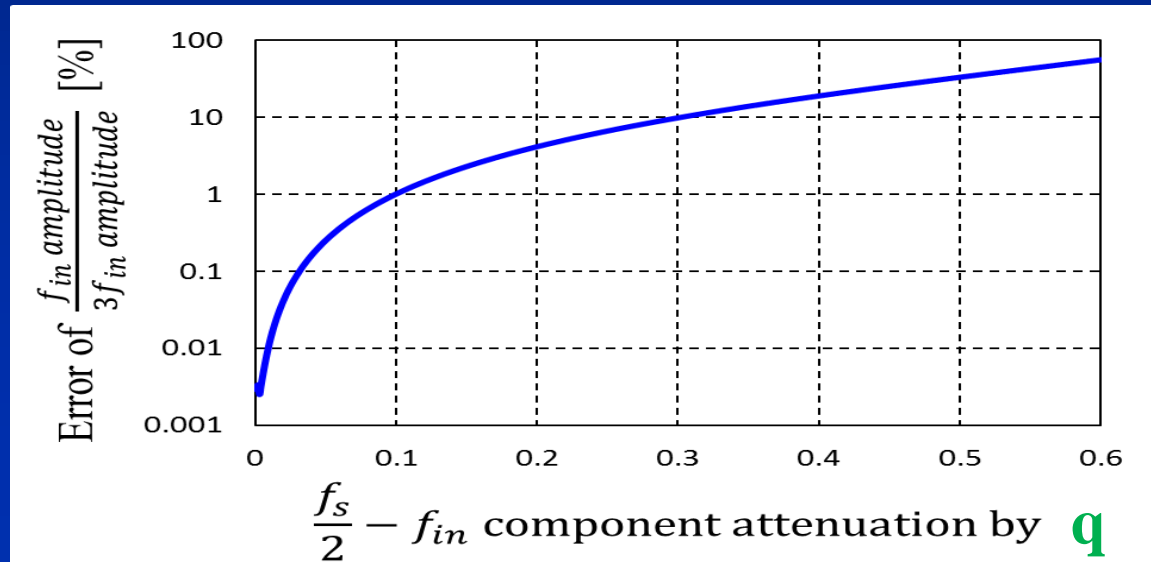
$$+ \mathbf{q} Q \cos\left(2\pi \left(\frac{f_s}{2} - f_{in}\right) nT_s\right)$$

$$+ \mathbf{r} R \cos\left(2\pi \left(\frac{f_s}{2} - 3f_{in}\right) nT_s\right)$$

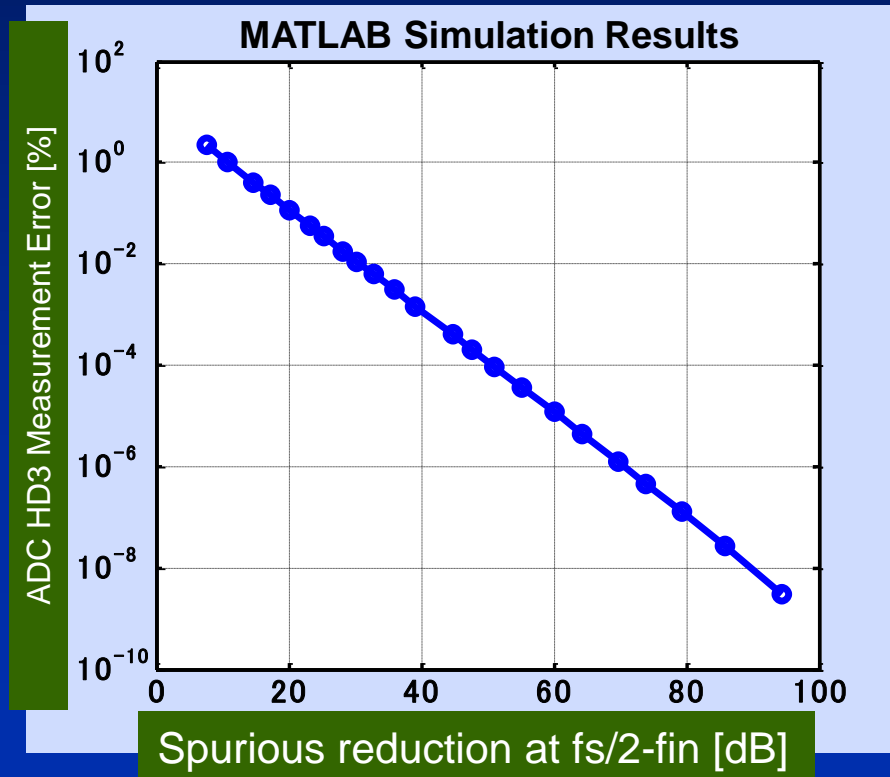
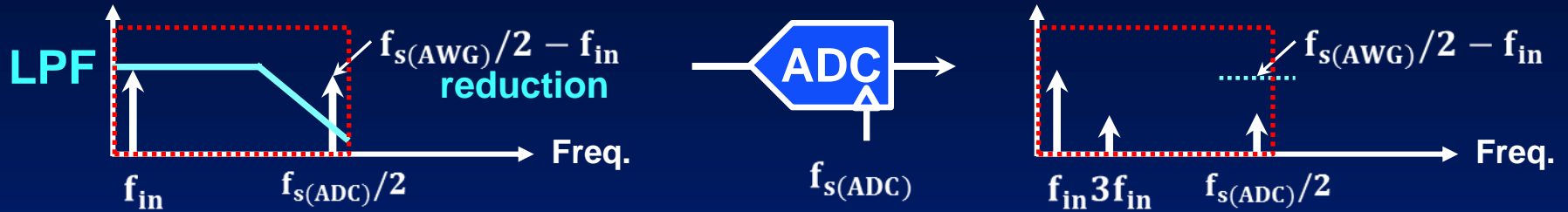
$f_{s(AWG)}/2 - f_{in}$
spurious reduction



Accurate HD3 measurement



Spurious Reduction at $f_{s(AWG)}/2 - f_{in}$



Spurious reduction
at $f_s/2 - f_{in}$

HD3 measurement
Error

10dB



1%

20dB



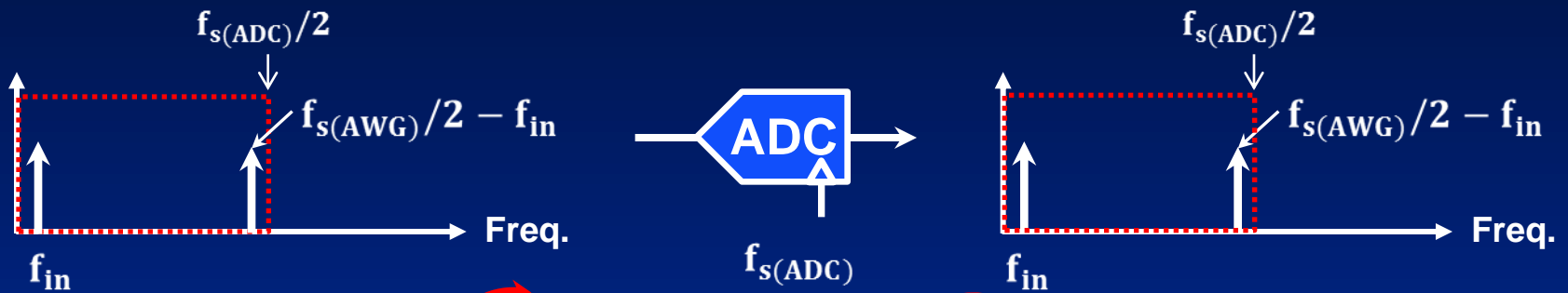
0.1%

For accurate ADC HD3 measurement
attenuate the spurious at $\frac{f_{s(AWG)}}{2} - f_{in}$

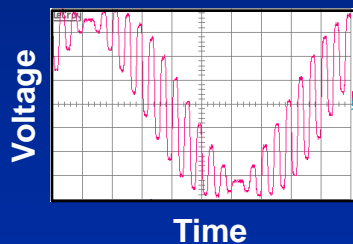
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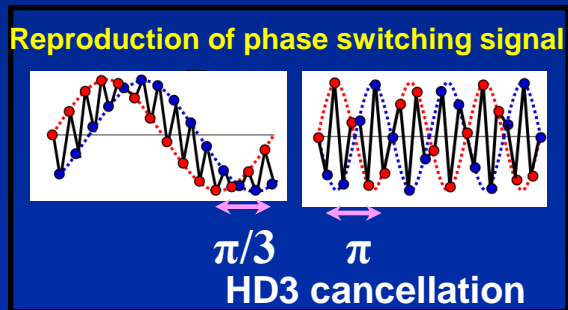
When Spurious @ $f_{s(AWG)}/2 - f_{in}$ is Within $f_{s(ADC)}/2$



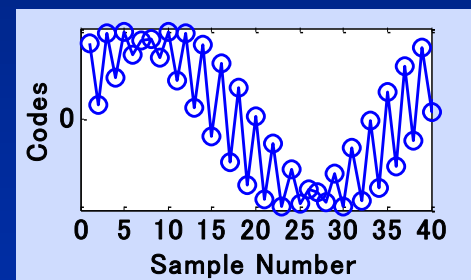
Analog input



Inside ADC



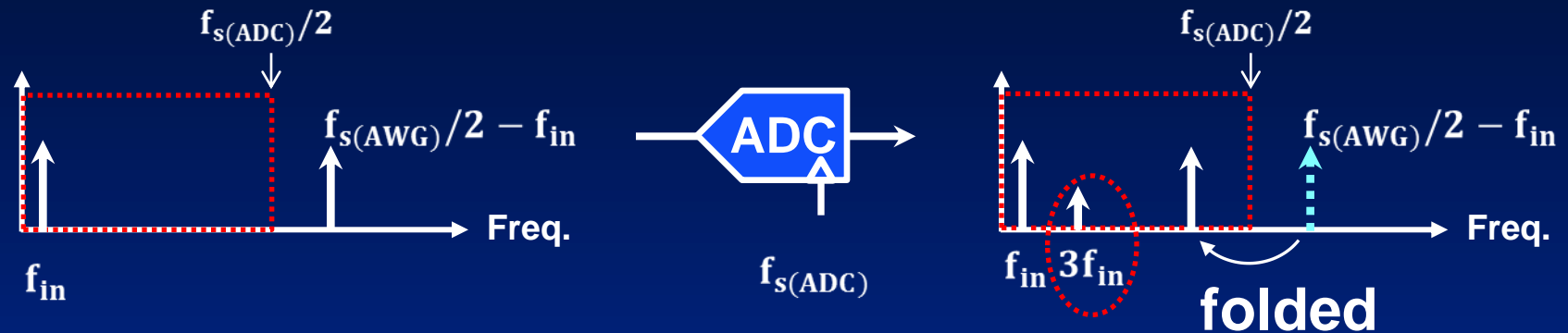
Digital output



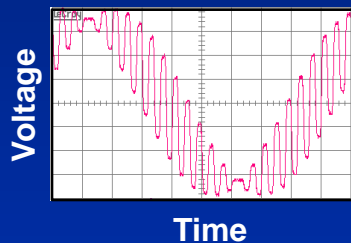
Similar

ADC HD3 can NOT be measured accurately

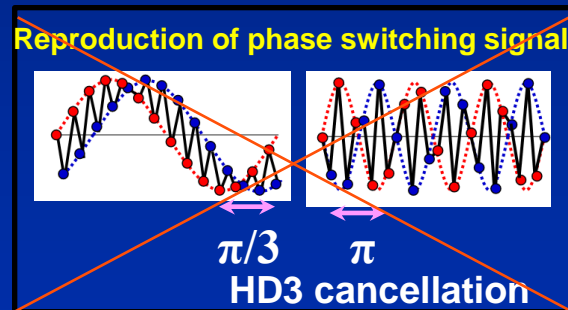
When Spurious @ $f_{s(\text{AWG})}/2 - f_{\text{in}}$ is Beyond $f_{s(\text{ADC})}/2$



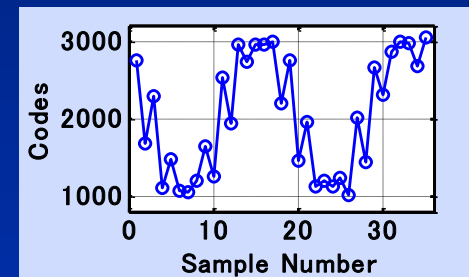
Analog input



Inside ADC



Digital output

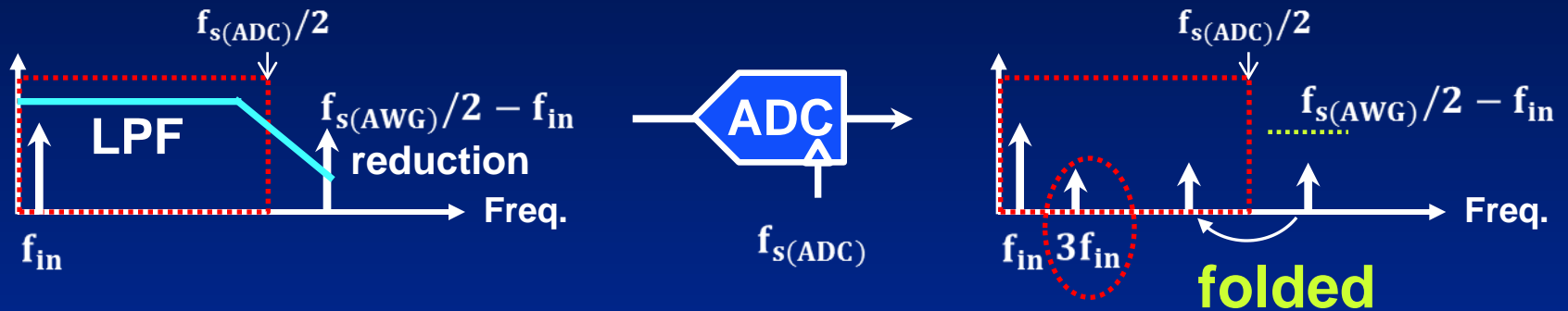


Different

ADC HD3 can be measured accurately

Conditions for Accurate ADC HD3 Measurement

$$\frac{f_{s(\text{AWG})}}{2} - f_{\text{in}} > \frac{f_{s(\text{ADC})}}{2}, \text{ and/or Spurious Reduction at } \frac{f_{s(\text{AWG})}}{2} - f_{\text{in}}$$



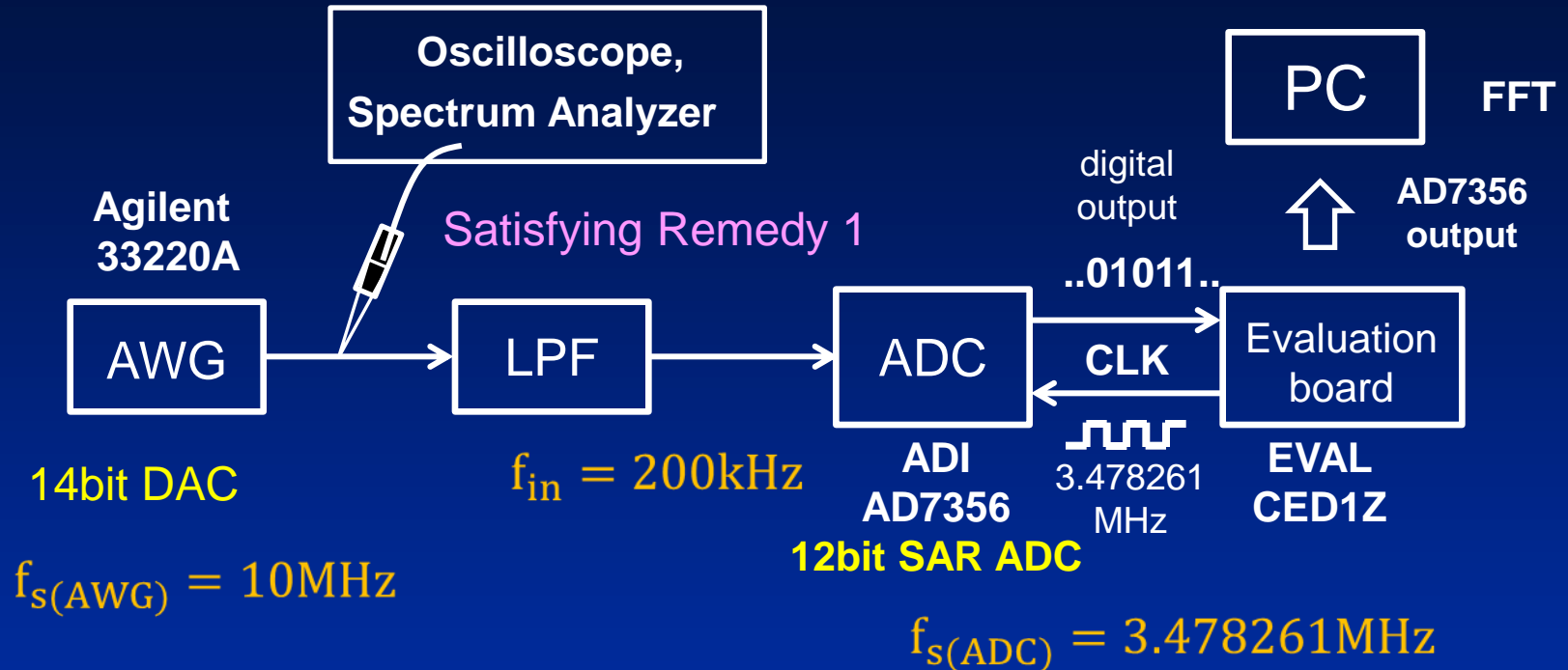
- ① Spurious power reduction at $f_{s(\text{AWG})}/2 - f_{\text{in}}$
- ② $\frac{f_{s(\text{AWG})}}{2} - f_{\text{in}} > \frac{f_{s(\text{ADC})}}{2}$ Change sampling frequency by folding

Accurate measurement of ADC output HD3 with phase switching signal

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ADC 3rd Harmonic Measurement Diagram



ADC test signal generation

- Conventional
- Phase switching

Satisfying Remedy 2

$$\frac{f_{s(AWG)}}{2} - f_{in} > \frac{f_{s(ADC)}}{2}$$

Experimental Environment for ADC Testing

PC1 for test signal program

Spectrum Analyzer
for test signal analysis

$f_s(\text{AWG})=10\text{MHz}$, $f_{in}=200\text{kHz}$
 $f_s(\text{ADC})=3.476261\text{MHz}$

Oscilloscope for test signal analysis

AWG 33220A
for test signal
generation

Common mode
noise suppression
by a choke coil

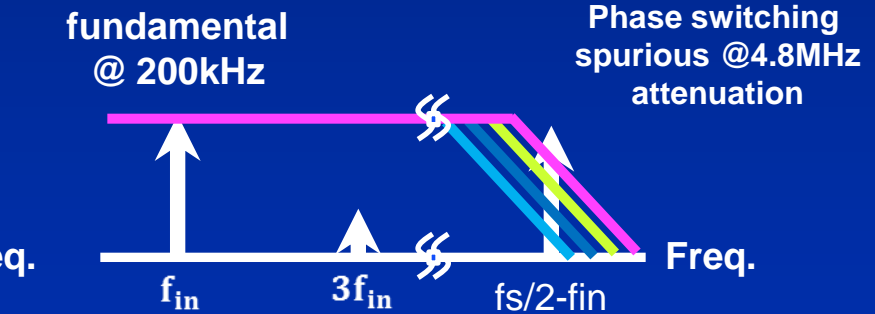
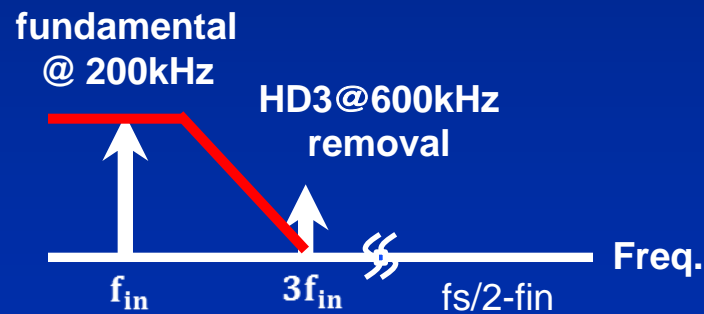
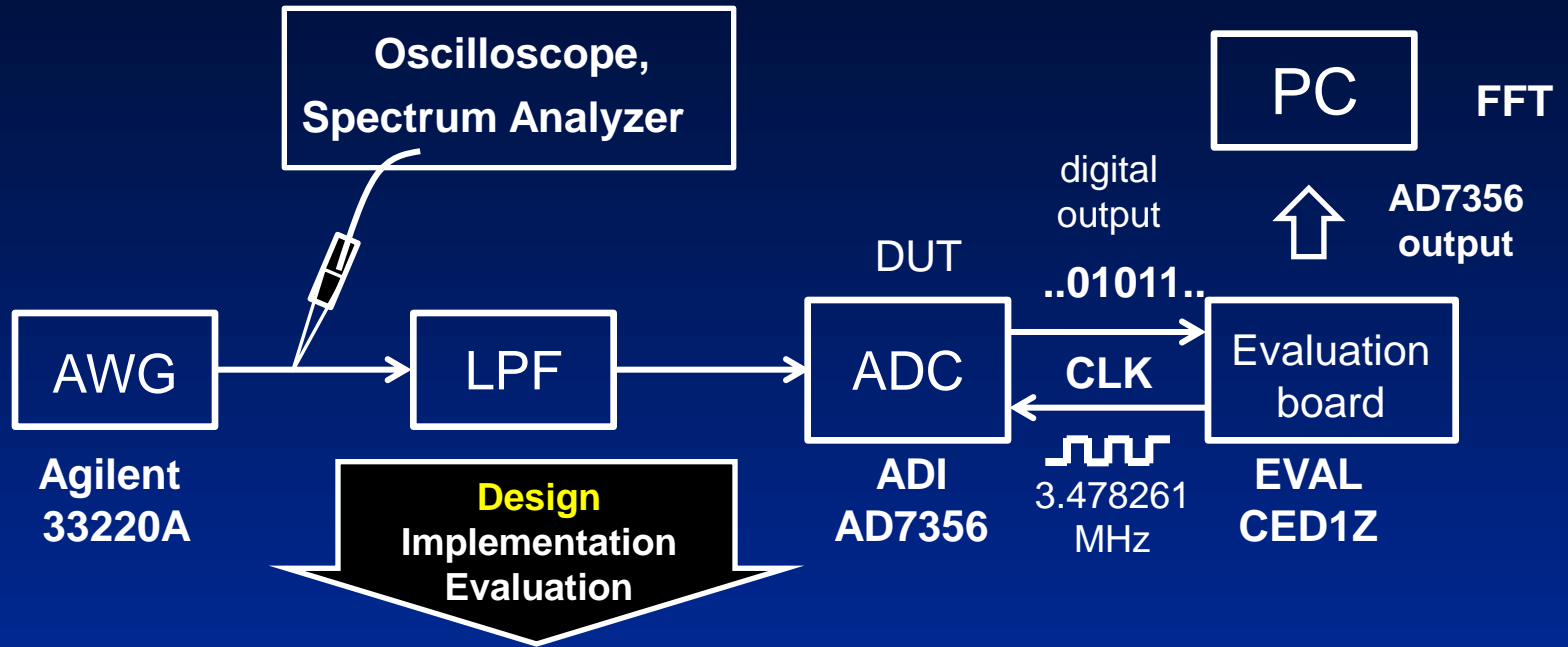
PC2 for
ADC output
analysis

Low Pass Filters

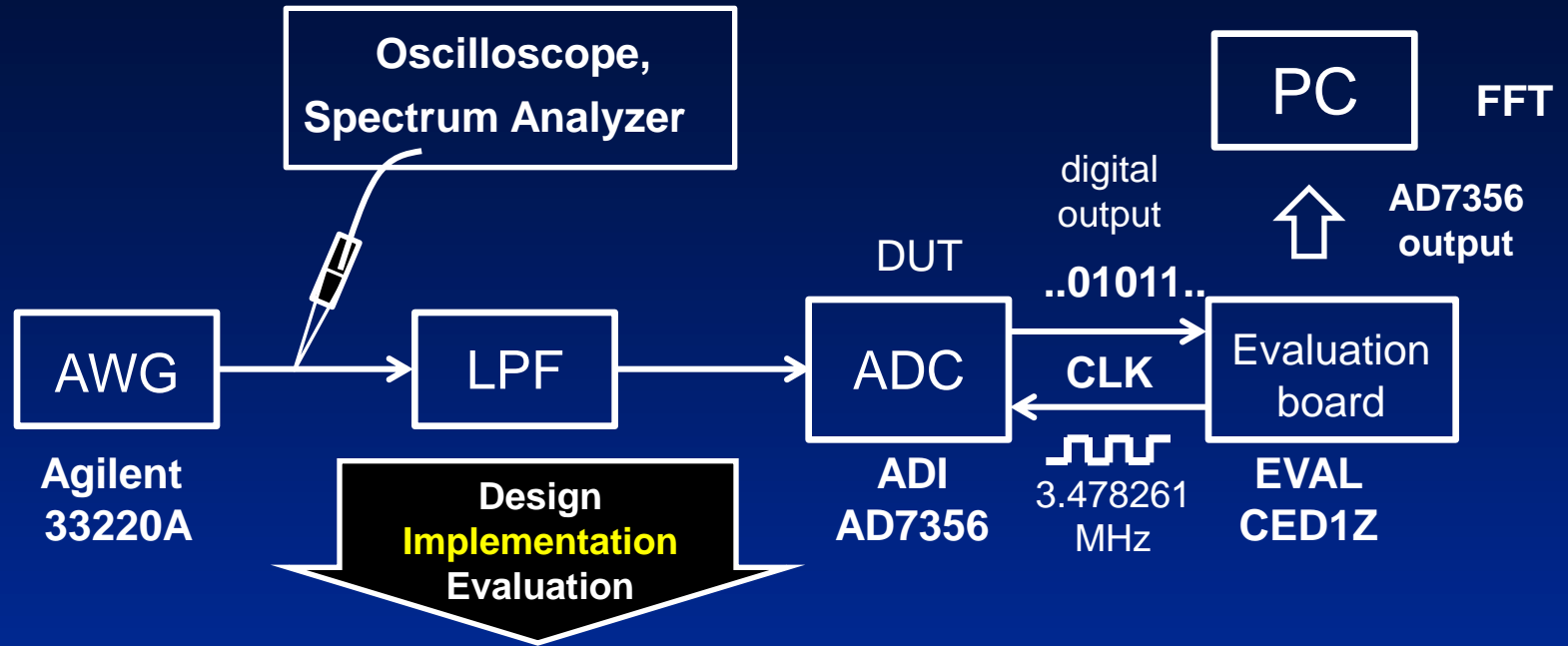
EVAL-CED1Z for generating
sampling clock , etc.

ADC AD7356 (12bit)
6 samples
Device Under Test

Analog LPF Design for ADC Testing



LPF Implementation

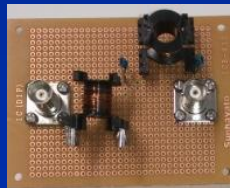


$f_c=250\text{kHz}$

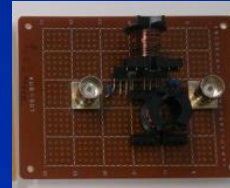


for HD3 reduction

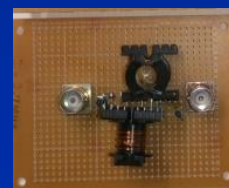
$f_c=1\text{MHz}$



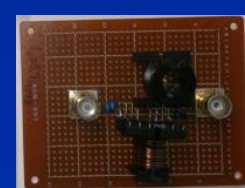
$f_c=2\text{MHz}$



$f_c=2.7\text{MHz}$

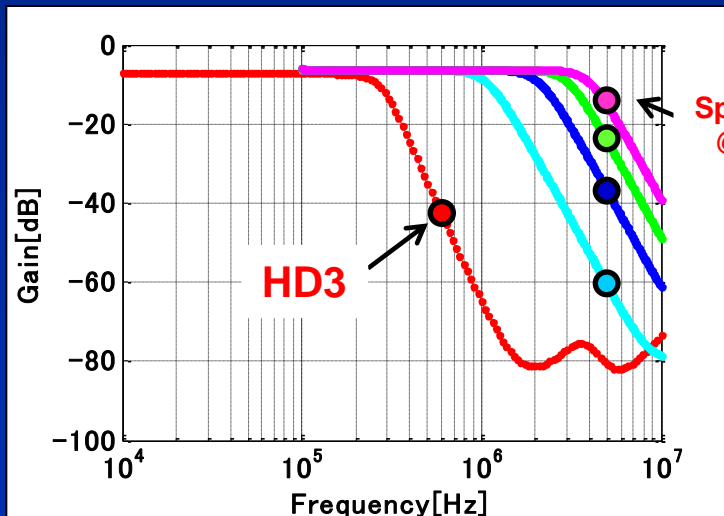
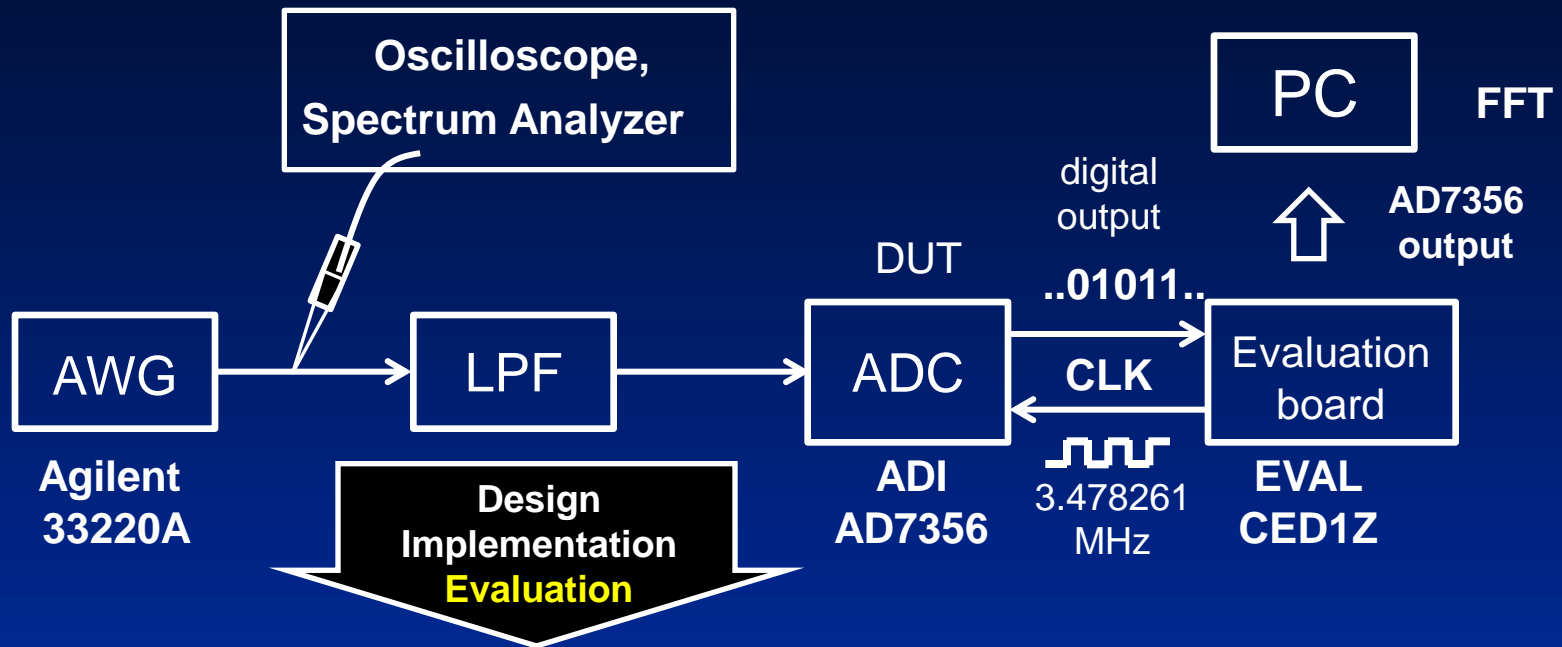


$f_c=3.7\text{MHz}$



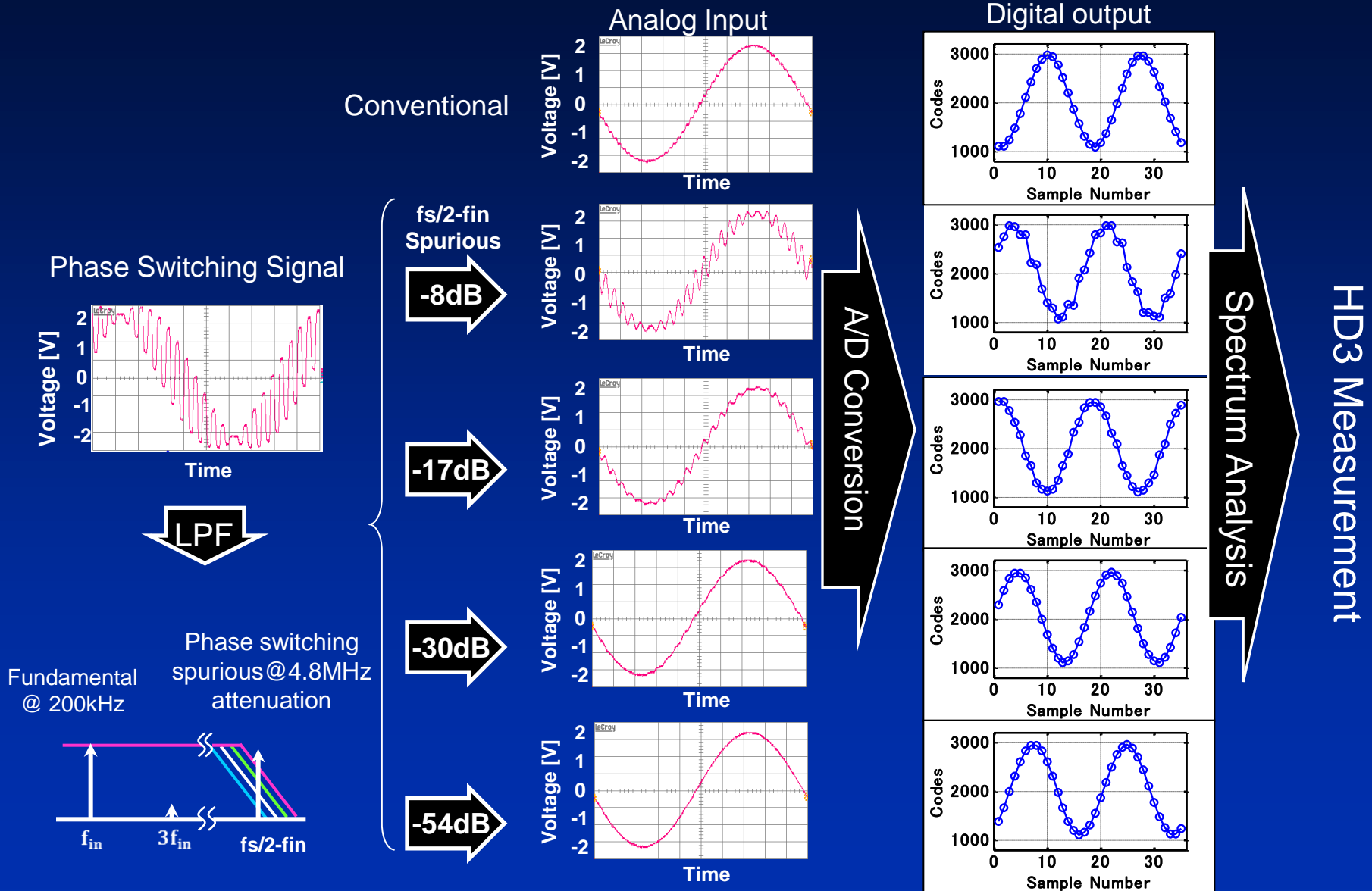
for $f_s/2$ -bin spurious reduction

LPF Evaluation with Frequency Response Analysis

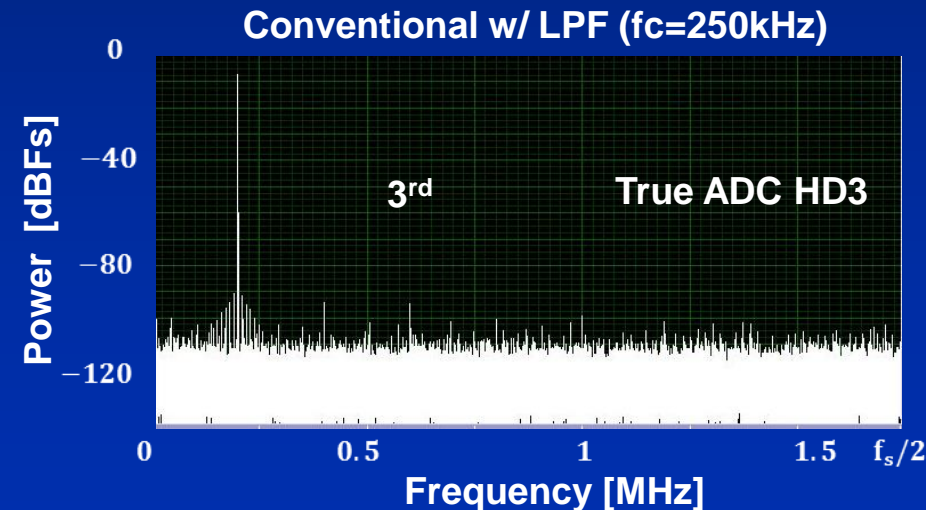
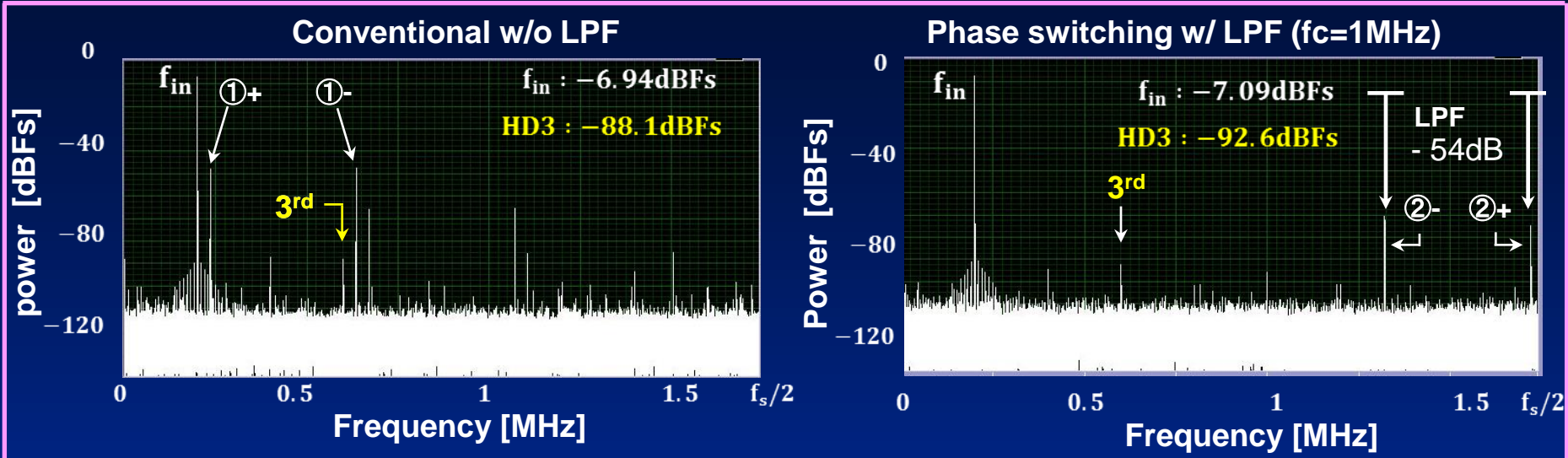


- $f_c=250\text{kHz} \rightarrow -40\text{dB @ } 600\text{kHz}$
- $f_c=1\text{MHz} \rightarrow -54\text{ dB @ } 4.8\text{MHz}$
- $f_c=2\text{MHz} \rightarrow -30\text{ dB @ } 4.8\text{MHz}$
- $f_c=2.7\text{MHz} \rightarrow -17\text{ dB @ } 4.8\text{MHz}$
- $f_c=3.7\text{MHz} \rightarrow -8.0\text{ dB @ } 4.8\text{MHz}$

Measured Waveforms of ADC Input and Output



Comparison of Conventional and Proposed Methods



True ADC HD3: -94.6dBFs

Conventional method
Measured HD3: -88.1dBFs
Error 6.8%

Proposed method
Measured HD3: -92.6dBFs
Error 2.1%

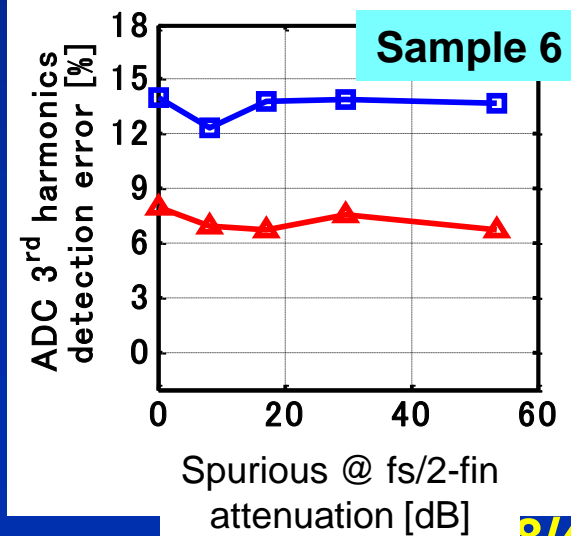
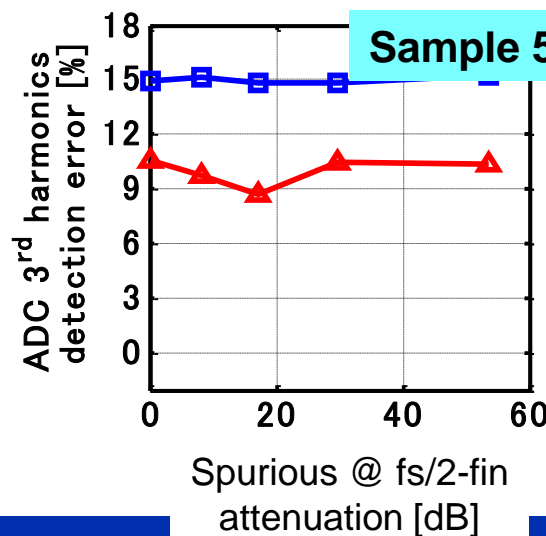
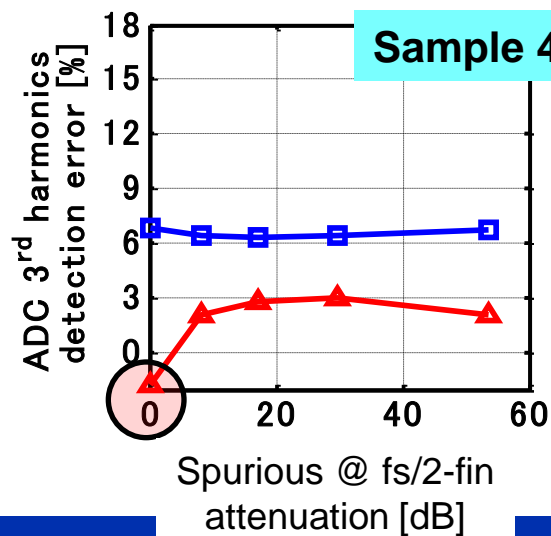
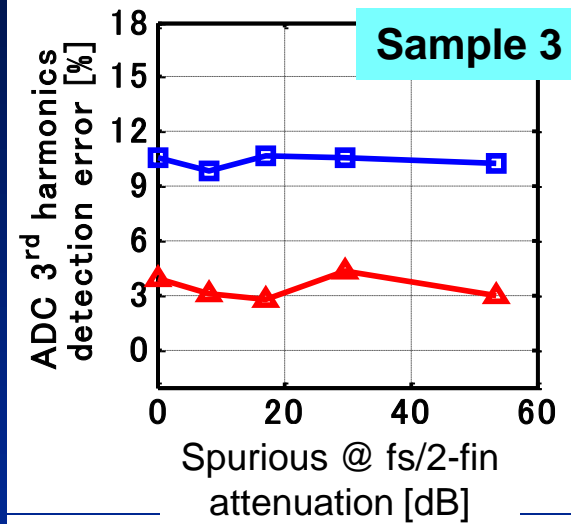
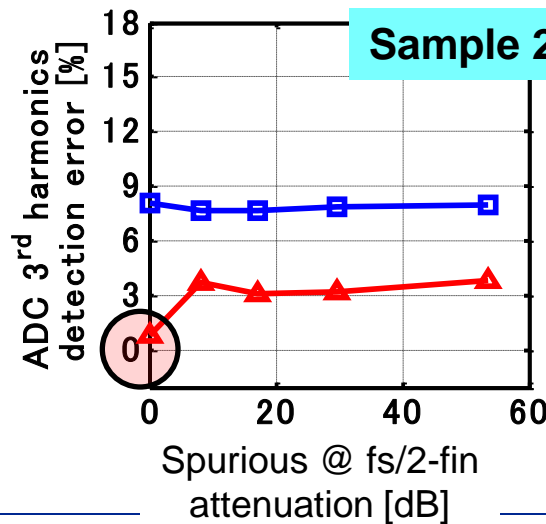
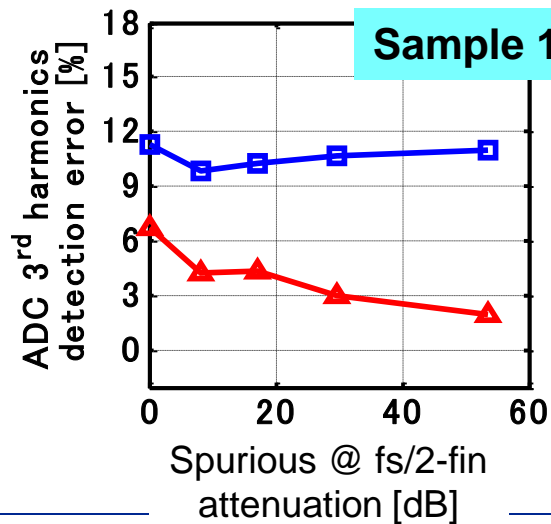
$$\textcircled{1} \pm : f_{s(\text{AWG})} \pm f_{\text{in}}$$

$$\textcircled{2} \pm : \frac{f_s}{2} \pm f_{\text{in}}$$

ADC HD3 Measurement Results



ADC (AD7356) HD3 measurement error reduction is verified

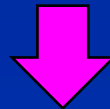


Outline

- Research background
- Proposed solution
- Problems of proposed solution
- Remedy 1
- Remedy 2
- Experimental results
- Conclusion

Conclusions

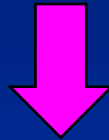
- Low distortion sine signal generation
- Without AWG hardware modification
- Just AWG program change
and a simple analog LPF
- Verified with
AWG (Agilent 33220)
6 samples of 12bit SAR ADC (AD7356)



Greatly improved quality of ADC Linearity testing
at virtually no extra cost.

Future Work

- Generalization to other types of low distortion sinusoidal signal generation
 - HD2, HD2&HD3 cancellation for 1-tone
 - IMD3 cancellation for 2-tone



- We have partially verified these.
- Detailed theoretical analysis, simulations, and experiments are underway.