



Self-heat Modeling of Multi-finger n-MOSFETs for RF-CMOS Applications

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Outline

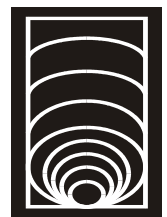
- Research Background
- Purposes of This Work
- Investigation of Self-heating in a Multi-finger n-MOSFET with a 2-D Device Simulator
- Model Derivations
- Measurements and Model Verifications
- Summary and Future Research



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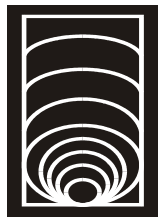
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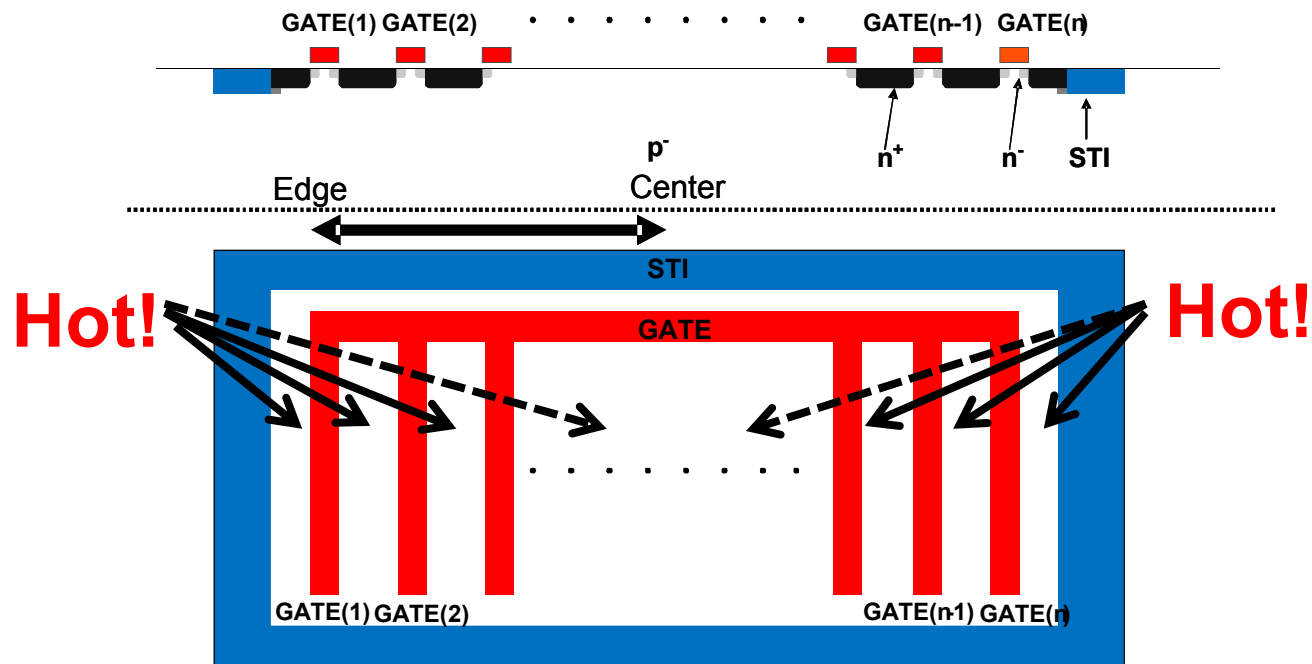
Research Background (1)

- A multi-finger structure is popularly used in MOSFETs for various RF-CMOS circuits including power amplifiers, mixers, and oscillators
- There is an inconsistency between *S-parameters* and static drain current simulations despite accurate model parameter extractions



Research Background (2)

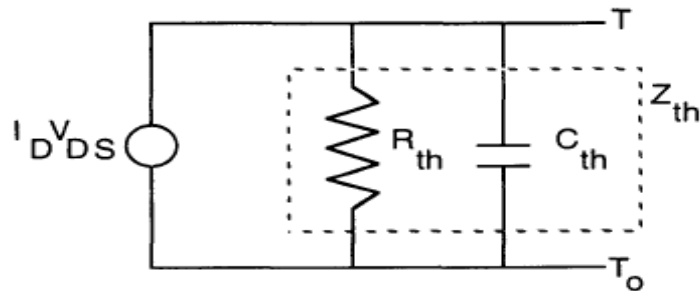
- In bulk MOSFETs for the multi-finger structure, self-heating effect (SHE) may occur especially if shallow trench isolation (STI) technology is adopted



Research Background (3)

- A sub-circuit based self-heat model does not converge in large circuits

1. Temperature terminals are added to the model equivalent circuit as a sub-circuit



Z_{th} : Thermal Impedance

V_{DS}, I_D : Drain Voltage and Current

2. Operation temperature

$$T = T_0 + \underbrace{(I_D V_{DS}) Z_{th}}_{\text{Rises in device temperature}}$$



Rises in device temperature

Ambient temperature



3. Main model circuit simulation



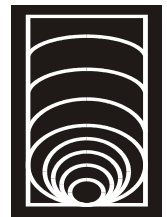
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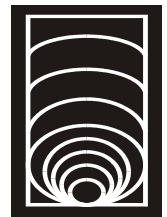
Purposes of This Work

- To analyze self-heat mechanisms in multi-finger n-channel MOSFETs
- To develop a general self-heat model without using thermal sub-circuits
- To analyze and modeling fin-number dependencies of thermal resistance with DC and *S-parameter* measurements and simulations

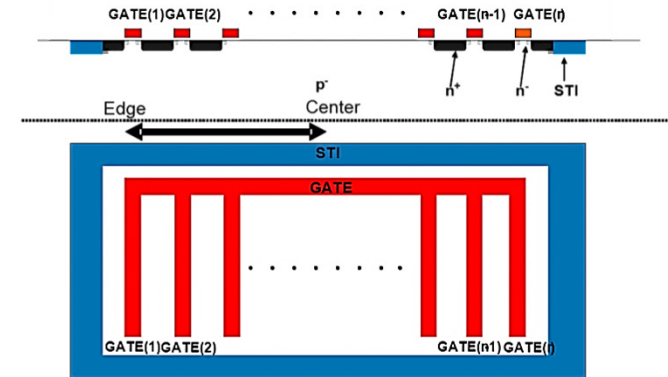
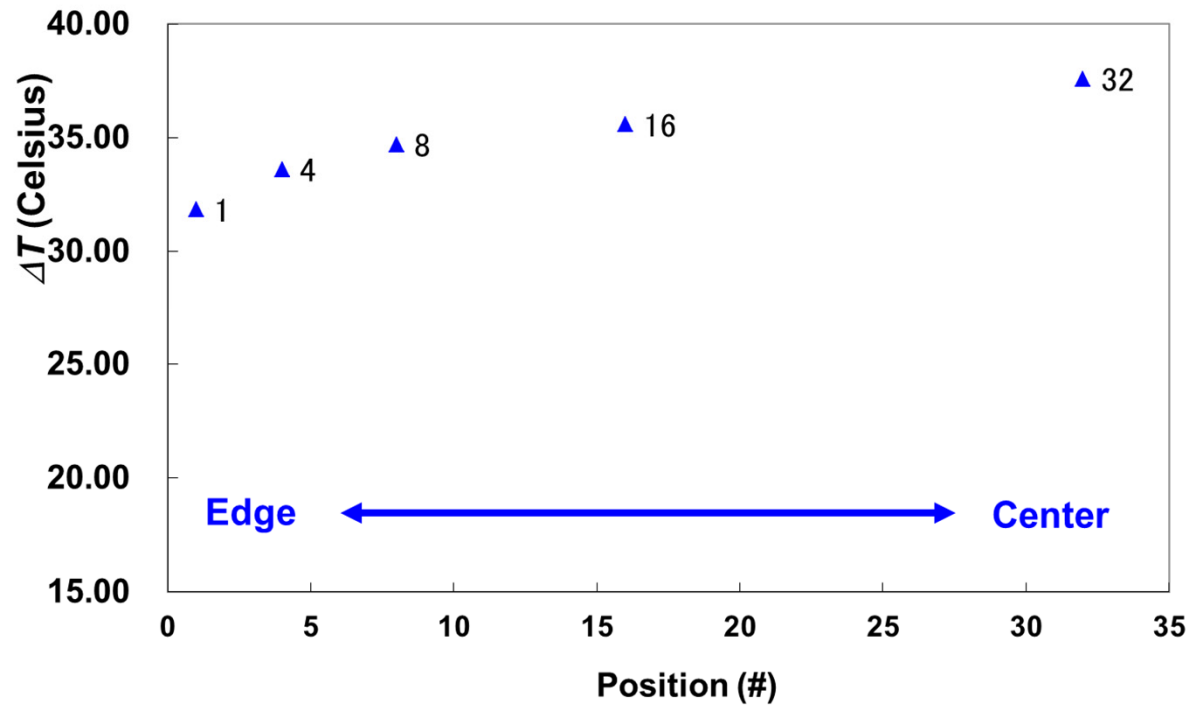


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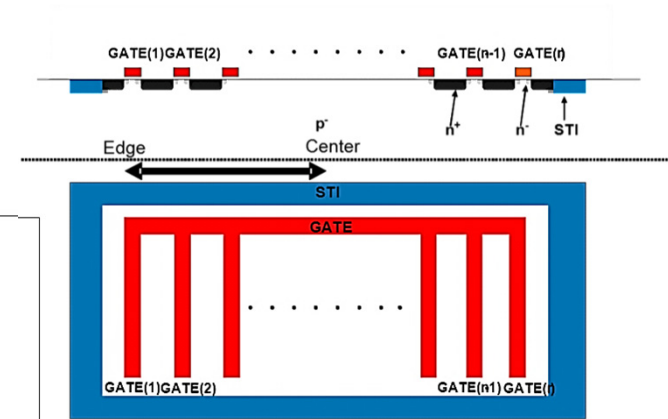
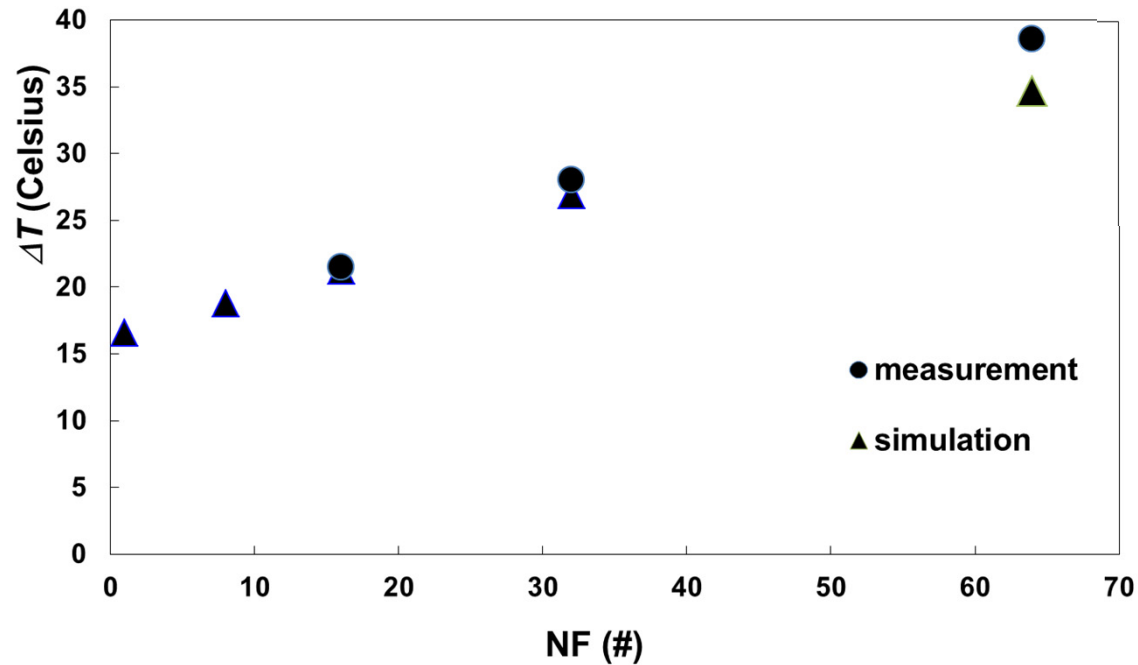
Device Simulation of Self-heating Induced Temperature Distribution



- Simulated with a 2-D device simulator (PISCES-2HB)
- A slow pulsed DC source was used for better convergence



Dependence of ΔT on the number of fins



- The gate width of each fin is 20 μm
- Simulated ΔT is obtained at the center fin
- Measurement was made by using DC source/monitor



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Temperature Dependence on Resistance

The DC and Isothermal current is written as

$$I_{ds}(V_{ds}, T_{dev}) = I_{iso} \left[V_{ds}, R_{th} \cdot V_{ds} \cdot I_{ds}(V_{ds}, T_{dev}) + T_{dev} \right] \quad (1)$$

ΔT is defined as

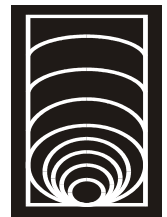
$$\Delta T = I_{ds} \cdot V_{ds} \cdot R_{th} \quad (2)$$

R_{th} can be written as an electrical resistance equation by

$$R_{th} = \rho \frac{L}{S} \quad (3)$$

Temperature dependence is given by

$$R_{th}(T_{dev} + \Delta T) = \rho(T_{dev} + \Delta T) \frac{L}{S} \quad (4)$$



Thermal Resistance

Since ρ is linearly proportional to the rise in temperature, we have

$$\rho(T_{dev} + \Delta T) = \rho(T_{dev}) + c \cdot \Delta T \quad (5)$$

By plugging eq. (5) into eq. (4), we obtain

$$R_{th}(T_{dev} + \Delta T) = R_{th0} + c \cdot \frac{L}{S}(T_{dev}) \cdot R_{th0} \cdot \Delta T \quad (6)$$

Now we define K_{th} as

$$K_{th} = c \cdot \frac{L}{S}(T_{dev}) \cdot R_{th0} \quad (7)$$

R_{th} can be simply represented as

$$R_{th} = R_{th0} + K_{th} \cdot \Delta T \quad (8)$$



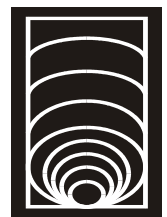
Thermal Impedance

For AC analysis, thermal capacitance, C_{th} , should be included in parallel with R_{th} , which is written as

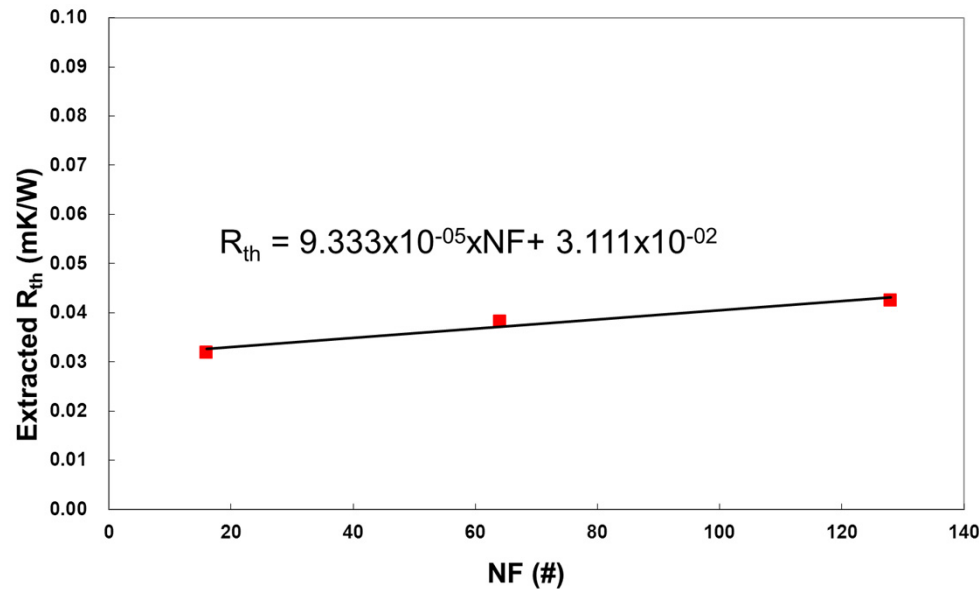
$$Z_{th} = \frac{R_{th}}{1 + j \cdot \omega \cdot C_{th} \cdot R_{th}} \quad (9)$$

Now eq. (2) becomes

$$\Delta T = I_{ds} \cdot V_{ds} \cdot Z_{th} \quad (10)$$



Fin-number Dependence on R_{th}



R_{th} is proportional to the number of fins, NF , of n-MOSFETs. R_{th} is a linear function as

$$R_{th}^{NF} = A \cdot NF + R_{th} \quad (11)$$

Z_{th} is replaced with

$$Z_{th}^{NF} = \frac{R_{th}^{NF}}{1 + j \cdot \omega \cdot C_{th} \cdot R_{th}^{NF}} \quad (12)$$



Drain Current with Self-heating

Temperature dependence of effective mobility is referred as

$$\mu_{eff}(T) = \mu_{eff}(T_{dev}) \frac{T}{T_{dev}} \quad (13)$$

Effective mobility with self-heating can be

$$\mu_{eff}(T_{dev} + \Delta T) = \frac{\mu_{eff}}{1.0 + \Delta T / T_{dev}} \quad (14)$$

Finally, a drain current with self-heating of a multi-finger n-MOSFET is written as

$$I_{ds_th} = \frac{I_{ds}}{1.0 + \Delta T_{NF} / T_{dev}} \quad (15)$$



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BSIM6 Model as a Modeling Vehicle

BSIM6 model

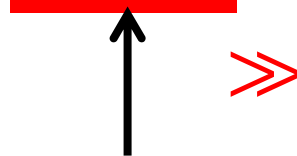
- is continuous in all operation regions
- has accurate derivatives to predict harmonic distortion
- is satisfied both Gummel symmetry and AC symmetry
- has better physical capacitance behavior
- supports Verilog-A code which is supplied by the authors



'Cold' DC Measurement (1)

- AC conductance method* with a Network Analyzer ('Cold' DC measurement) is developed

$$\frac{dI_d}{dV_d} = \frac{\partial I_d}{\partial T} \cdot \frac{\partial T}{\partial V_d} + \frac{\partial I_d}{\partial V_d} \Big|_{T_{dev}}$$

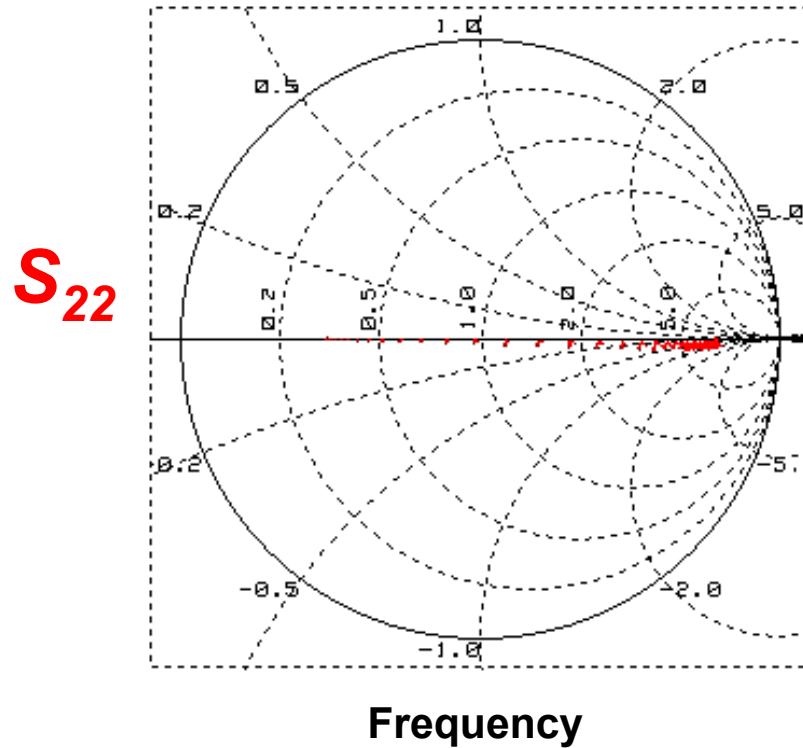


**Can be neglected
at high frequencies**

*R. H. Tu. et.al, IEEE EDL, vol. 16, Feb. 1995.



'Cold' DC Measurement (2)



$$S_{22} - V_d$$



$$G_{ds} - V_d$$

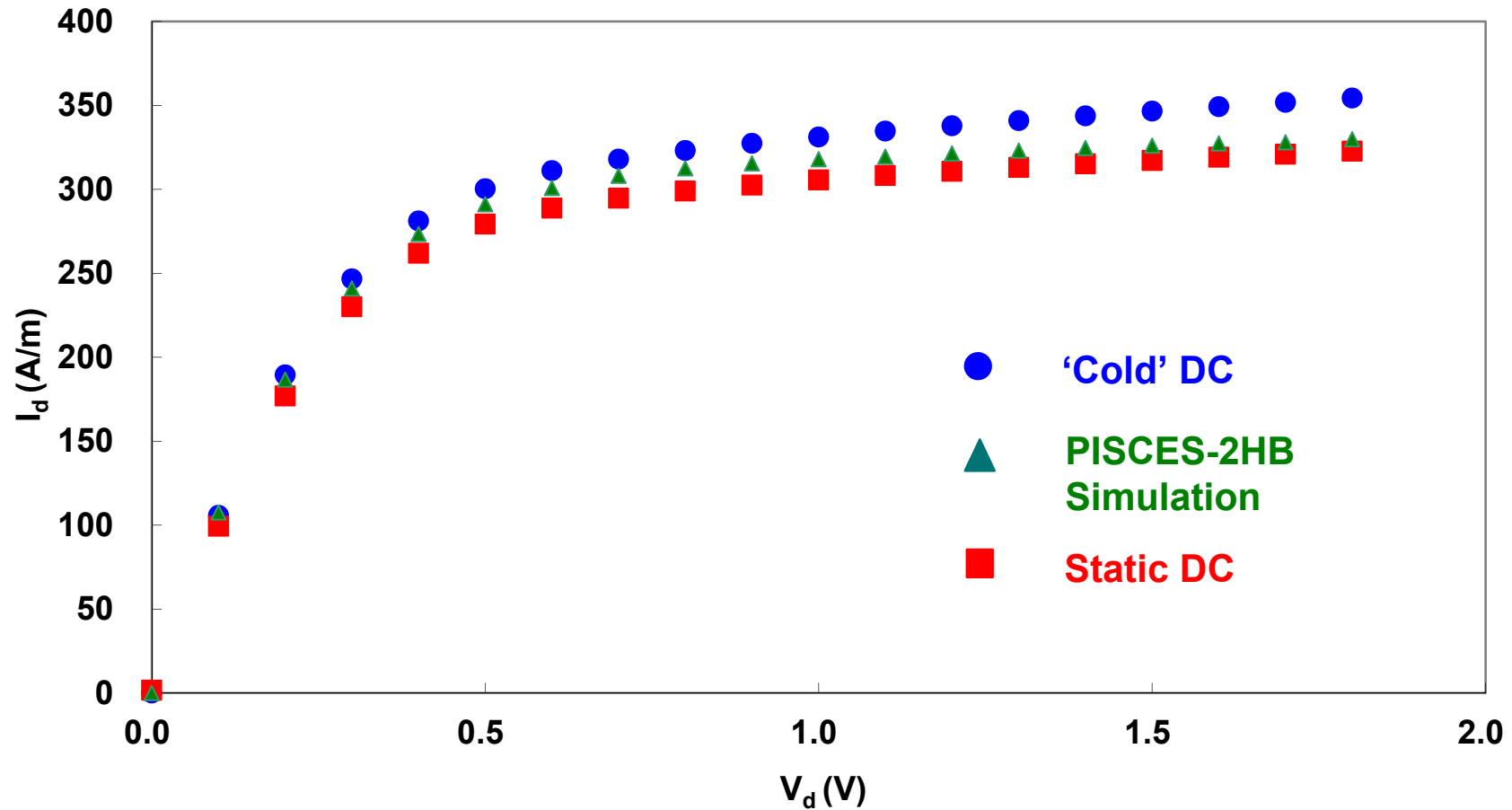


$$I_d - V_d$$



I_d - V_d Measurement

Drain current characteristics of
64-fin n-MOSFET

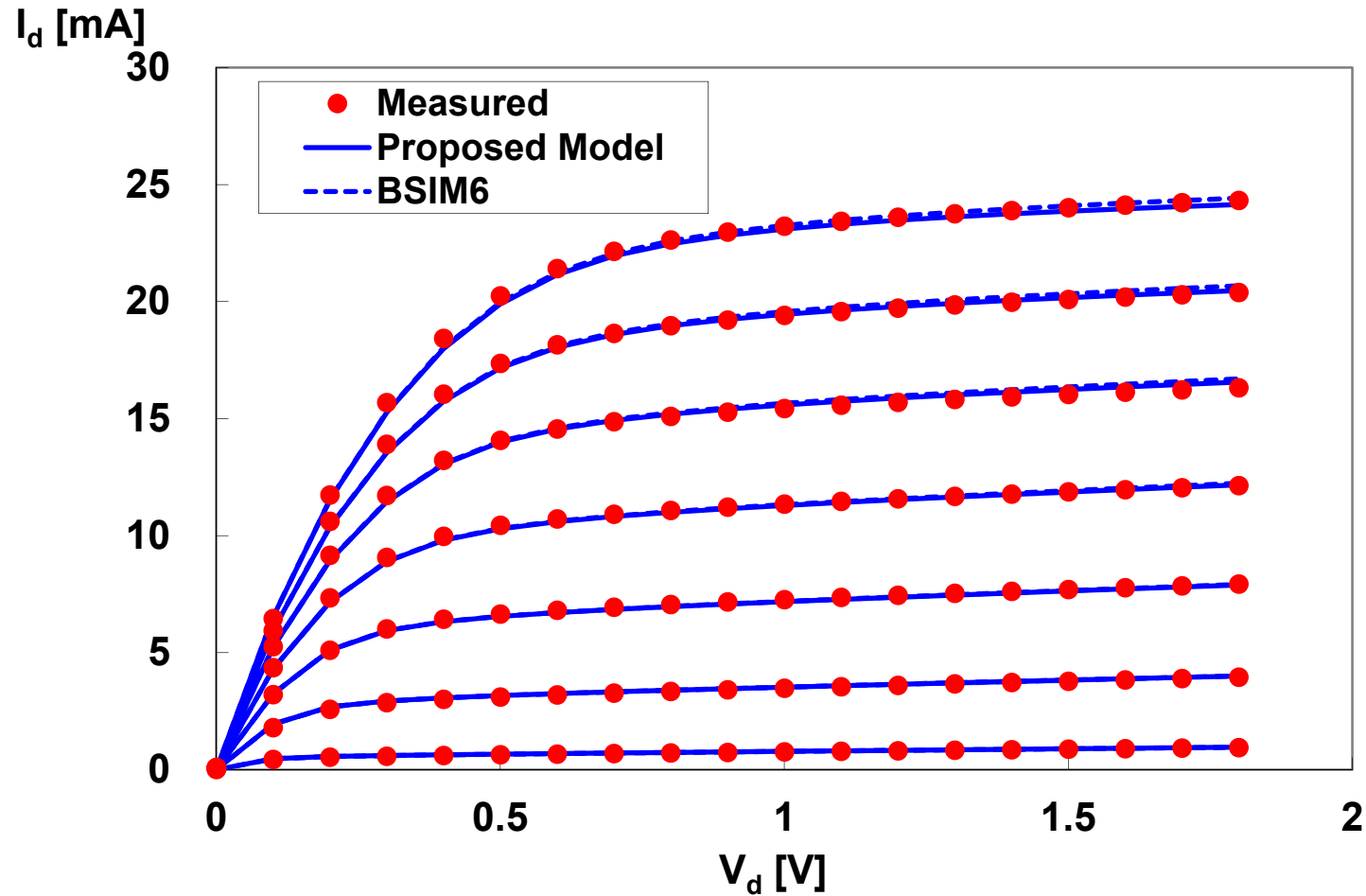


Model Parameter Extractions

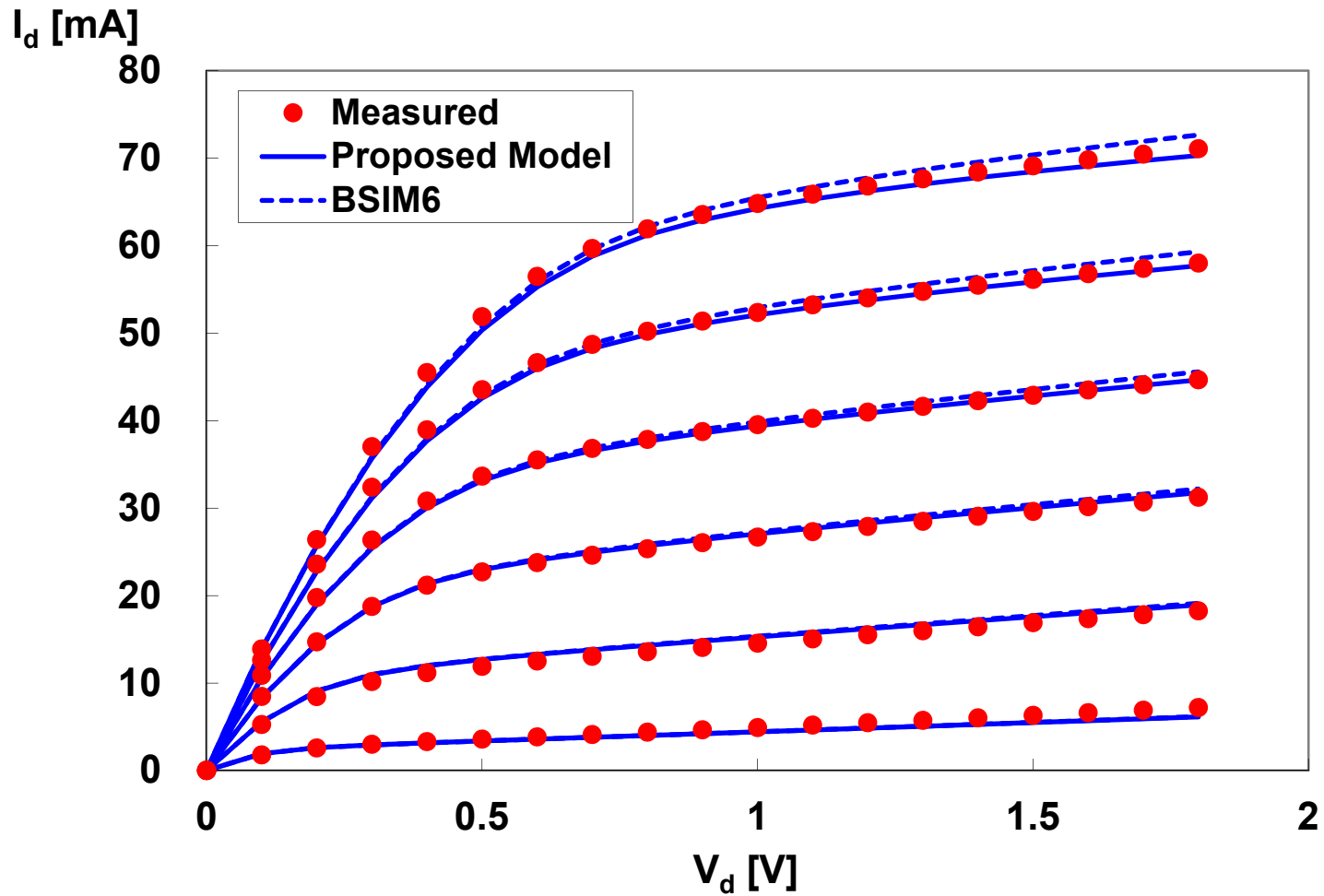
1. Input process parameters for BSIM6
2. DC I-V measurement
3. Measurements of *S-parameters* and de-embedded parasitic components, which are used for 'Cold' DC calculations and AC parameter extractions
4. Extractions of DC parameters including BSIM6 and SHE parameters
5. AC parameter (*L*, *C*, and *R*) extractions
6. Model verifications with small circuit modules



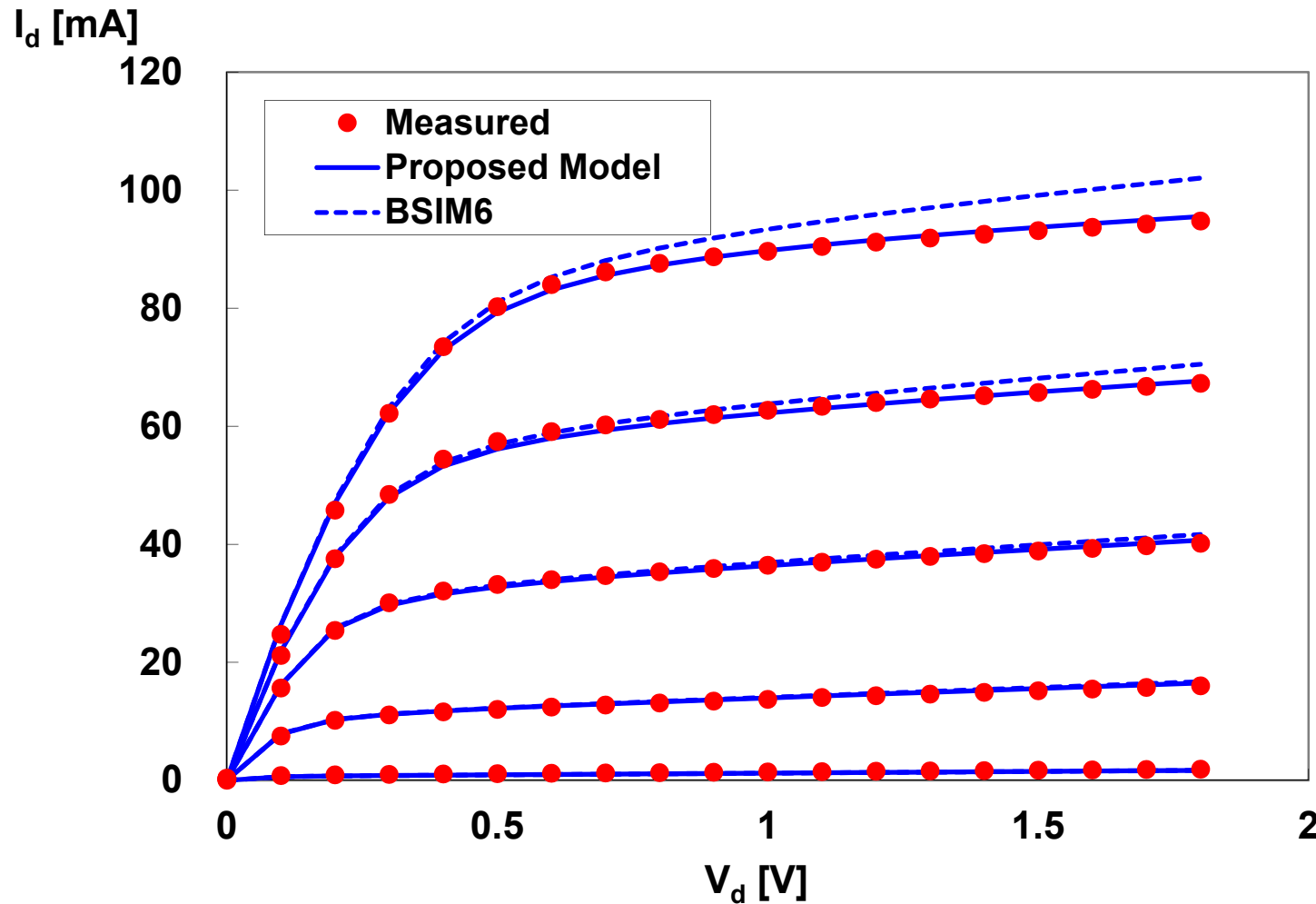
DC Drain Current Characterization of 16-fin n-MOSFET



DC Drain Current Characterization of 64-fin n-MOSFET

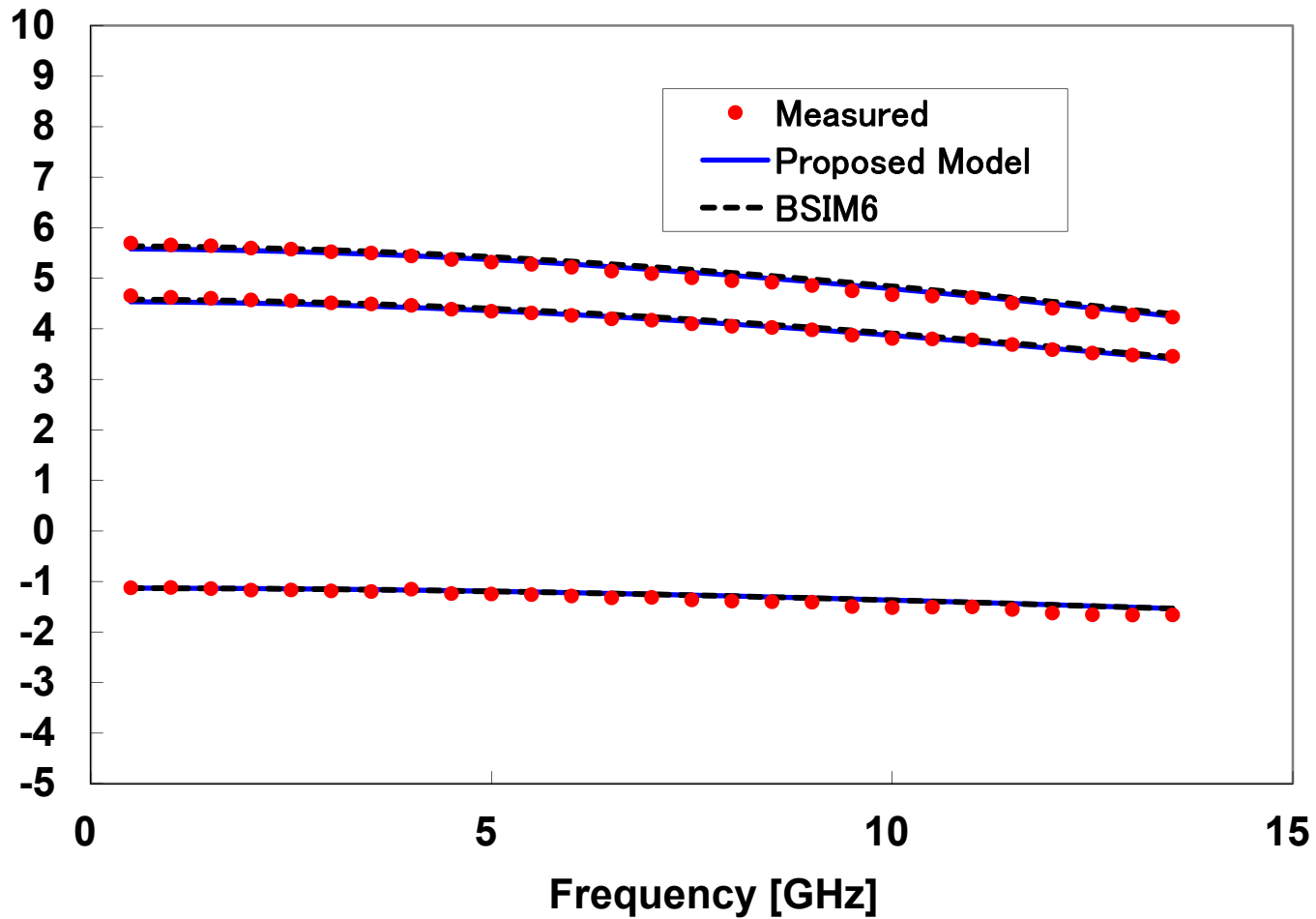


DC Drain Current Characterization of 128-fin n-MOSFET



S_{21} Characterization of 16-fin n-MOSFET

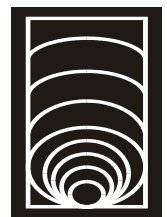
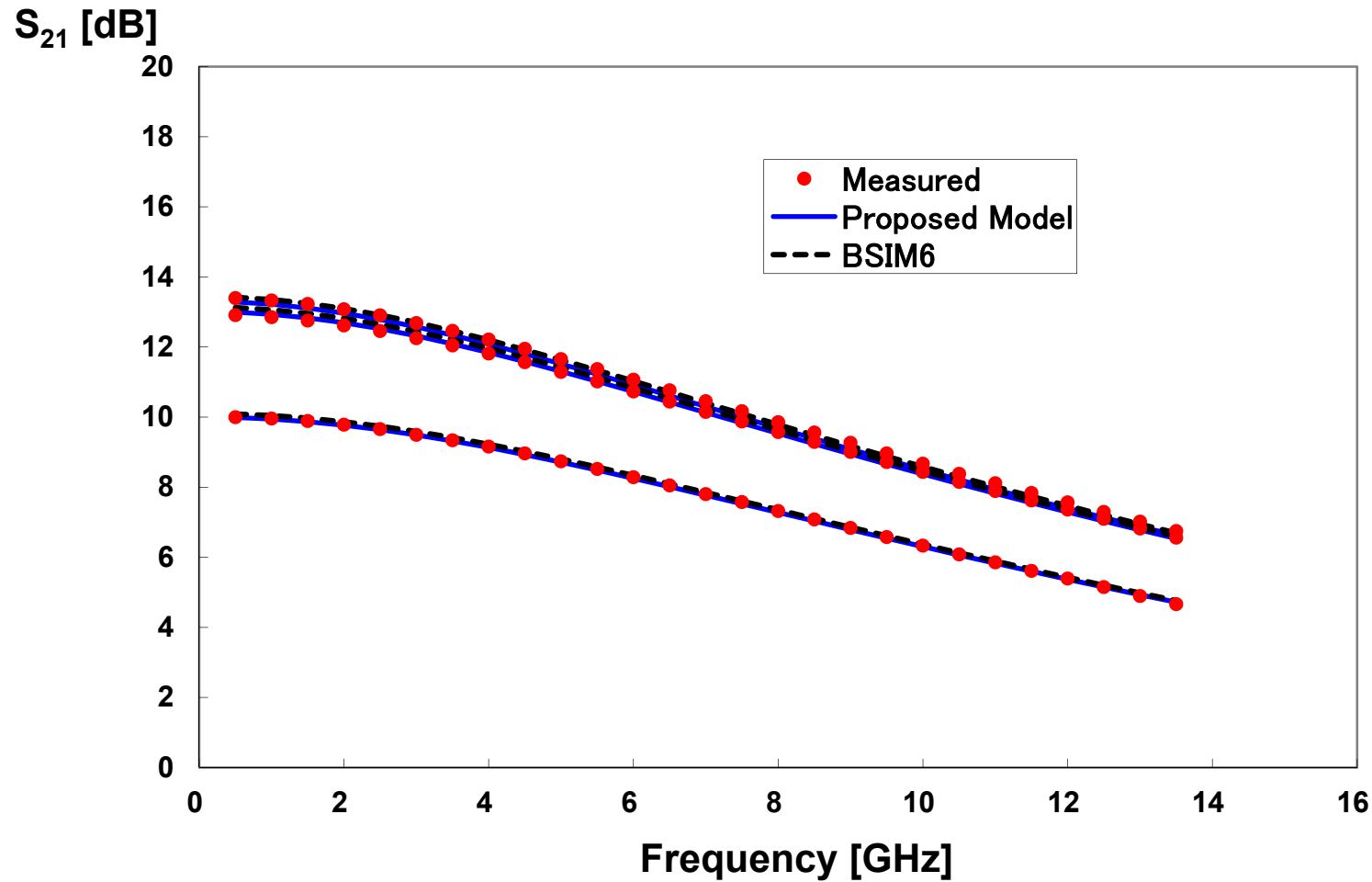
S_{21} [dB]



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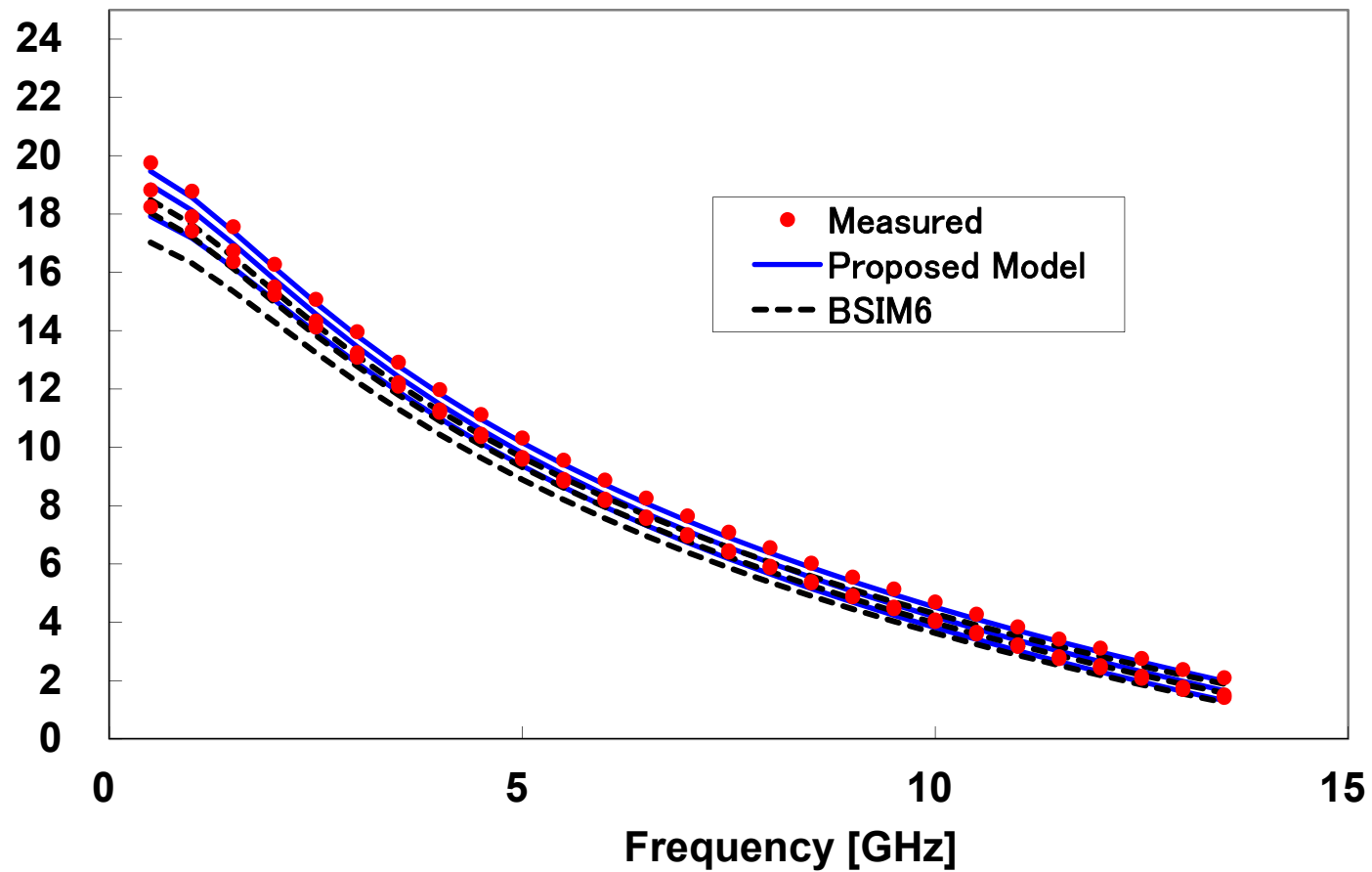


S_{21} Characterization of 64-fin n-MOSFET

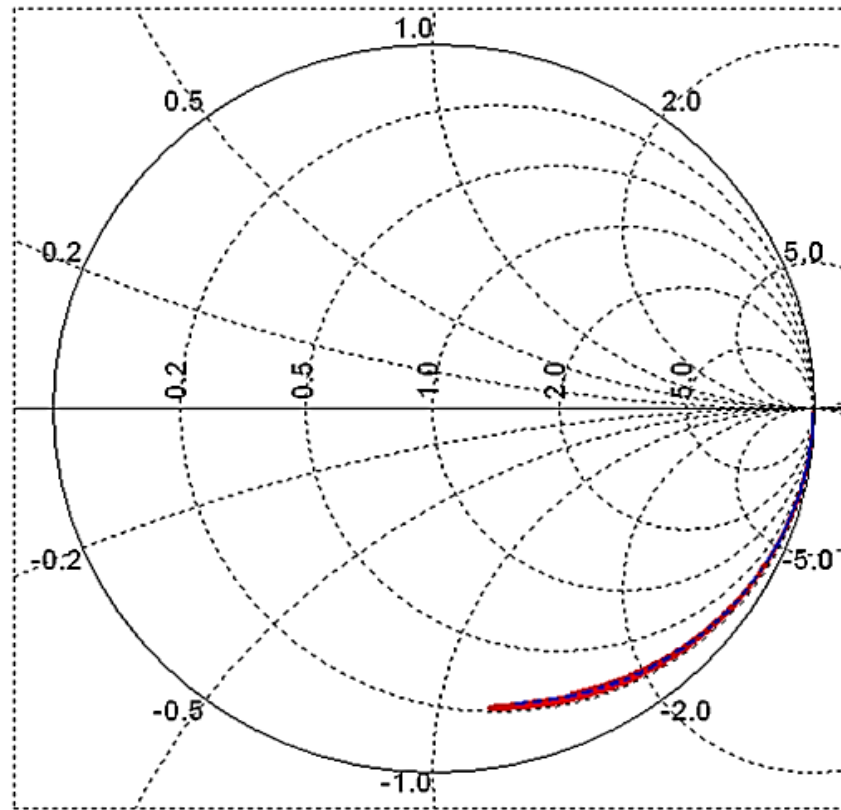


S_{21} Characterization of 128-fin n-MOSFET

S_{21} [dB]



S_{11} Characterization of 128-fin n-MOSFET



— Measurement
- - - Proposed Model

frequency



Simulation Speed Comparison of n-MOSFETs Ring Oscillators

# of Stages	17	35	71	143
BSIM6 simulation time [sec]	0.32	2.93	4.81	9.92
Proposed Model simulation time [sec]	0.34	2.96	4.91	10.86

- HSPICE was used for the simulations on Windows PC (Pentium i5)
- Each stage consists of a 128-fin n-MOSFET



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Summary

- SHE has been verified with a 2-D device simulator
- The proposed model was implemented into BSIM6 model with the Verilog-A language
- The proposed model has been verified with DC and small-signal S-parameter measurements
- The self-heat model can be applied to other MOSFET models



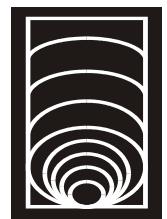
Future Research

- Since the Verilog-A code itself is not so fast for circuit simulation, the proposed model will be converted to a C code model for practical use
- Thermal capacitance measurement and the extraction will be developed for more gate fins of multi-finger MOSFETs
- A temperature-dependent method for circuit simulations will be considered



APPENDIX

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Verilog-A Source Implementations

- Time dependent heating implementation

```
Tdev = idt((Ids * Tem0 / Tdev * vds - (Tdev - Tem0)  
/ (RTH + (Tdev - Tem0)*KTH * RTH)) +Tem0;
```

- Small-signal AC simulations

```
if ((COSELFHEAT == 1)&&analysis("ac"))  
  begin  
    freq = 1.0 / (2.0 * `PI * $realtime());  
    cdrain = Ids / (1.0-2.0*`PI*freq*RTH*CTH);  
  end
```

