



International Conference on
Mechanical, Electrical and Medical Intelligent System

KL-08 Nov. 5, 2018 (Mon)

Performance Improvement of Analog / Mixed-Signal IC Using Signal Processing Techniques

Haruo Kobayashi

Gunma University





Contents

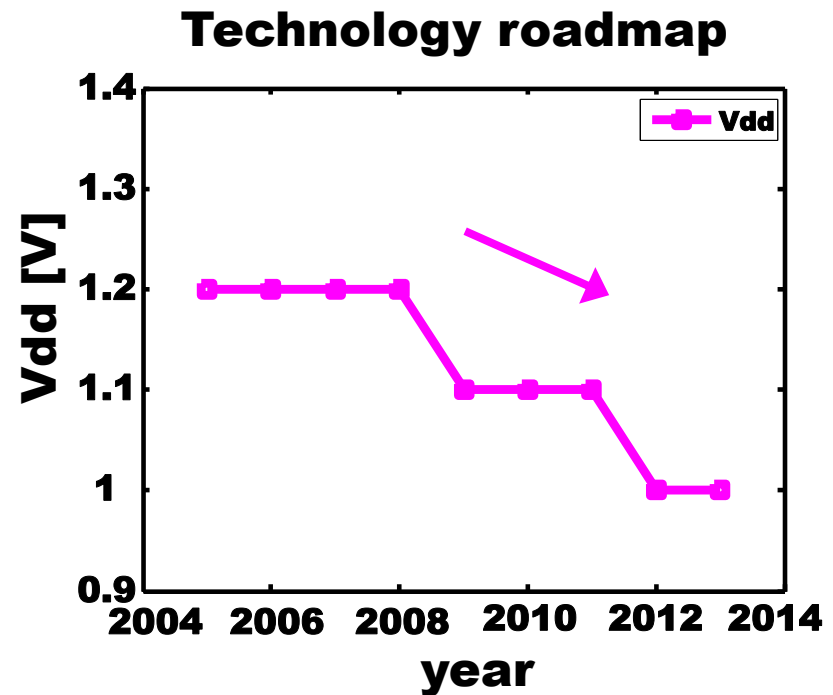
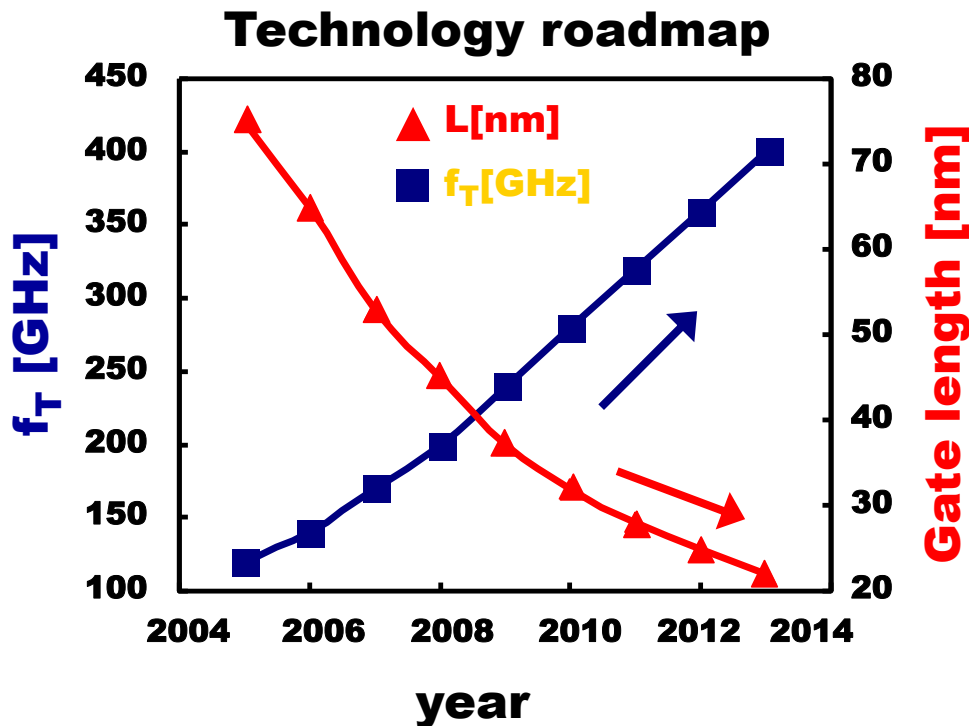
- Motivation
- Digitally Assisted Analog Technology
 - Region1 : Continuous Amplitude, Continuous Time
 - Region2 : Continuous Amplitude, Discrete Time
 - Region3 : Discrete Amplitude, Continuous Time
 - Region4 : Discrete Amplitude, Discrete Time
- Discussion
- Conclusion



Contents

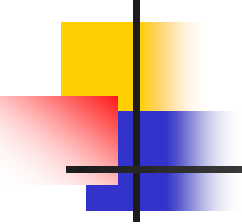
- Motivation
- Digitally Assisted Analog Technology
 - Region1 : Continuous Amplitude, Continuous Time
 - Region2 : Continuous Amplitude, Discrete Time
 - Region3 : Discrete Amplitude, Continuous Time
 - Region4 : Discrete Amplitude, Discrete Time
- Discussion
- Conclusion

Nano CMOS Era Has Come



□ Fine CMOS ⇒ High Speed (Time domain resolution)

Low breakdown (Vdd → low), Drain resistance → small



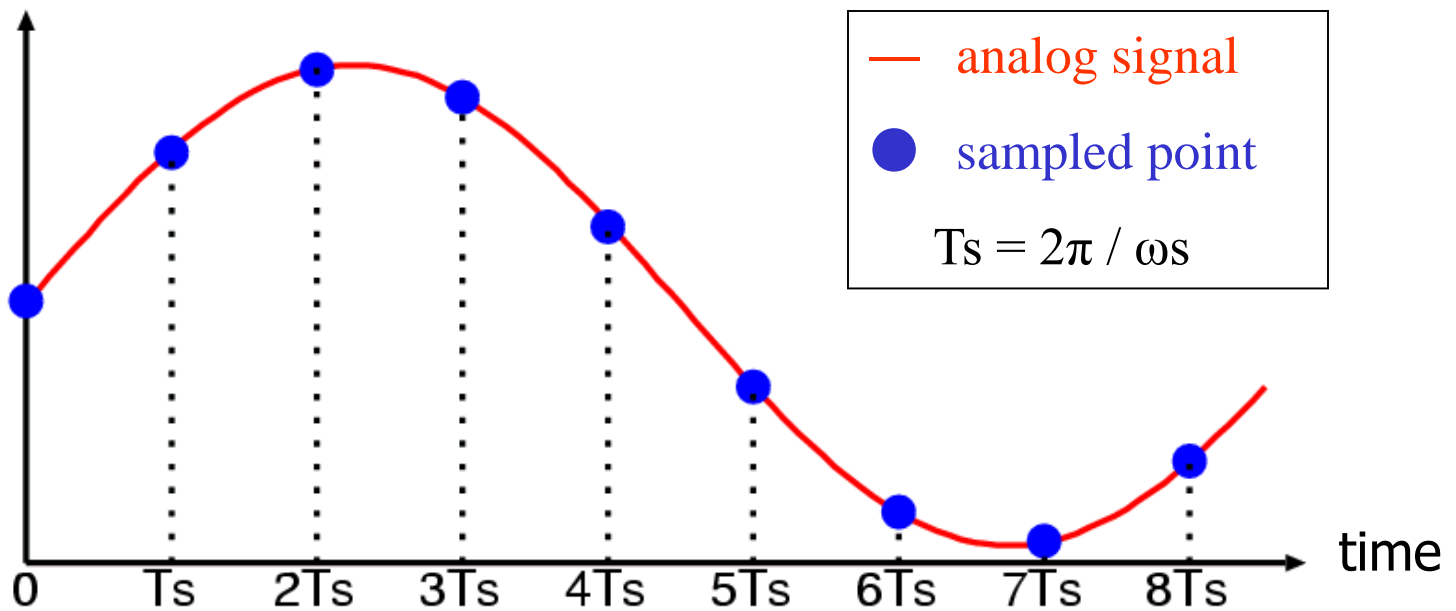
Position of Analog Technology in Modern Electronics

- Digital technology is great !
- Today,
analog technology is
for digital technology.

Cooperation between analog and digital
is important !

Digital Signal Feature (1)

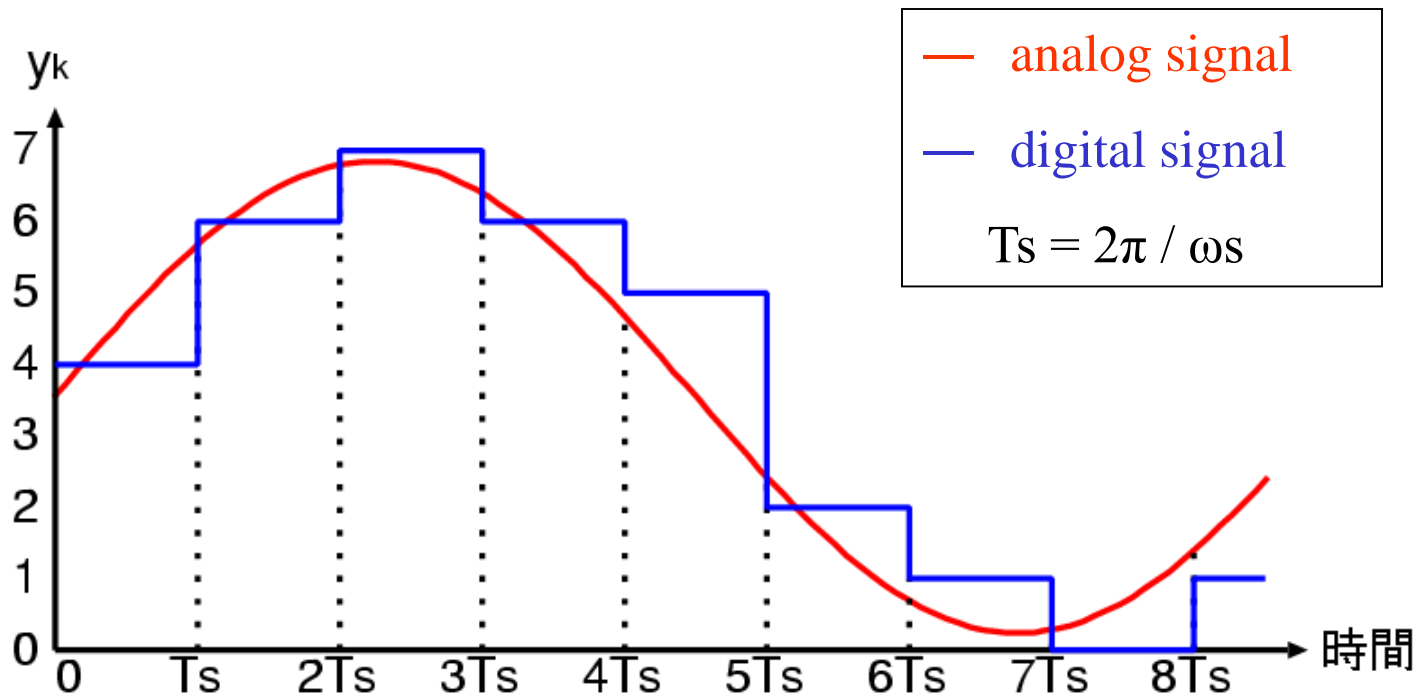
Discrete Time (Sampling)



Take sampled points and discard the others.

Digital Signal Feature (2)

Amplitude Quantization



Amplitude rounding

4 Regions for Circuit Design

	Continuous Time	Discrete Time
Continuous Amplitude	Region 1 Analog	Region 2 Switched Cap. Sampler
Discrete Amplitude	Region 3 TDC, PWM	Region 4 Digital

Region 1 : Good for bipolar

Region 2, 3, 4: Good for CMOS

Statement of this paper

Analog in nano CMOS era → Use all 4 regions



Contents

- Motivation
- **Digitally Assisted Analog Technology**
 - Region1 : Continuous Amplitude, Continuous Time
 - Region2 : Continuous Amplitude, Discrete Time
 - Region3 : Discrete Amplitude, Continuous Time
 - Region4 : Discrete Amplitude, Discrete Time
- Conclusion



Contents

● Motivation

● Digitally Assisted Analog Technology

Region1 : Continuous Amplitude, Continuous Time

Region2 : Continuous Amplitude, Discrete Time

Region3 : Discrete Amplitude, Continuous Time

Region4 : Discrete Amplitude, Discrete Time

● Conclusion

[1] M. Hirano, N. Kushita, Y. Moroshima, H. Harakawa, T. Oikawa, N. Tsukiji, T. Ida, Y. Shibasaki, H. Kobayashi, "Silicon Verification of Improved Nagata Current Mirrors",

IEEE International Conference on Solid-State and Integrated Circuit Technology (Nov. 2018)

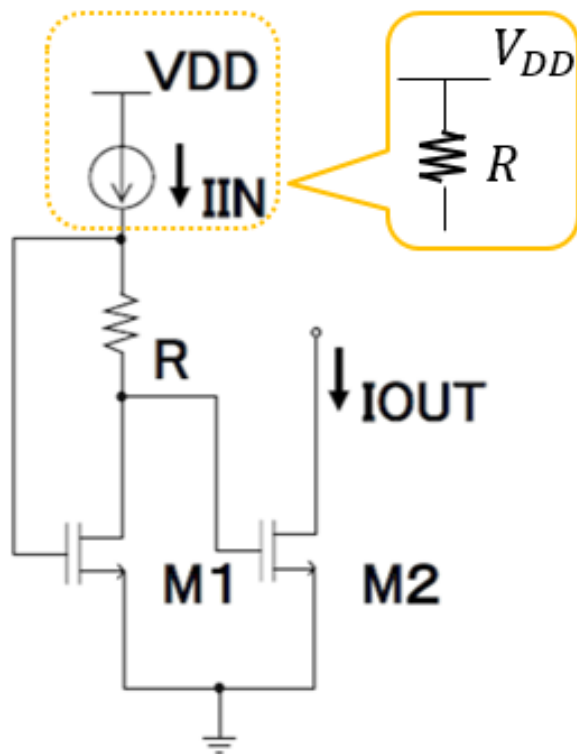
[2] M. Hirano, N. Tsukiji, Haruo Kobayashi,

"Simple Reference Current Source Insensitive to Power Supply Voltage Variation

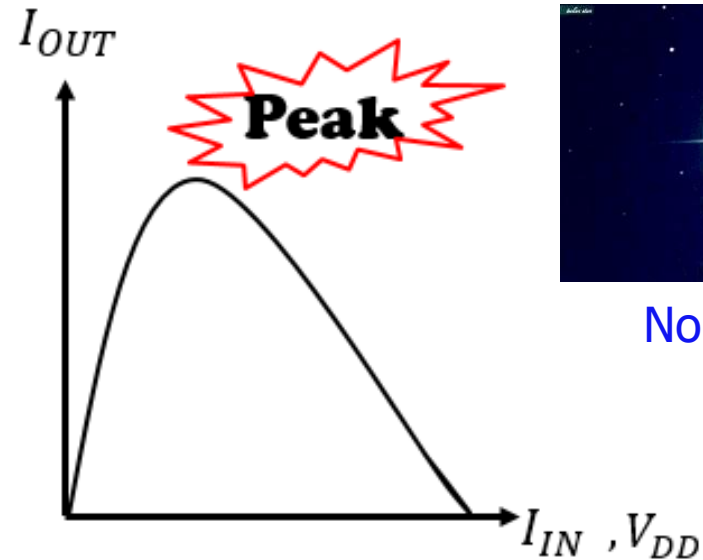
- Improved Minoru Nagata Current Source",

IEEE International Conference on Solid-State and Integrated Circuit Technology (Oct., 2016)

Original Nagata Current Mirror



MOS Nagata
Current Mirror Circuit



North Star

Peaking current characteristics

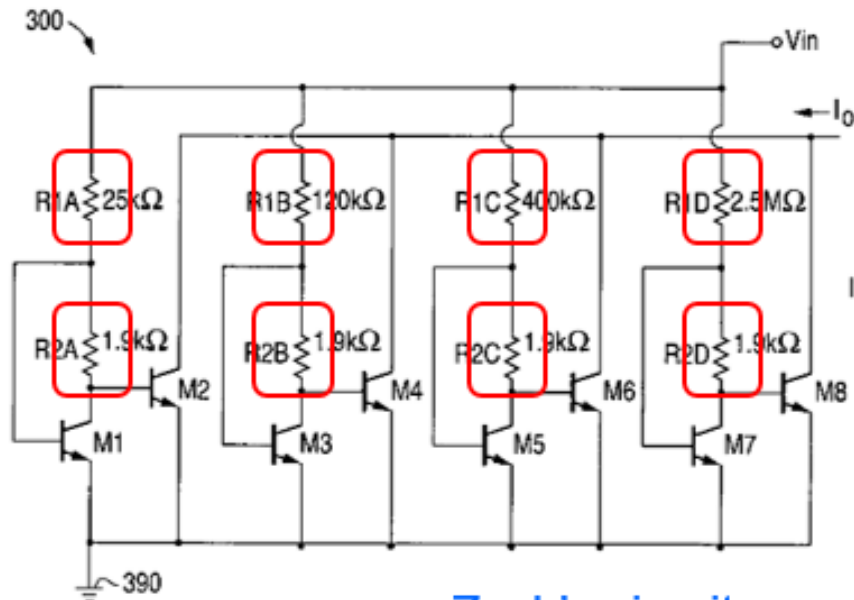
At peak vicinity



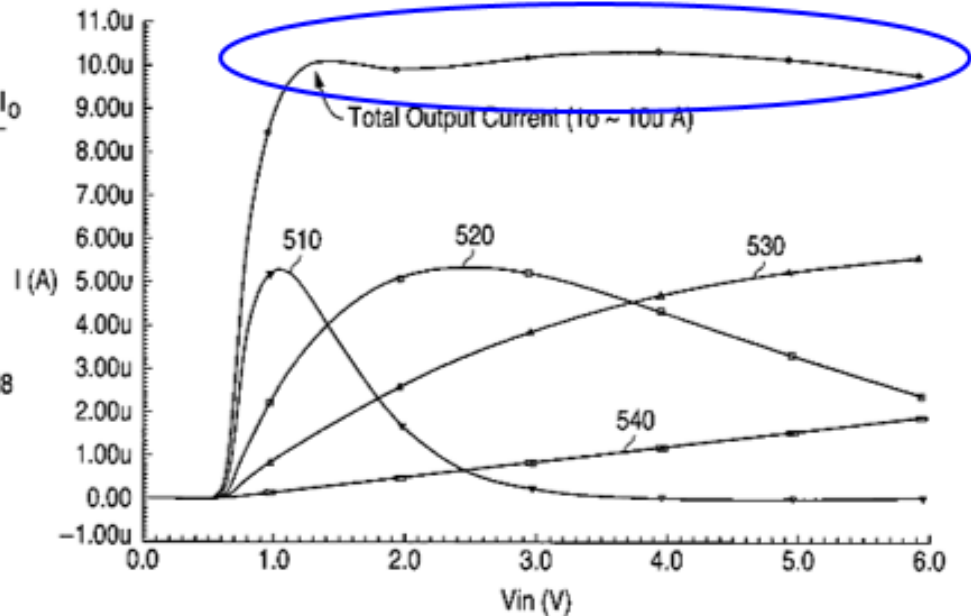
Small output current change
against input current change

Simple  Widely used. Ex: in DC-DC converter ICs

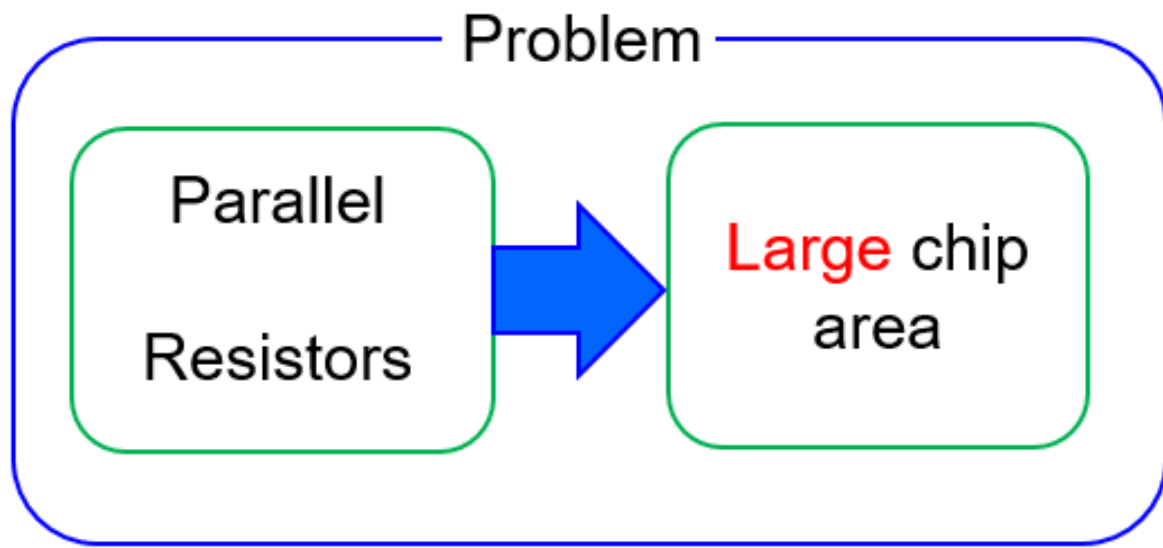
Previous Improved Circuit



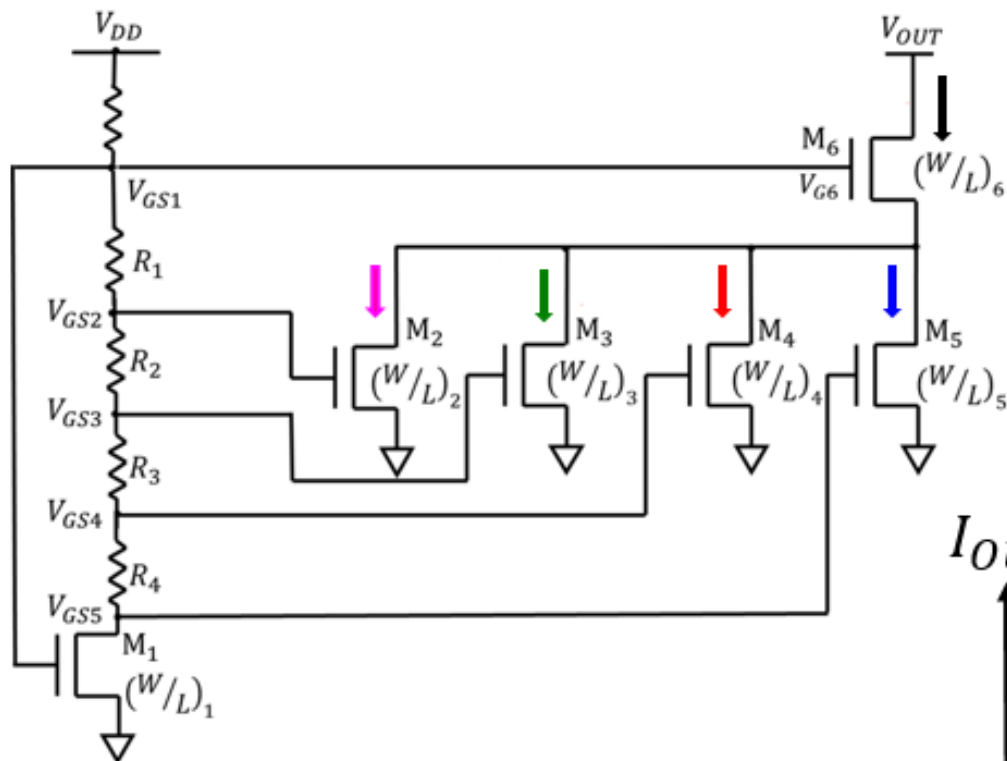
Zach's circuit



Inventor
 Zachary Zehner Nosker
 Obtained Ph.D.
 from Gunma Univ.
 Kobayashi Lab.



MOS Reference Current Source Operation



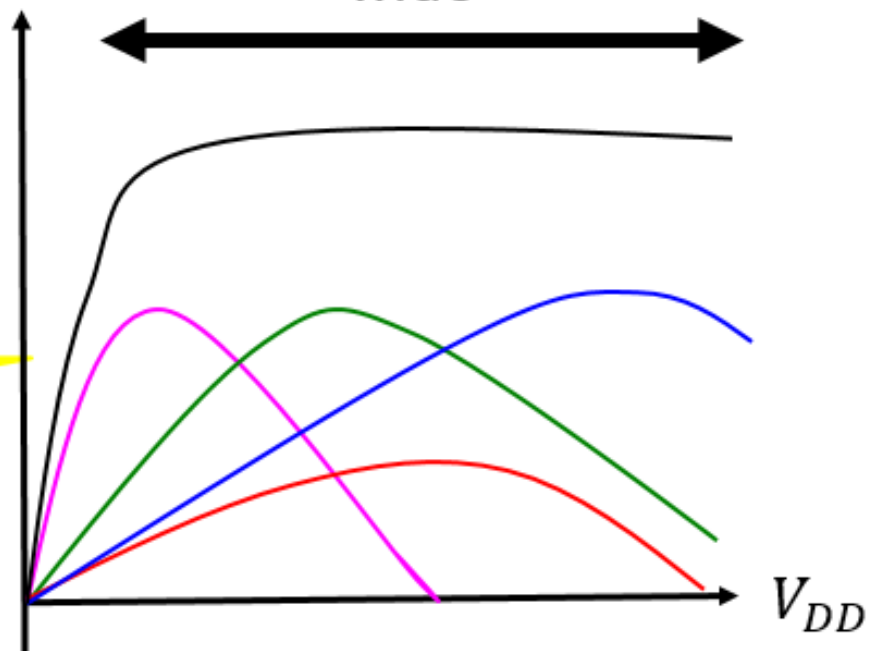
Almost **constant**
total current

I_{OUT}

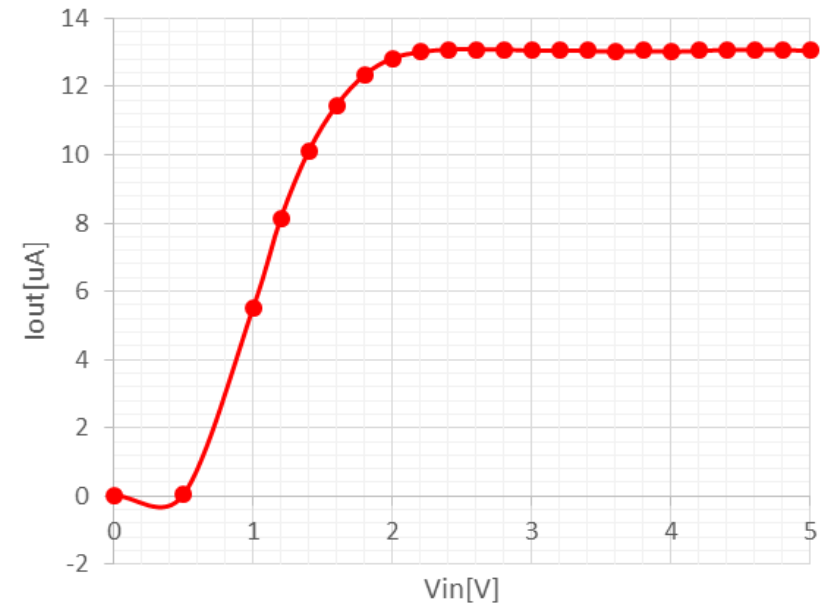
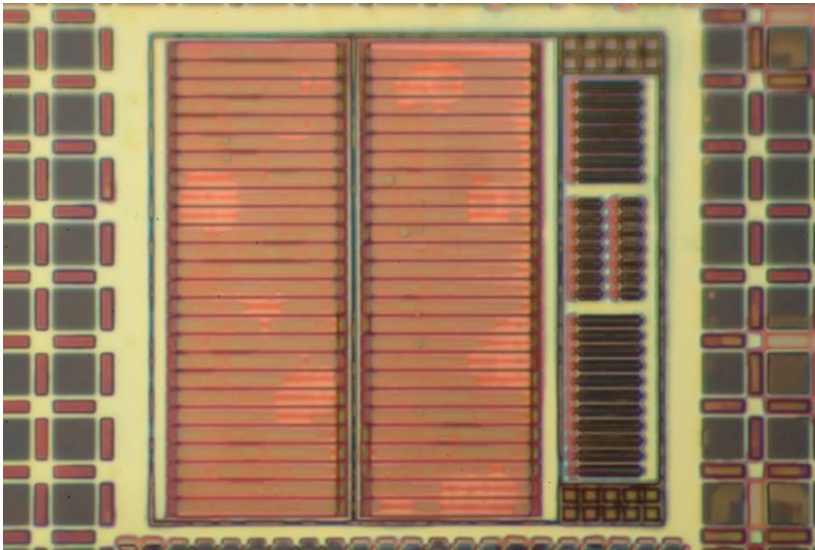
wide



Simple
Reference Current Source



Chip Fabrication / Measurement



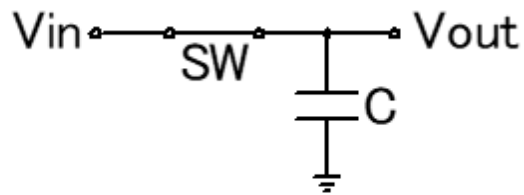
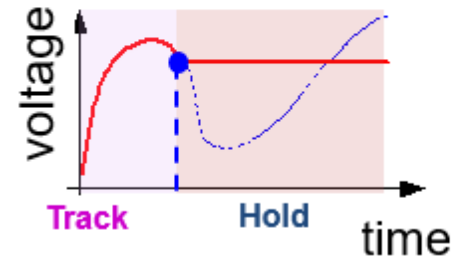
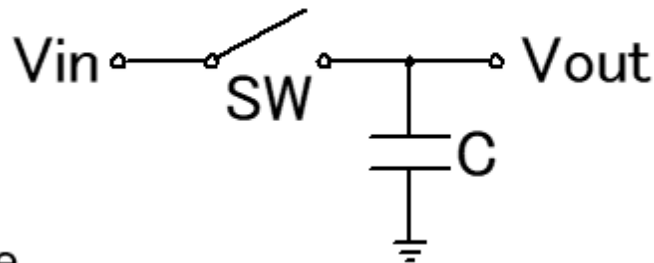
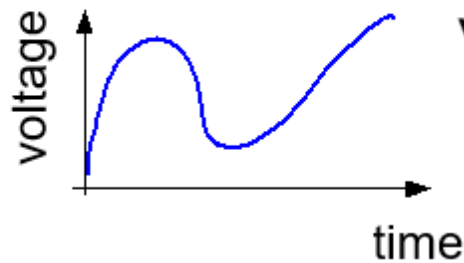
Supported by ASO Inc.



Contents

- Motivation
- **Digitally Assisted Analog Technology**
 - Region1 : Continuous Amplitude, Continuous Time
 - Region2 : Continuous Amplitude, Discrete Time**
 - Region3 : Discrete Amplitude, Continuous Time
 - Region4 : Discrete Amplitude, Discrete Time
- Conclusion

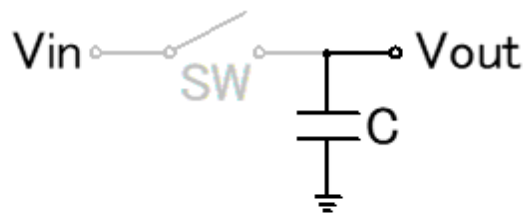
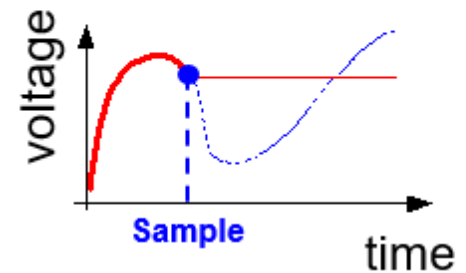
Sampling Circuit



SW ON

- $V_{out}(t) = V_{in}(t)$

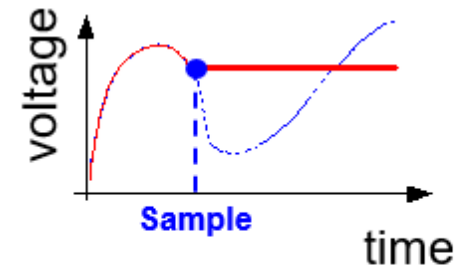
Track mode



SW OFF

- $V_{out}(t) = V_{in}(t_{OFF})$

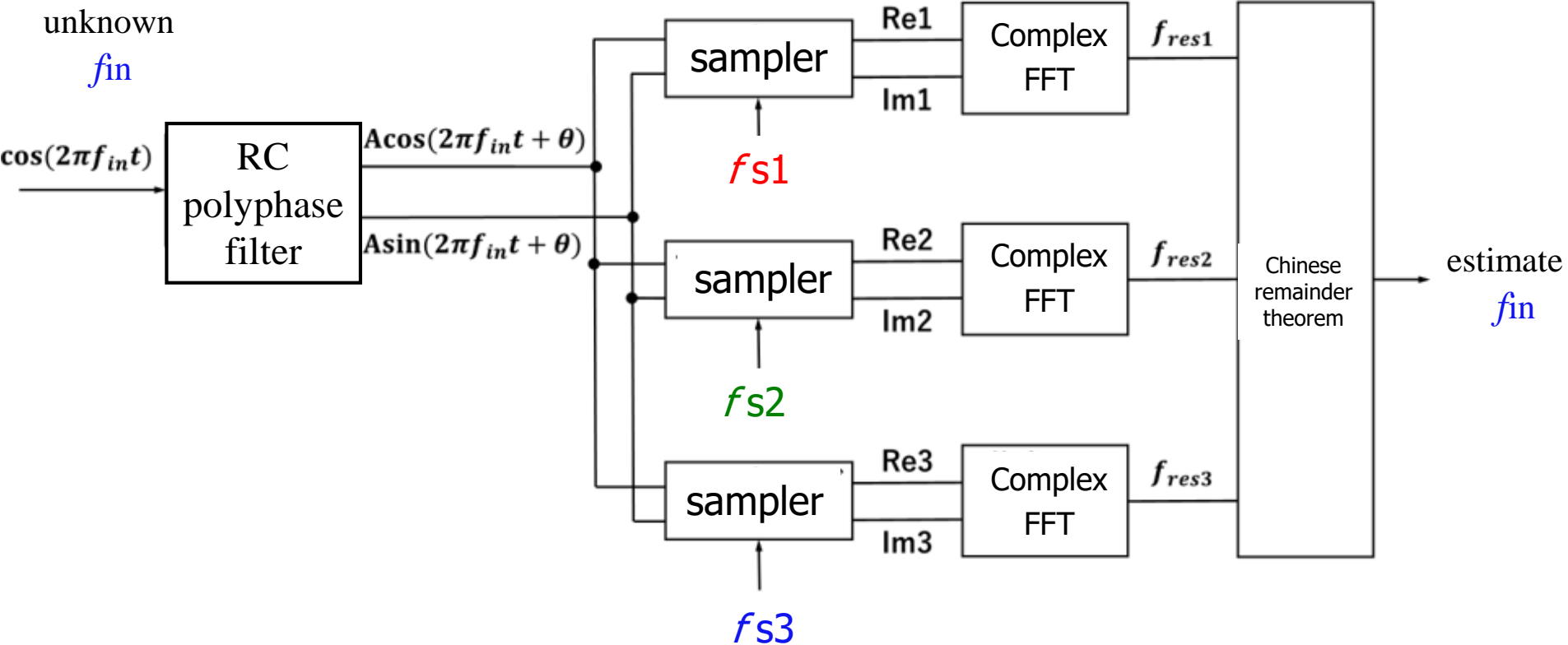
Hold mode



Frequency Estimation Sampler System

Giga Hertz order

Based on residue number system



kHz order sampling frequencies

f_{s1} , f_{s2} , f_{s3} are prime numbers.

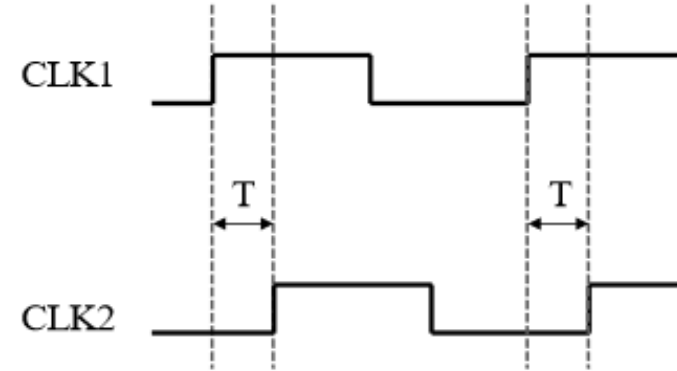


Contents

- Motivation
- **Digitally Assisted Analog Technology**
 - Region1 : Continuous Amplitude, Continuous Time
 - Region2 : Continuous Amplitude, Discrete Time
 - Region3: Discrete Amplitude, Continuous Time**
 - Region4: Discrete Amplitude, Discrete Time
- Conclusion

Time-to-Digital Converter (TDC)

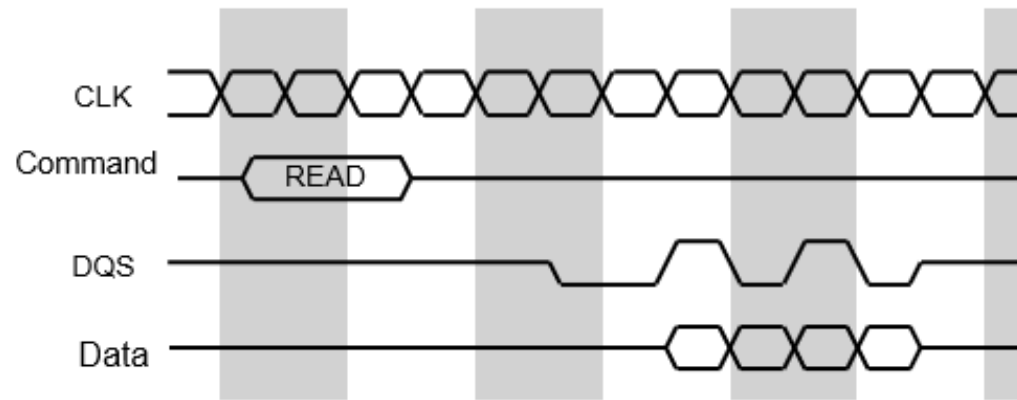
- Testing the timing between two repetitive digital signals
Ex. Data and clock in Double Data Rate memory in Double Data Rate memory



- Short testing time
- Good accuracy



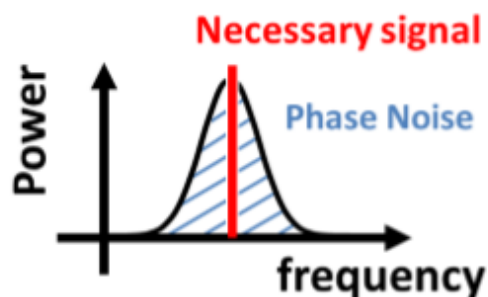
Implement BOST with small circuitry



BOST: Built-Out Self-Test

Phase Noise Test with $\Delta\Sigma$ TDC

Phase noise in oscillator



Malfunction of electronic systems

- RF circuit & system
- ADC

Important

Phase noise test **at low cost**, in short time

Conventional



- **Expensive** : Spectrum analyzer
- **Long** : test time (~10seconds)



cost

high



Proposed

Simple circuit



$\Delta\Sigma$ Time-to-Digital Converter

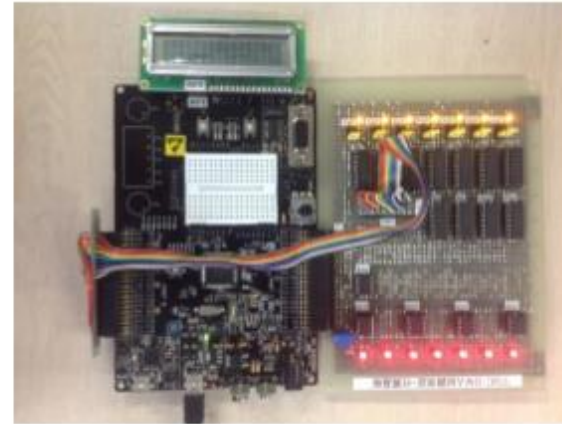
Low cost, high quality
Phase noise test

[1] Y. Osawa, H. Kobayashi, "Phase Noise Measurement Techniques Using Delta-Sigma TDC", IEEE International Mixed-Signals, Sensors and Systems Test Workshop (IMS3TW'14), Porto Alegre, Brazil (Sept. 17-19, 2014).

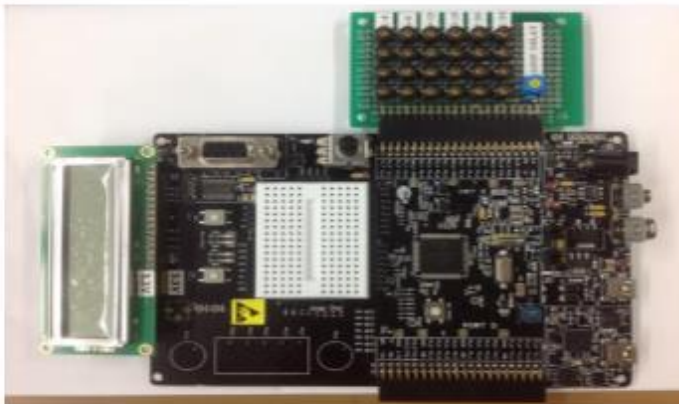
TDC BOSTs for Timing Signal Testing



Single-bit $\Delta\Sigma$ TDC with analog FPGA



Multi-bit $\Delta\Sigma$ TDC with analog FPGA



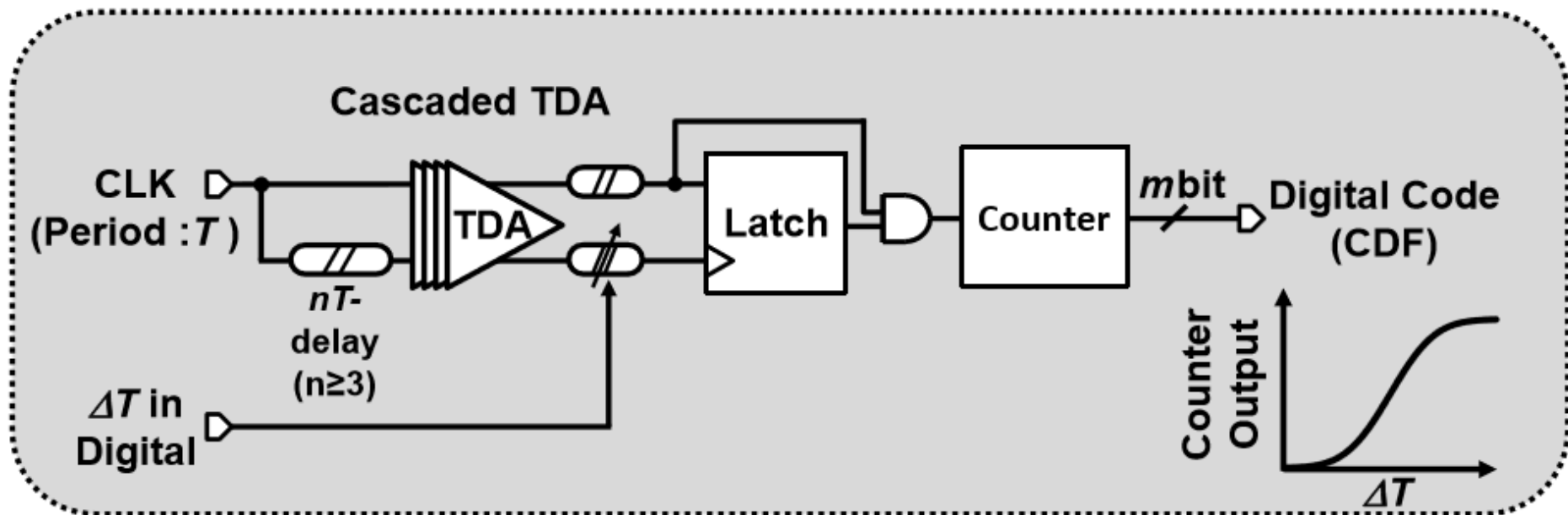
Flash-type TDC with analog FPGA



Flash-type TDC with digital FPGA

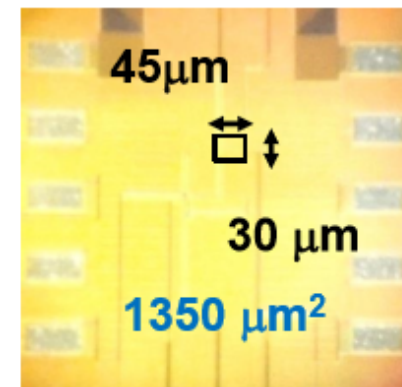
[2] T. Chujo, H. Kobayashi, "Experimental Verification of Timing Measurement Circuit With Self-Calibration", IEEE International Mixed-Signals, Sensors and Systems Test Workshop (IMS3TW'14), Porto Alegre, Brazil (Sept. 17-19, 2014).

On-chip Jitter Measurement Circuit



Experiments show that
1.67 ps RMS timing jitter
can be measured

Process : 65 nm CMOS
Supply Voltage : 1.2 V



[3] K. Niitsu, H. Kobayashi, "CMOS Circuits to Measure Timing Jitter Using a Self-Referenced Clock and a Cascaded Time Difference Amplifier with Duty-Cycle Compensation," IEEE Journal of Solid-State Circuits, Nov. 2012.



Contents

- Motivation
- **Digitally Assisted Analog Technology**
 - Region1 : Continuous Amplitude, Continuous Time
 - Region2 : Continuous Amplitude, Discrete Time
 - Region3 : Discrete Amplitude, Continuous Time
 - Region4 : Discrete Amplitude, Discrete Time**
- Conclusion

[1] Ru YI, Minghui WU, Koji ASAMI, Haruo KOBAYASHI, Ramin KHATAMI, Atsuhiko KATAYAMA, Isao SHIMIZU, Kentaroh KATOH
“Digital Compensation for Timing Mismatches in Interleaved ADCs”,
IEEE 22nd Asian Test Symposium, Yilan, Taiwan, (Nov. 2013).

Research Background & Objective



ATE System

■ Background

High-speed sampling time-interleaved ADC for ATE system

Timing skew → Big issue

Error compensation of timing skew effects

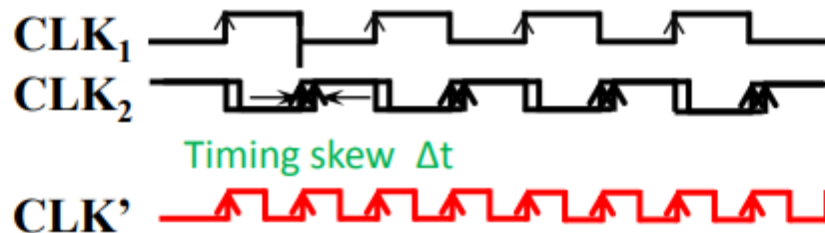
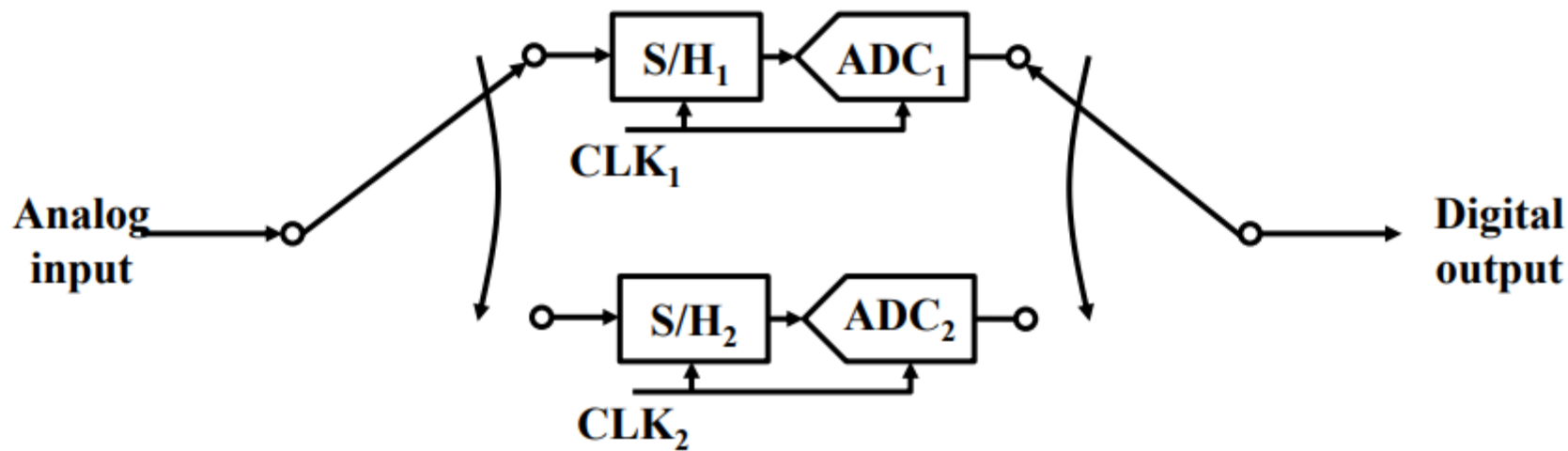
Conventional
Analog method + Digital method

■ Objective

Proposal
Full digital method

High accuracy, Stable, Reliable

2-channel Interleaved ADC



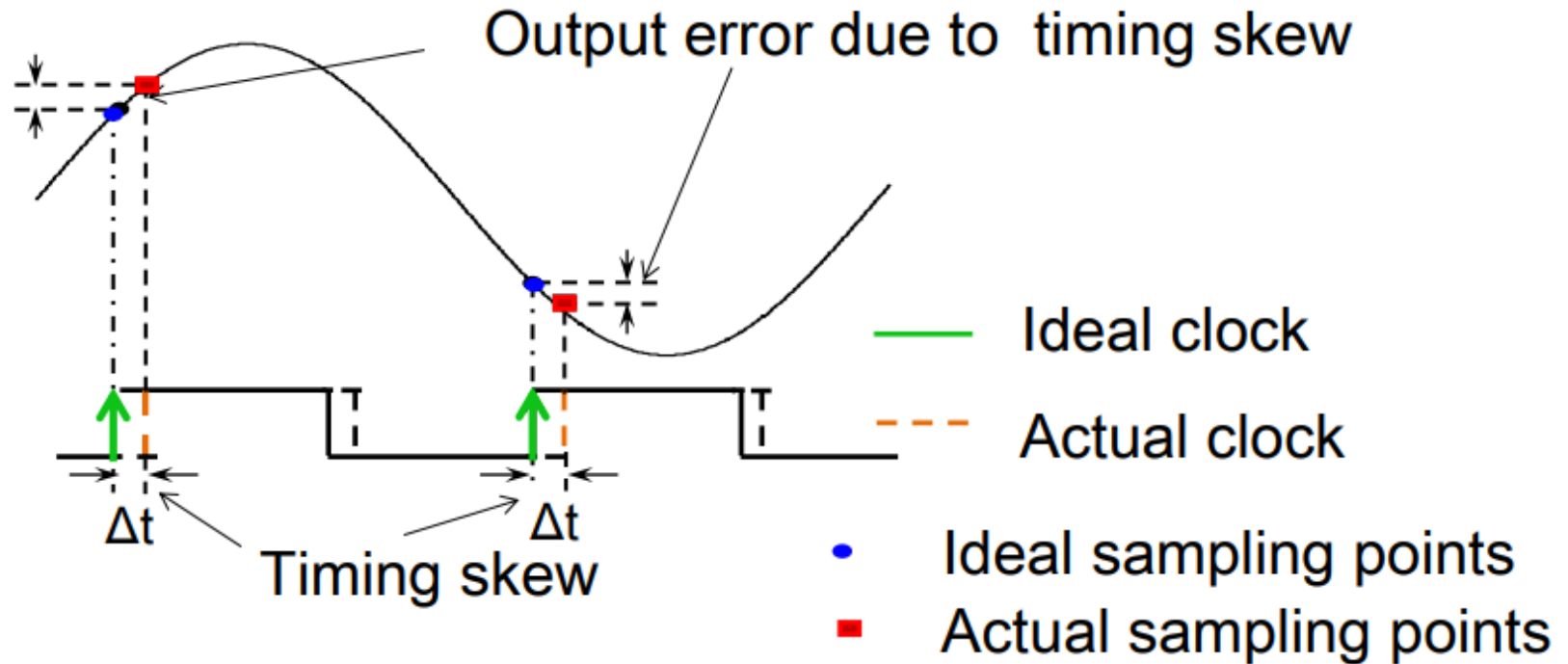
Time →

CLK1 reference

CLK2 delayed by half period

2 times sampling rate

Timing Error in Sampling

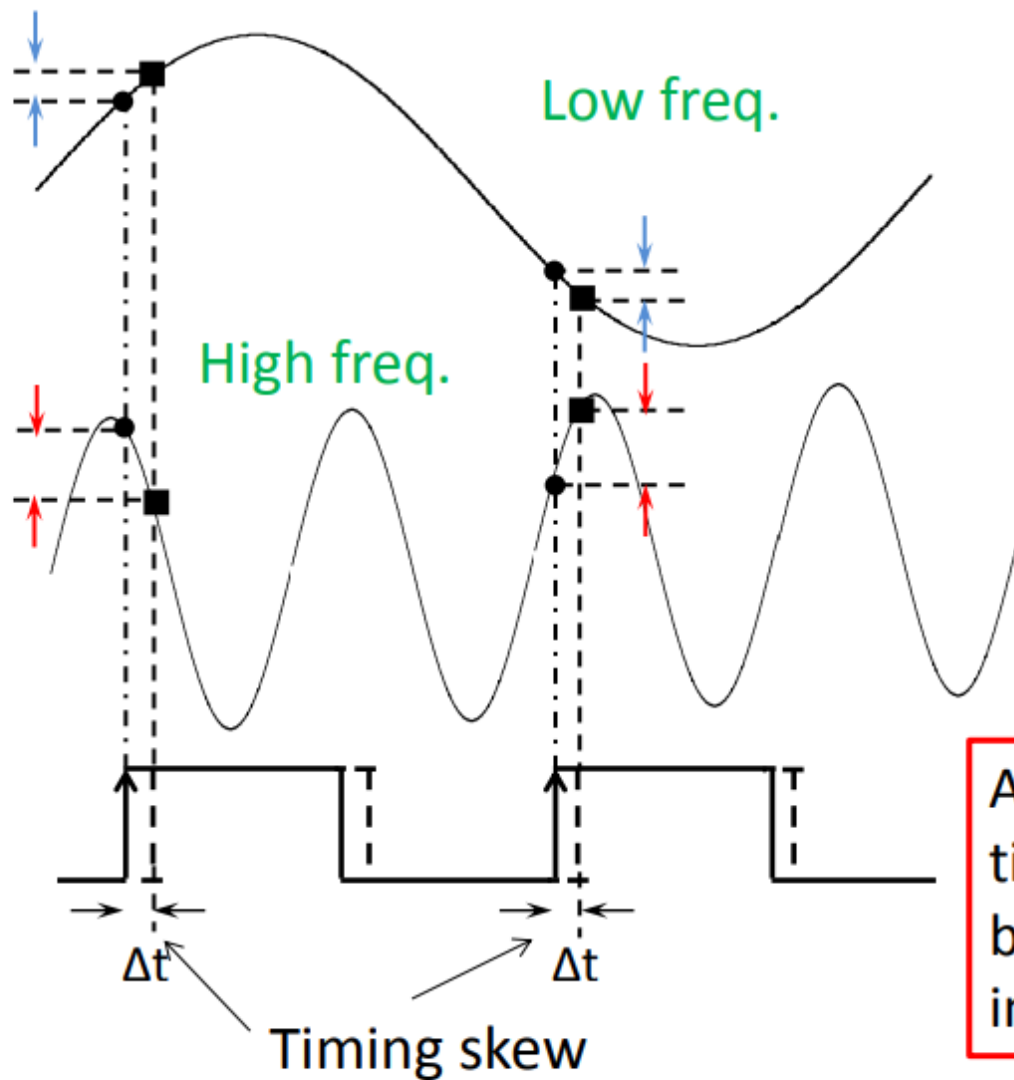


Timing error (horizontal error)



Sampled voltage error (vertical error)

Input Frequency & Output




Input freq. \uparrow

\Downarrow

Output error \uparrow

As input frequency increases, timing skew problem becomes serious in interleaved ADC.

Proposed Calibration System

- 
- Full digital
 - Timing Skew Detection
 - Cross-correlation of two channel ADC outputs
 - Timing Skew Effect Compensation
 - Delay linear digital filter
 - Calibration Control
 - Successive approximation algorithm
 - Foreground calibration

Correlation of $R(0)$ and $R(1)$

CH1 ADC output: $f[n]$

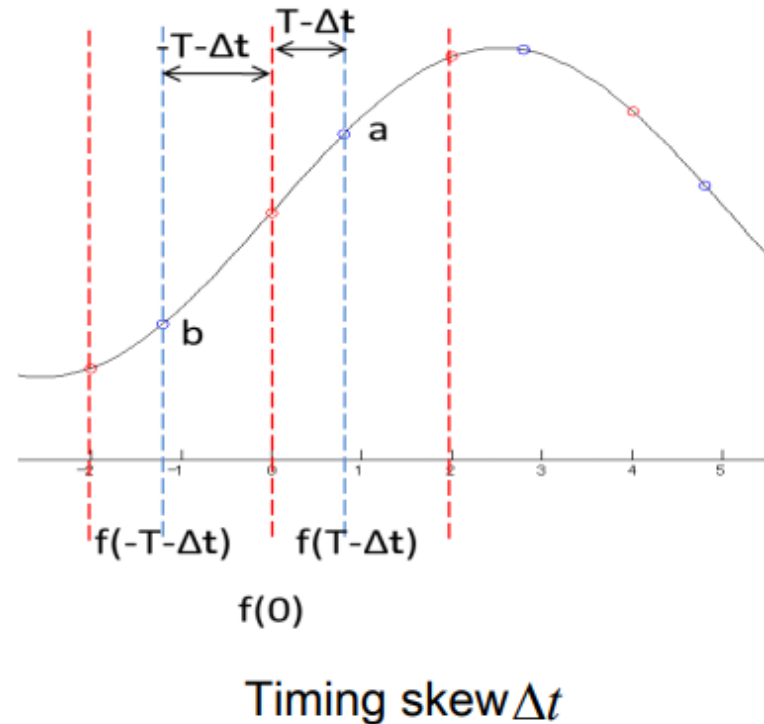
CH2 ADC output: $g[n] = f[n + T - \Delta t]$

lag 0,

$$R(0) = R_{ff}[0] = \lim_{n \rightarrow \infty} \frac{1}{n} \sum_{-n}^n f[n] f[n + T - \Delta t]$$

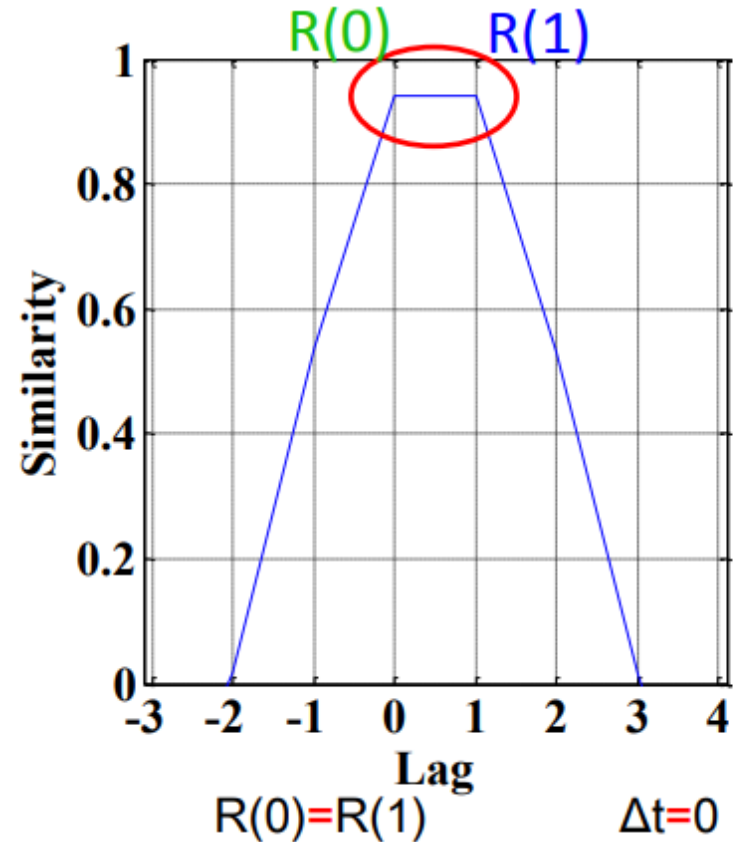
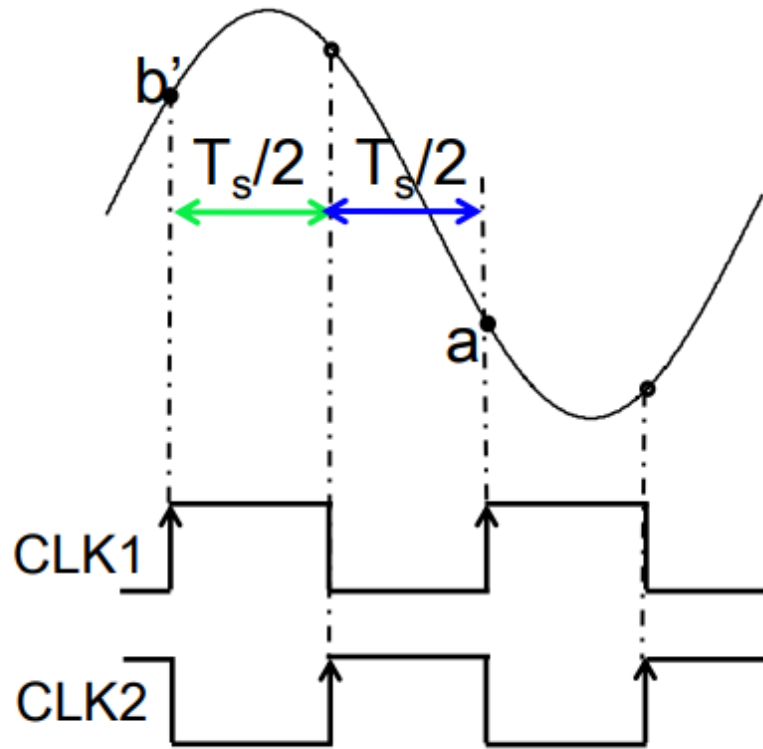
lag 1,

$$R(1) = R_{ff}[-2T] = \lim_{n \rightarrow \infty} \frac{1}{n} \sum_{-n}^n f[n] f[n - T - \Delta t]$$

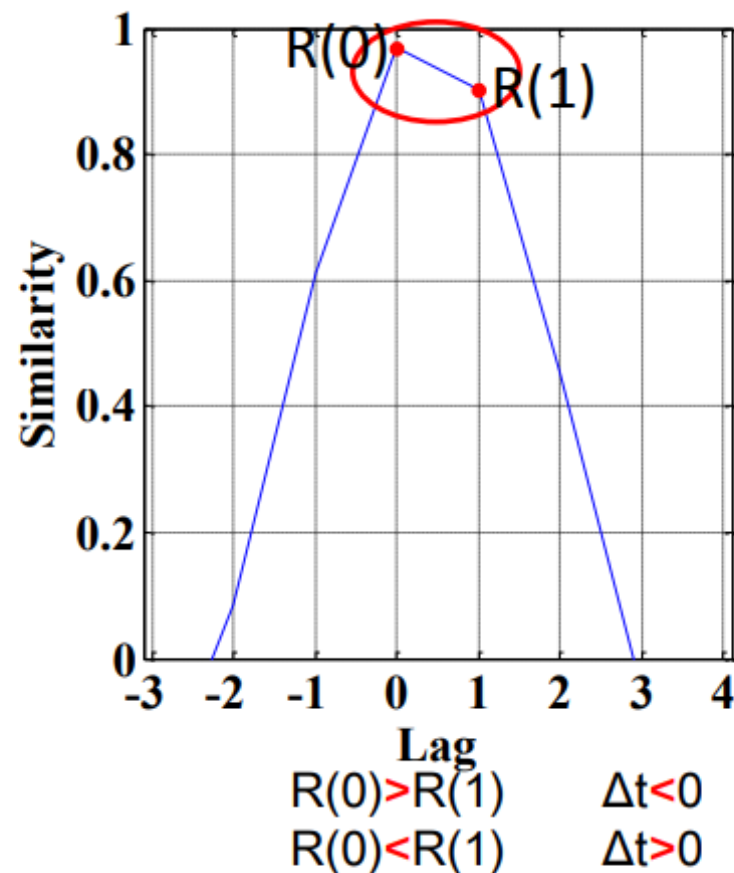
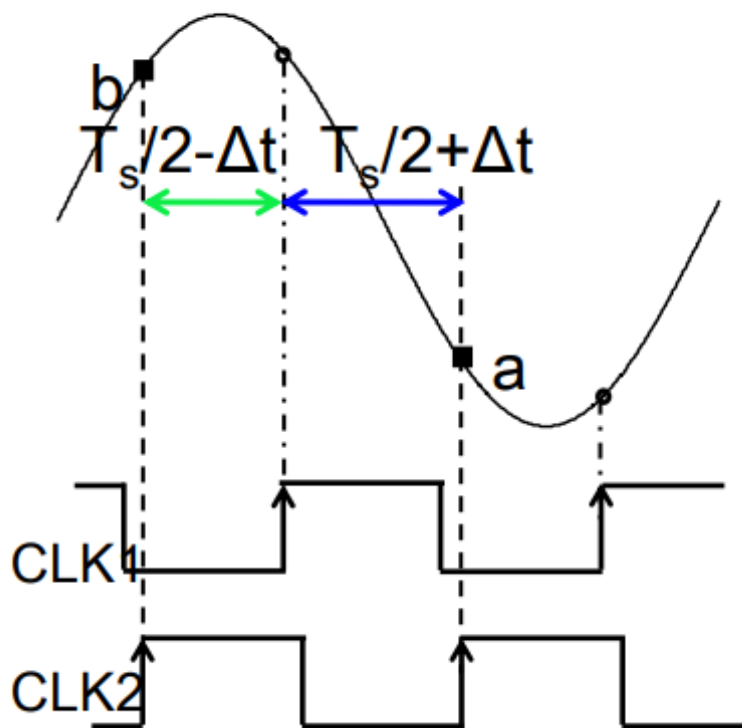


Timing Skew Detection

Cross-Correlation without Timing Skew



Cross-Correlation with Timing Skew



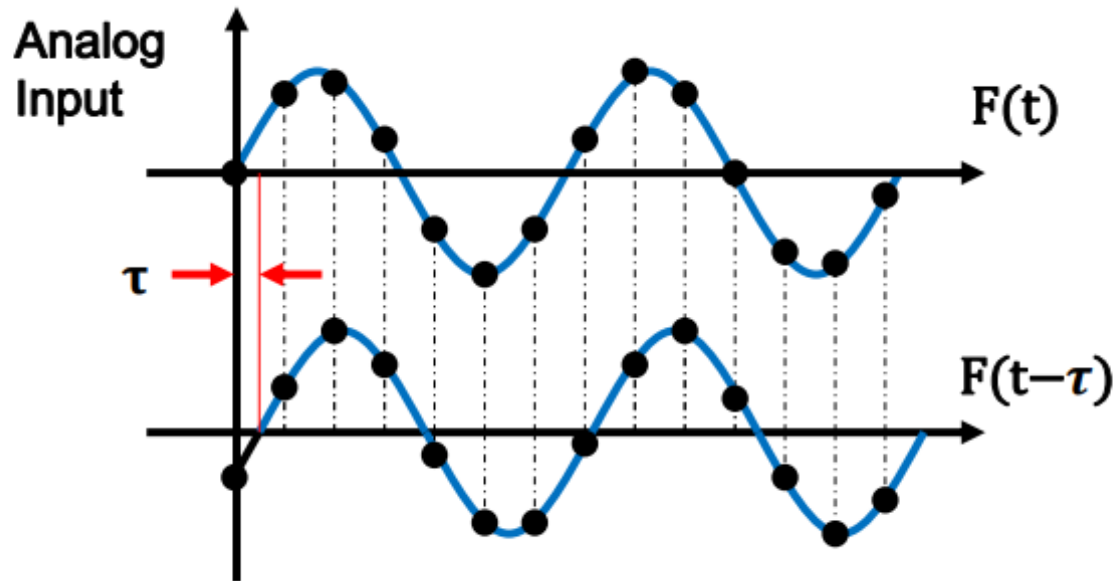
cross-correlation value



Sign of Δt ($\Delta t > 0$ or $\Delta t < 0$)
 Magnitude of $|\Delta t|$



Linear Phase Delay Digital Filter



Linear phase \longrightarrow maintain time-domain waveform

Conventional Linear-Phase Digital Filter :

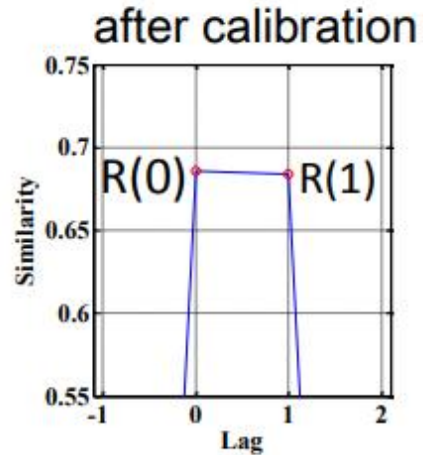
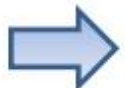
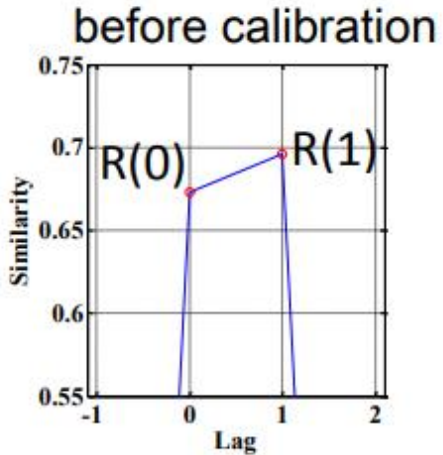
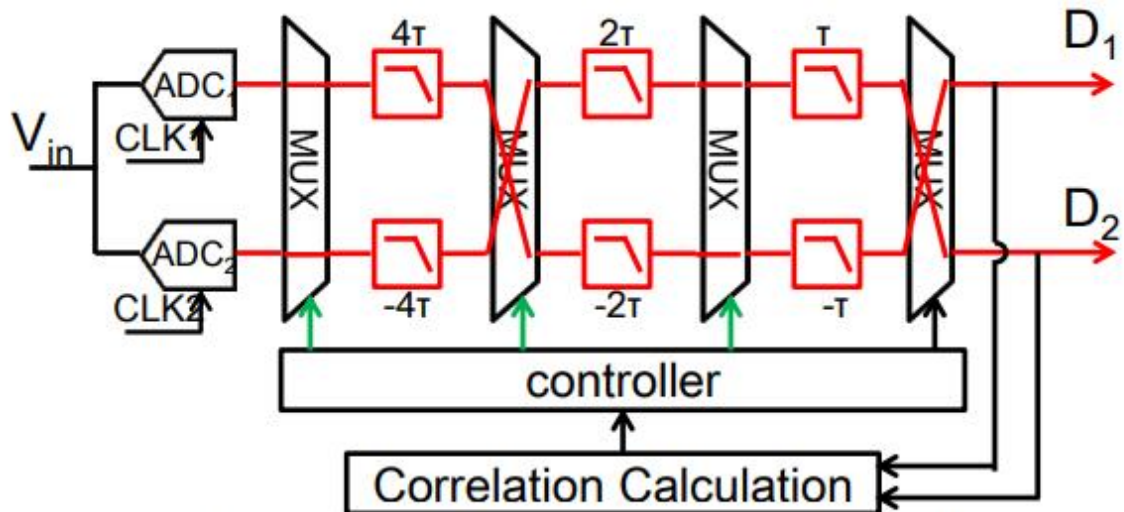
Group delay time resolution $T_s/2$

Proposed Linear-Phase Delay Digital Filter [1] :

Arbitrary small time resolution τ

[1] K. Asami, et. al., "Timing Skew Compensation Technique using Digital Filter with Novel Linear Phase Condition," IEEE International Test Conference (Nov. 2010).

Calibration Done



Binary-search, successive approximation



Contents

- Motivation
- Digitally Assisted Analog Technology
 - Region1 : Continuous Amplitude, Continuous Time
 - Region2 : Continuous Amplitude, Discrete Time
 - Region3 : Discrete Amplitude, Continuous Time
 - Region4 : Discrete Amplitude, Discrete Time
- Conclusion



Conclusion

In nano CMOS era,
analog/mixed signal IC design should use
all of 4 circuit design regions.

Analog & digital signal processing



Good performance, low power,
low cost, fast turn around time
can be achieved.