

Operational Amplifier Stability Research

Research Objective

Stability Criteria

Our proposal

For Analysis and design of operational amplifier stability

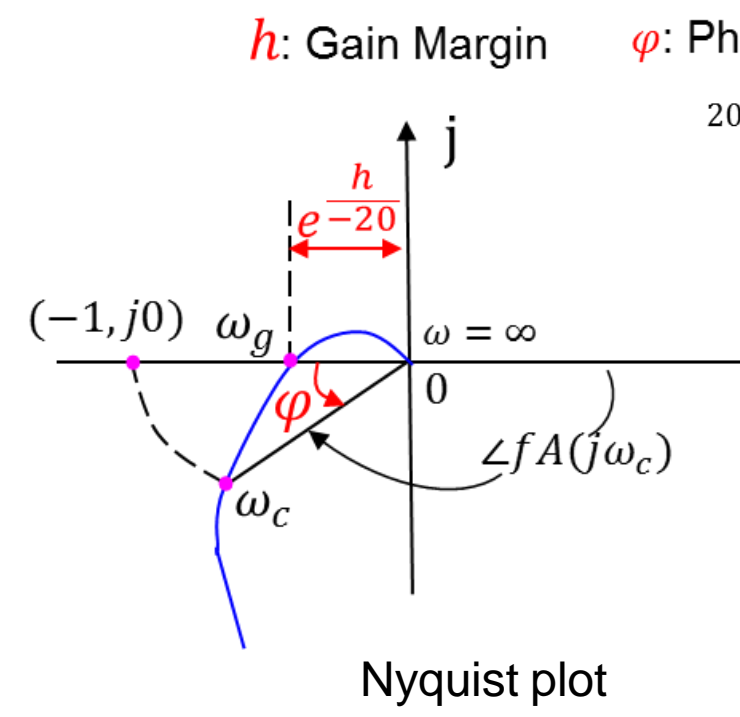
Use Routh-Hurwitz stability criterion

We can obtain Explicit stability condition for circuit parameters (which can NOT be obtained only with Bode plot).

Relationship between R-H parameters and phase margin

- Electronic Circuit Design Field
 - Bode plot (>90% frequently used)
 - Nyquist plot
 - Control Theory Field
 - Bode plot
 - Nyquist plot
 - Nicholas plot
 - Routh-Hurwitz stability criterion
- Very popular in control theory field but rarely seen in electronic circuit books/papers

Nyquist stability Criteria



Signal Expression

Time domain $A(t)$ $\xrightarrow{\text{Laplace transform}}$ Frequency domain $A(j\omega)$ $A(s)$

Routh-Hurwitz stability Criteria

$$\frac{A(s)}{1+A(s)} = \frac{N(s)}{D(s)} = \frac{b_n s^m + b_{m-1} s^{m-1} + \dots + b_1 s + b_0}{\alpha_n s^n + \alpha_{n-1} s^{n-1} + \dots + \alpha_1 s + \alpha_0}$$

Routh table

S^n	α_n	α_{n-2}	α_{n-4}	α_{n-6}	...
S^{n-1}	α_{n-1}	α_{n-3}	α_{n-5}	α_{n-7}	...
S^{n-2}	$\beta_1 = \frac{\alpha_{n-1}\alpha_{n-2} - \alpha_n\alpha_{n-3}}{\alpha_{n-1}}$	$\beta_2 = \frac{\alpha_{n-1}\alpha_{n-4} - \alpha_n\alpha_{n-5}}{\alpha_{n-1}}$	β_3	β_4	...
S^{n-3}	$\gamma_1 = \frac{\beta_1\alpha_{n-3} - \alpha_{n-1}\beta_2}{\beta_1}$	$\gamma_2 = \frac{\beta_1\alpha_{n-5} - \alpha_{n-1}\beta_3}{\beta_1}$	γ_3	γ_4	...
...
S^0	α_0				

Bode plot does NOT show explicit stability conditions of circuit parameters.

Equivalence at Mathematical Foundations

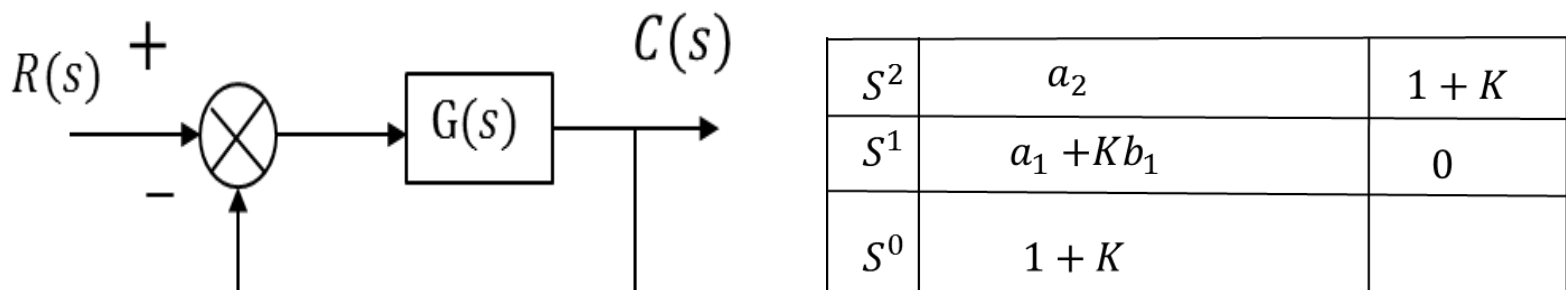
Two Examples

Ex.1 $G(s) = \frac{K(1+b_1s)}{1+a_1s+a_2s^2}$ One Zero, Two Poles

Ex.2 $G(s) = \frac{K(1+b_1s)}{1+a_1s+a_2s^2+a_3s^3}$ One Zero, Three Poles

Stability Condition

Based on Routh-Hurwitz Criterion

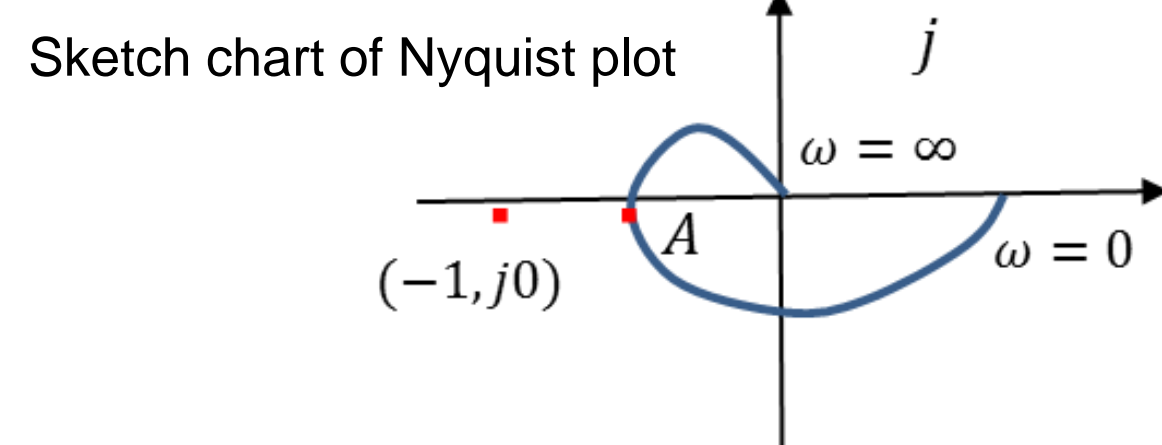


S^2	a_2	$1+K$
S^1	a_1+Kb_1	0
S^0	$1+K$	

Ex.1 $K > -\frac{a_1}{b_1}$ At condition: $b_1 > 0$
 $K < -\frac{a_1}{b_1}$ At condition: $b_1 < 0$

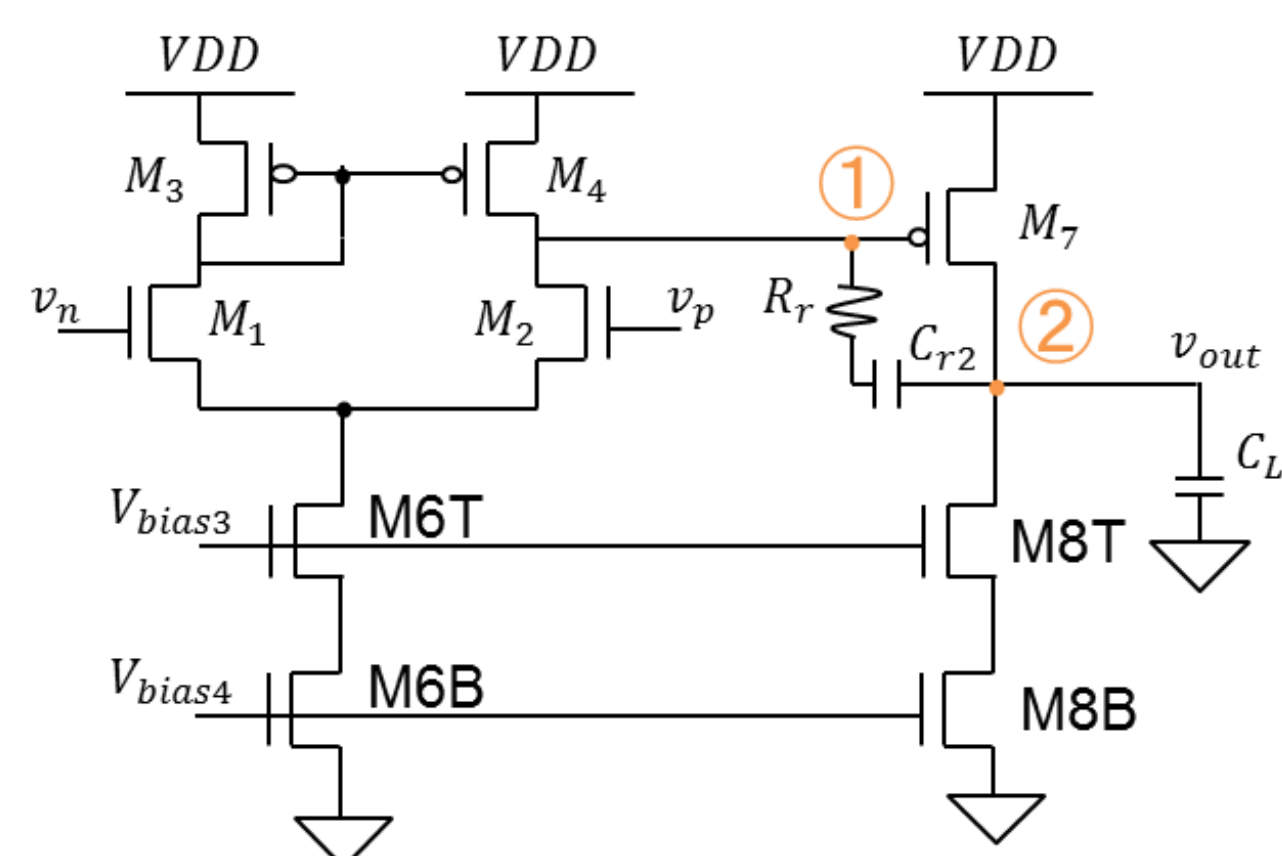
Ex.2 $K > \frac{a_3 - a_1 a_2}{a_2 b - a_3}$ At condition: $a_2 b - a_3 > 0$
 $K < \frac{a_3 - a_1 a_2}{a_2 b - a_3}$ At condition: $a_2 b - a_3 < 0$

Based on Nyquist Criterion

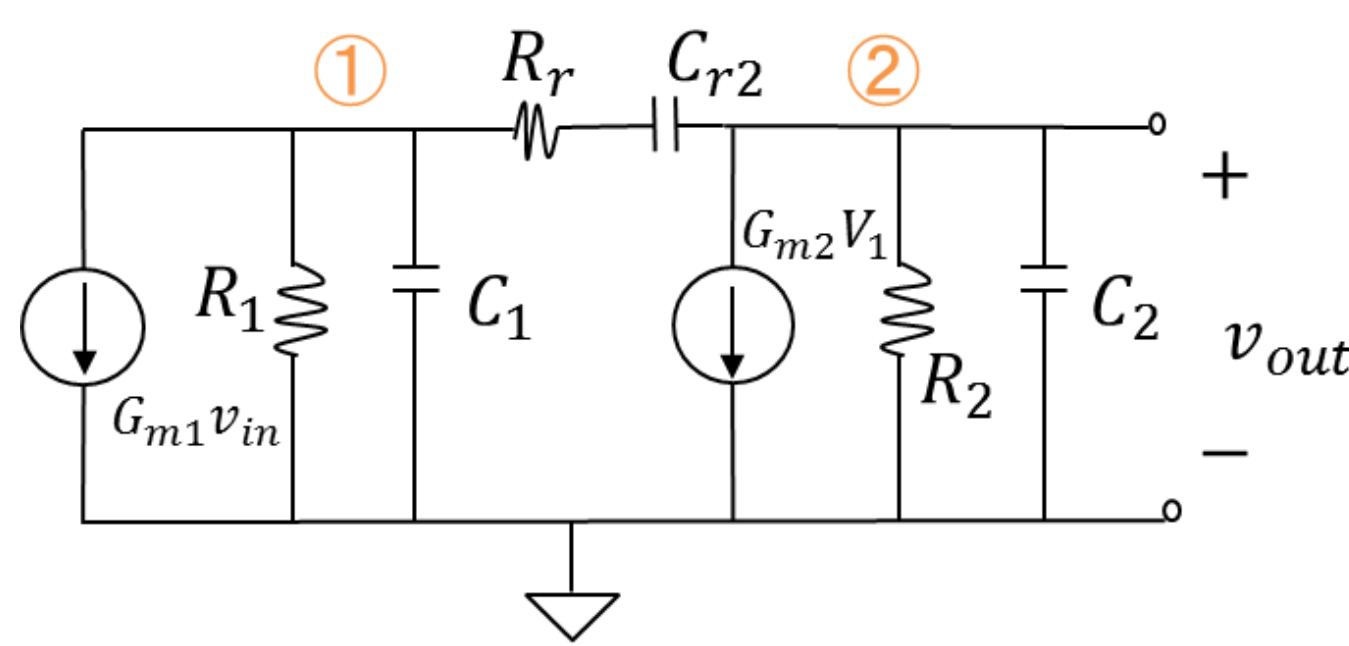


Ex.1 $-\frac{a_1}{b_1} < K < \frac{a_1}{b_1}$ At condition: $a_1 b_1 > 0$
 $\frac{a_1}{b_1} < K < -\frac{a_1}{b_1}$ At condition: $a_1 b_1 < 0$

Ex.2 $\frac{a_3 - a_1 a_2}{a_2 b - a_3} < K < \frac{a_3 - a_1 a_2}{a_3 - a_2 b}$ At condition: $(a_3 - a_1 a_2)(a_3 - a_2 b) > 0$
 $\frac{a_3 - a_1 a_2}{a_3 - a_2 b} < K < \frac{a_3 - a_1 a_2}{a_2 b - a_3}$ At condition: $(a_3 - a_1 a_2)(a_3 - a_2 b) < 0$



(a) Transistor level circuit



(b) Small-signal model

Ex.2 Two-stage amplifier with compensation network using a nulling resistor

Simulation Verification

Equivalence with Bode method

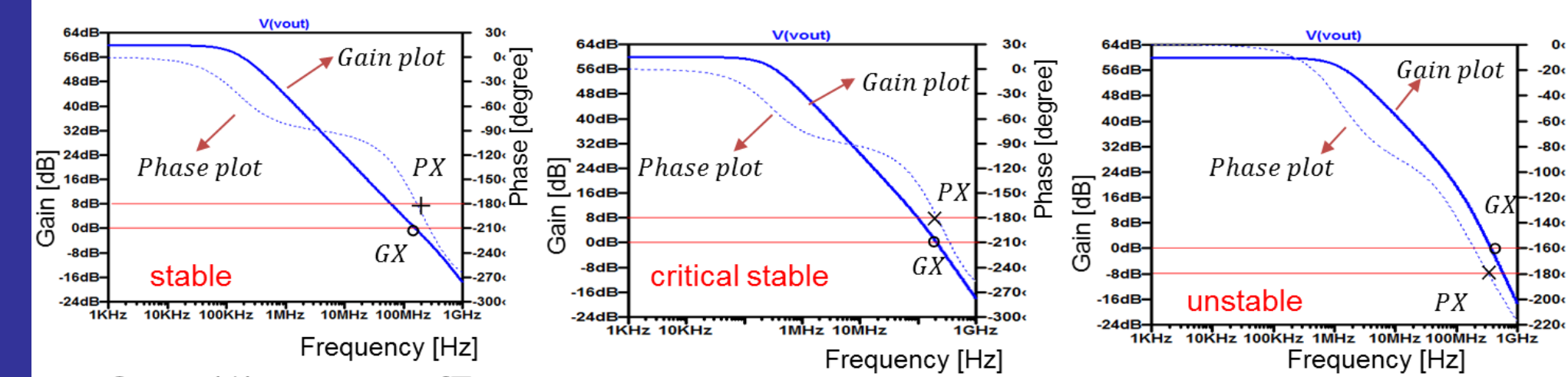
Closed-loop transfer function:

$$H(s) = \frac{A_0(1+b_1s)}{1+fA_0+(a_1+fA_0b_1)s+a_2s^2}$$

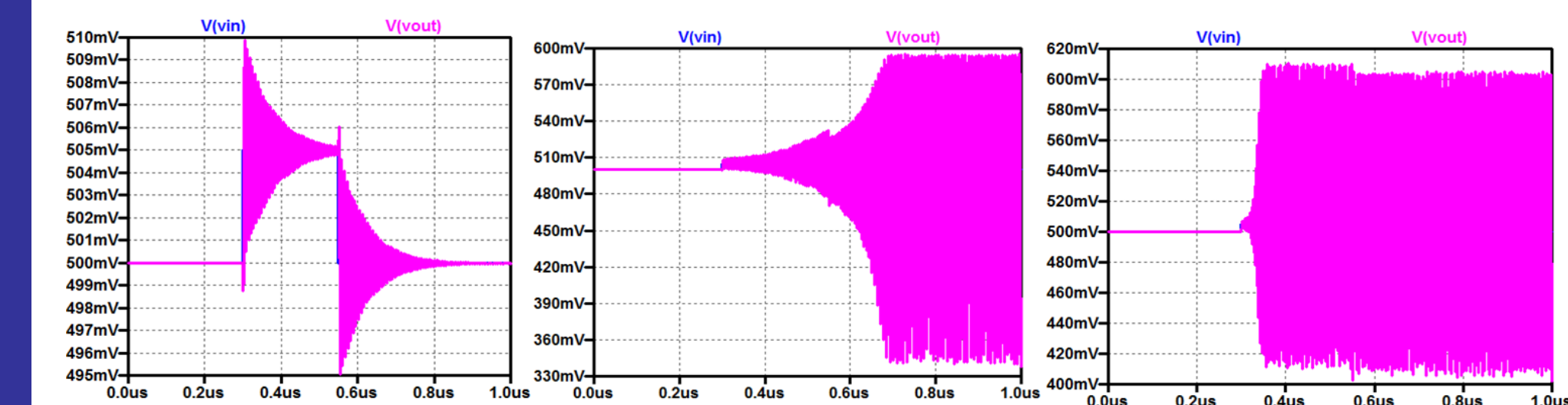
Explicit stability condition of parameters:

$$a_1 + fA_0b_1 = R_1C_1 + R_2C_2 + (R_1 + R_2)C_r + (G_{m2} - fG_{m1})R_1R_2C_r > 0$$

$$C_r > 79.57fF$$



Consistency of Bode Plots and R-H Results



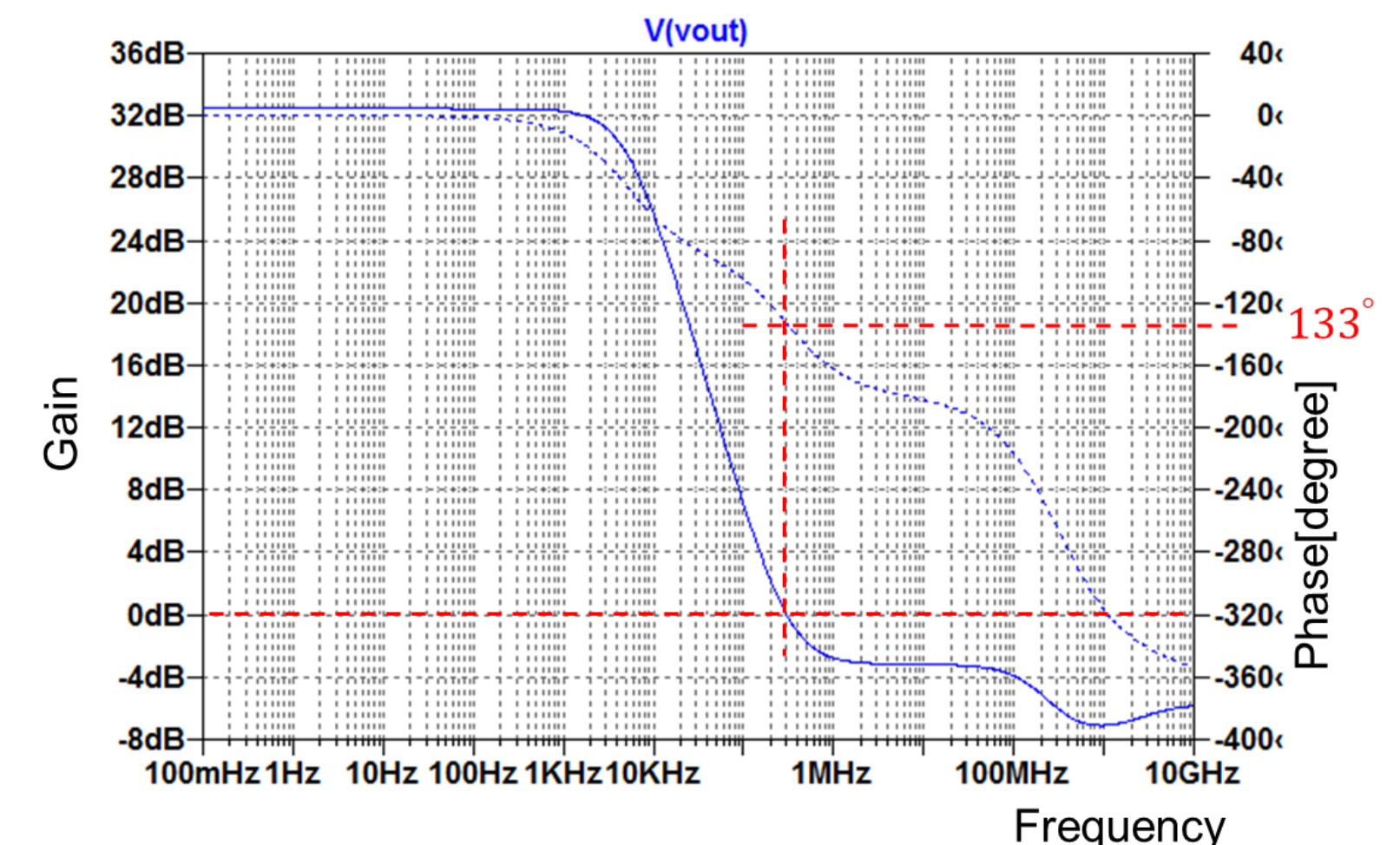
Consistency of Transient Analysis and R-H Results

Practicability

For stable feedback system, necessary PM value: 45° or 60°

$$C_{r1} = f_1(PM) = 6.343e^{-15}PM^3 - 2.091e^{-13}PM^2 + 2.493e^{-12}PM - 9.822e^{-12}$$

PM=45degree, $C_{r1} = 2.5694e^{-10}F = 0.25694nF$



Phase Margin = 180° - 133° = 47°

experimental result is close to calculation result

Summary

Discussion

Conclusion

- Depict small signal equivalent circuit
- Derive open-loop transfer function
- Derive closed-loop transfer function & obtain characteristics equation
- Apply R-H stability criterion & obtain explicit stability condition
- Use this stability condition for circuit parameters

- Equivalence between Nyquist and R-H stability criteria
- Equivalency of mathematical foundations
- R-H method, explicit circuit parameter conditions
- Consistency with Bode plot method, LTSpice simulation

R-H method can be used with conventional Bode plot method.