

Virtual TJCAS 2020
Taiwan and Japan Conference on
Circuits and Systems

STABILITY TEST FOR HIGH-ORDER
LADDER LOW-PASS FILTERS

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(Nov. 7th, 2020)



Outline

1. Research Background

- Motivation, objectives and achievements
- Self-loop function in a transfer function

2. Ringing Test for Passive Networks

- Stability test for RLC low-pass filters

3. Ringing Test for Ladder Low-Pass Filters

- Stability test for active high-order ladder low-pass filters

4. Conclusions

1. Research Background

Noise in Electronic Systems

Performance of a system

Signal to
Noise Ratio:

$$\text{SNR} = \frac{\text{Signal power}}{\text{Noise power}}$$

Common types of noise:

- Electronic noise
- Thermal noise,
- Intermodulation noise,
- Cross-talk,
- Impulse noise,
- Shot noise, and
- Transit-time noise.

Performance of a device

Figure of
Merit:

$$F = \frac{\text{Output SNR}}{\text{Input SNR}}$$

Device noise:

- Flicker noise,
- Thermal noise,
- White noise.

Linear networks

- **Overshoot,**
- **Ringing**
- **Oscillation noise**



1. Research Background

Effects of Ringing on Electronic Systems

Ringing represents a **distortion** of a signal.

Ringing is **overshoot/undershoot voltage** or current when it's seen on time domain.

Ringing does the following things:

- **Causes** EMI noise,
- **Increases** current flow,
- **Consumes** the power,
- **Decreases the** performance, and
- **Damages** the devices.

Unstable system



STABILITY TEST

1. Research Background

Objectives of Study

- **Derivation of transfer function** in electronic systems using **superposition theorem**
- **Investigation of operating regions of linear negative feedback networks**
 - **Over-damping** (**high delay** in rising time)
 - **Critical damping** (max power propagation)
 - **Under-damping** (**overshoot and ringing**)
- **Stability test** for electronic networks based on **alternating current conservation**

1. Research Background

Achievements of Study

Superposition formula for multi-source networks

$$V_o(t) \sum_{i=1}^n \frac{1}{Z_i} + V_o(t) \sum_{i=1}^n \frac{1}{Z_{si}} + \frac{1}{\sum_{k=1}^n \frac{1}{Z_{pik}}} = \sum_{i=1}^n \left(\frac{V_i(t)}{Z_i} + I_{ai}(t) - I_{gi}(t) \right)$$

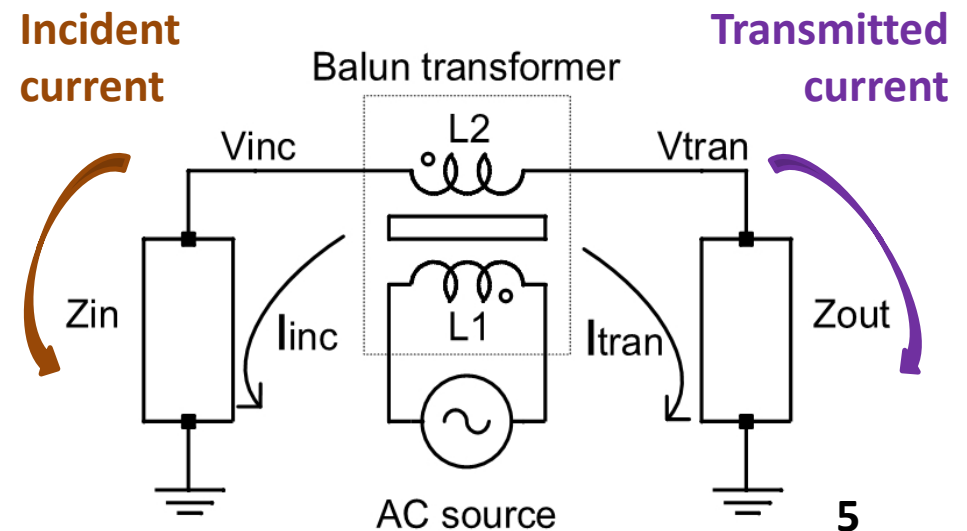
Self-loop function

$$L(\omega) = -\frac{V_{inc}}{V_{trans}}$$



10 mH
inductance

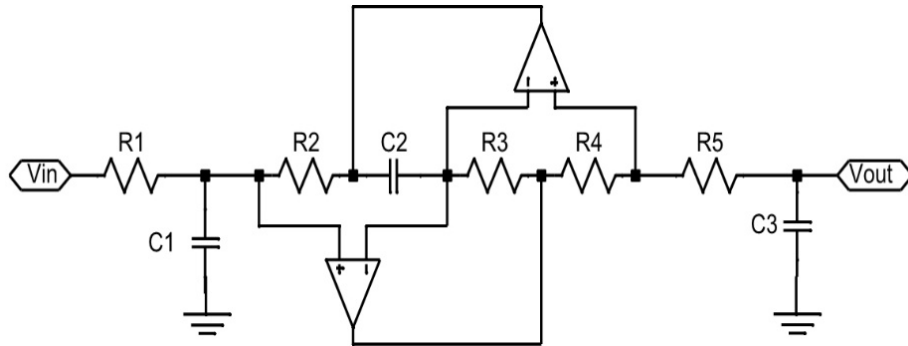
Alternating current conservation



1. Research Background

Approaching Methods

3rd-order ladder LPF

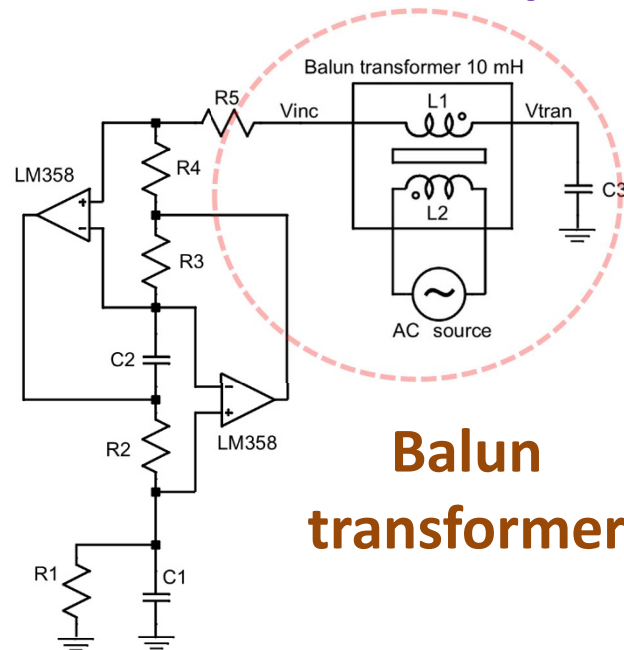


Balun transformer

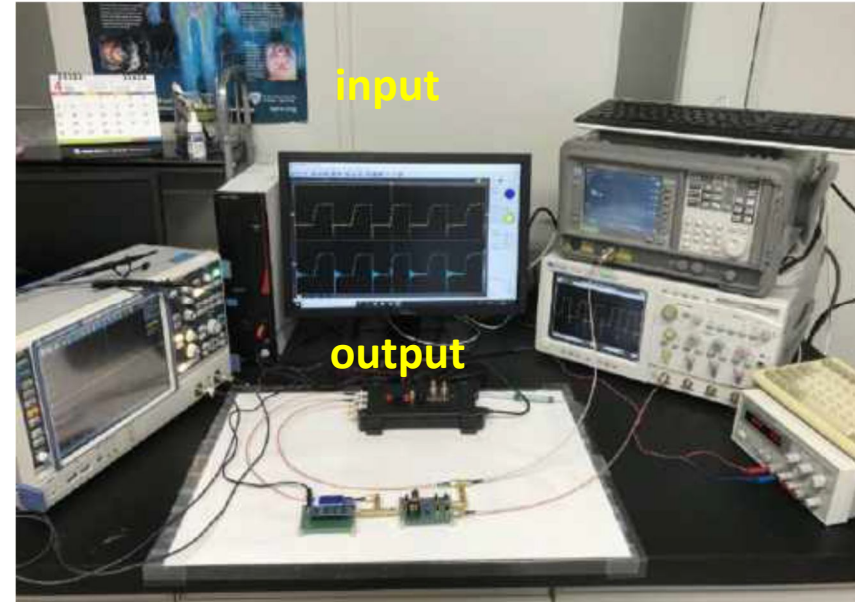


Implemented circuit

Derivation of self-loop function



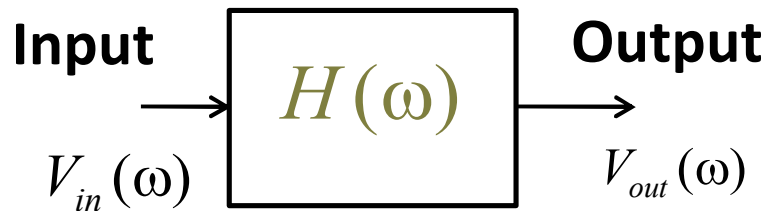
Balun transformer



1. Research Background

Self-loop Function in A Transfer Function

Linear system



Model of a linear system

$$H(\omega) = \frac{b_0(j\omega)^n + \dots + b_{n-1}(j\omega) + b_n}{a_0(j\omega)^n + \dots + a_{n-1}(j\omega) + a_n}$$

Transfer function

$$H(\omega) = \frac{V_{out}(\omega)}{V_{in}(\omega)} = \frac{A(\omega)}{1 + L(\omega)}$$

$A(\omega)$: Open loop function

$H(\omega)$: Transfer function

$L(\omega)$: Self-loop function

Variable: angular frequency (ω)

○ Polar chart → Nyquist chart

○ Magnitude-frequency plot

○ Angular-frequency plot

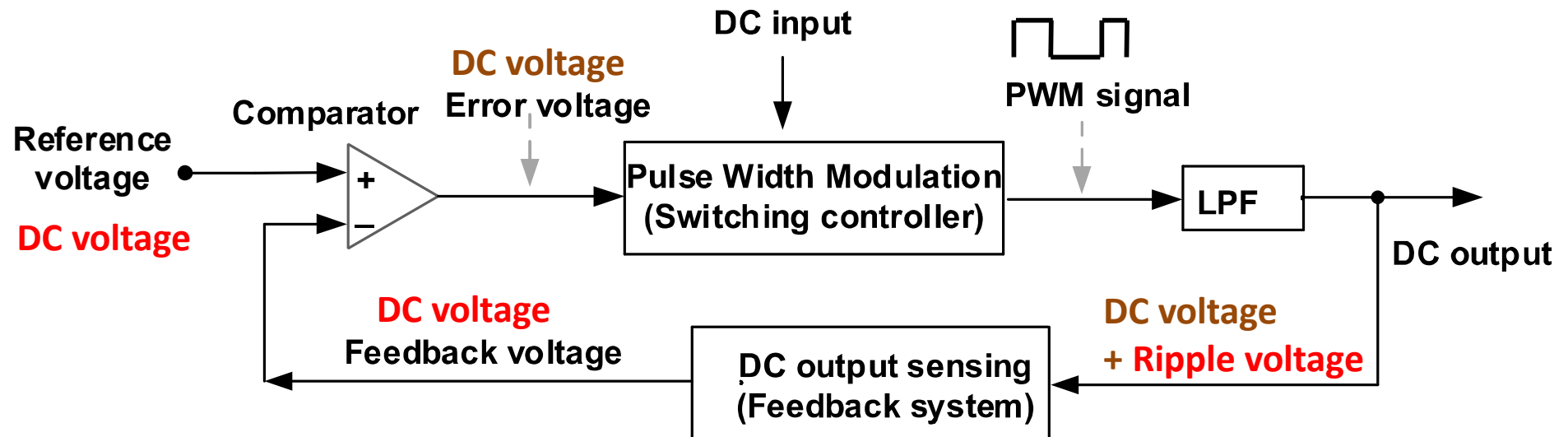
○ Magnitude-angular diagram → Nichols diagram

Bode plots

1. Research Background

Characteristics of Adaptive Feedback Network

Block diagram of a typical adaptive feedback system



Adaptive feedback is used to control the output source along with the decision source (**DC-DC Buck converter**).

Transfer function of an **adaptive feedback network** is **significantly different from** transfer function of a **linear negative feedback network**.

→ **Loop gain is independent** of frequency variable (**referent voltage, feedback voltage, and error voltage are DC voltages**).

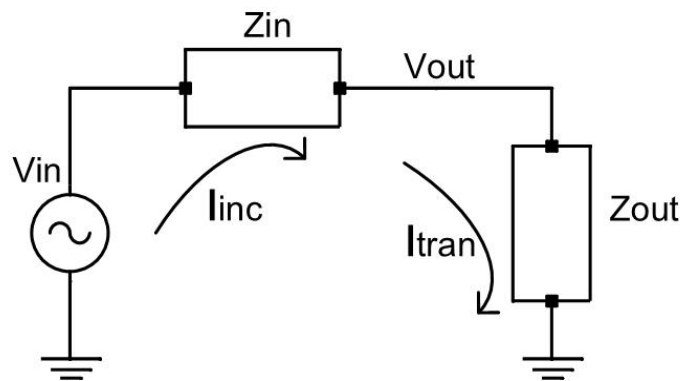
1. Research Background

Alternating Current Conservation

Transfer function

$$H(\omega) = \frac{V_{out}(\omega)}{V_{in}(\omega)} = \frac{1}{1 + \frac{Z_{in}}{Z_{out}}}$$

$$\Rightarrow L(\omega) = \frac{Z_{in}}{Z_{out}};$$



Simplified linear system

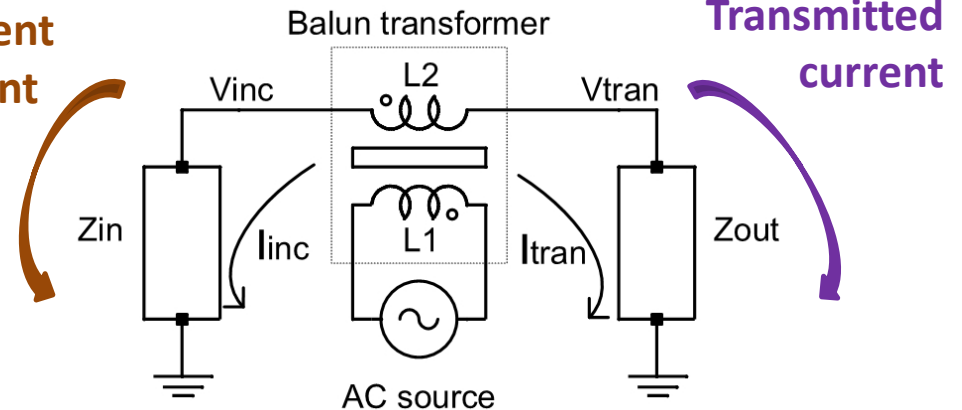
Self-loop function

$$\frac{V_{inc}}{Z_{in}} = -\frac{V_{trans}}{Z_{out}} \Rightarrow L(\omega) = -\frac{V_{inc}}{V_{trans}} = \frac{Z_{in}}{Z_{out}}$$



10 mH
inductance

Incident
current



Derivation of self-loop function

1. Research Background

Limitations of Conventional Methods

- **Middlebrook's measurement of loop gain**
→ Applying only in feedback systems (**DC-DC converters**).
- **Replica measurement of loop gain**
→ Using two identical networks (**not real measurement**).
- **Nyquist's stability condition**
→ Theoretical analysis for feedback systems (**Lab tool**).
- **Nichols chart of loop gain**
→ Only used in feedback control theory (**Lab tool**).
- **Conventional superposition**
→ Solving for every source (**several times**).

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2. Ringing Test for Passive Networks

- **Stability test for RLC low-pass filters**

3. Ringing Test for Ladder Low-Pass Filters

- Stability test for active high-order ladder low-pass filters

4. Conclusions

2. Ringing Test for Passive Networks

Characteristics of 2nd-order Transfer Function

Second-order transfer function:
$$H(\omega) = \frac{1}{1 + a_0(j\omega)^2 + a_1j\omega}$$

Case	Over-damping	Critical damping	Under-damping
Delta (Δ)	$\frac{1}{a_0} < \left(\frac{a_1}{2a_0}\right)^2 \Rightarrow \Delta = a_1^2 - 4a_0 > 0$	$\frac{1}{a_0} = \left(\frac{a_1}{2a_0}\right)^2 \Rightarrow \Delta = a_1^2 - 4a_0 = 0$	$\frac{1}{a_0} > \left(\frac{a_1}{2a_0}\right)^2 \Rightarrow \Delta = a_1^2 - 4a_0 < 0$
Module $ H(\omega) $	$\frac{1}{a_0} \sqrt{\omega^2 + \left(\frac{a_1}{2a_0} - \sqrt{\left(\frac{a_1}{2a_0}\right)^2 - \frac{1}{a_0}}\right)^2} \sqrt{\omega^2 + \left(\frac{a_1}{2a_0} + \sqrt{\left(\frac{a_1}{2a_0}\right)^2 - \frac{1}{a_0}}\right)^2}$	$\frac{1}{a_0} \sqrt{\omega^2 + \left(\frac{a_1}{2a_0}\right)^2} = \frac{1}{2} = -6dB$	$\frac{1}{a_0} \sqrt{\left(\omega - \sqrt{\frac{1}{a_0} - \left(\frac{a_1}{2a_0}\right)^2}\right)^2 + \left(\frac{a_1}{2a_0}\right)^2} \sqrt{\left(\omega + \sqrt{\frac{1}{a_0} - \left(\frac{a_1}{2a_0}\right)^2}\right)^2 + \left(\frac{a_1}{2a_0}\right)^2}$
Angular $\theta(\omega)$	$-\arctan\left(\frac{\omega}{\frac{a_1}{2a_0} - \sqrt{\left(\frac{a_1}{2a_0}\right)^2 - \frac{1}{a_0}}}\right) - \arctan\left(\frac{\omega}{\frac{a_1}{2a_0} + \sqrt{\left(\frac{a_1}{2a_0}\right)^2 - \frac{1}{a_0}}}\right)$	$-2 \arctan\left(\frac{2a_0\omega}{a_1}\right)$	$-\arctan\left(\frac{\omega - \sqrt{\frac{1}{a_0} - \left(\frac{a_1}{2a_0}\right)^2}}{\frac{a_1}{2a_0}}\right) - \arctan\left(\frac{\omega + \sqrt{\frac{1}{a_0} - \left(\frac{a_1}{2a_0}\right)^2}}{\frac{a_1}{2a_0}}\right)$
$\omega_{cut} = \frac{a_1}{2a_0}$	$ H(\omega_{cut}) < \frac{2a_0}{a_1}$ $\theta(\omega_{cut}) > -\frac{\pi}{2}$	$ H(\omega_{cut}) = \frac{2a_0}{a_1}$ $\theta(\omega_{cut}) = -\frac{\pi}{2}$	$ H(\omega_{cut}) > \frac{2a_0}{a_1}$ $\theta(\omega_{cut}) < -\frac{\pi}{2}$

2. Ringing Test for Passive Networks

Characteristics of 2nd-order Self-loop Function

Second-order self-loop function: $L(\omega) = j\omega[a_0 j\omega + a_1]$

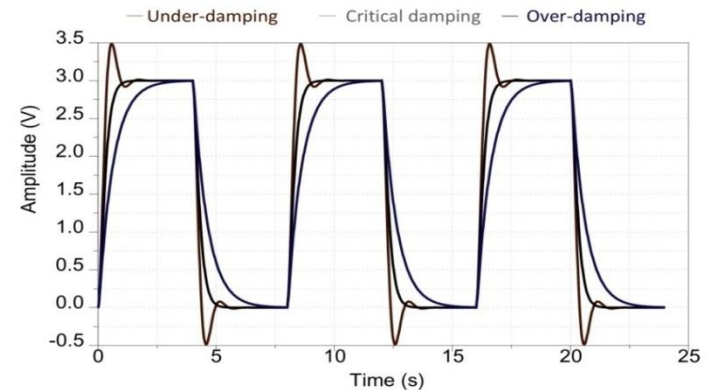
Case	Over-damping	Critical damping	Under-damping			
Delta (Δ)	$\Delta = a_1^2 - 4a_0 > 0$	$\Delta = a_1^2 - 4a_0 = 0$	$\Delta = a_1^2 - 4a_0 < 0$			
$ L(\omega) $	$\omega\sqrt{(a_0\omega)^2 + a_1^2}$	$\omega\sqrt{(a_0\omega)^2 + a_1^2}$	$\omega\sqrt{(a_0\omega)^2 + a_1^2}$			
$\theta(\omega)$	$\frac{\pi}{2} + \arctan \frac{a_0\omega}{a_1}$	$\frac{\pi}{2} + \arctan \frac{a_0\omega}{a_1}$	$\frac{\pi}{2} + \arctan \frac{a_0\omega}{a_1}$			
$\omega_1 = \frac{a_1}{2a_0}\sqrt{\sqrt{5}-2}$	$ L(\omega_1) > 1$	$\pi - \theta(\omega_1) > 76.3^\circ$	$ L(\omega_1) < 1$	$\pi - \theta(\omega_1) < 76.3^\circ$		
$\omega_2 = \frac{a_1}{2a_0}$	$ L(\omega_2) > \sqrt{5}$	$\pi - \theta(\omega_2) > 63.4^\circ$	$ L(\omega_2) = \sqrt{5}$	$\pi - \theta(\omega_2) = 63.4^\circ$	$ L(\omega_2) < \sqrt{5}$	$\pi - \theta(\omega_2) < 63.4^\circ$
$\omega_3 = \frac{a_1}{a_0}$	$ L(\omega_3) > 4\sqrt{2}$	$\pi - \theta(\omega_3) > 45^\circ$	$ L(\omega_3) = 4\sqrt{2}$	$\pi - \theta(\omega_3) = 45^\circ$	$ L(\omega_3) < 4\sqrt{2}$	$\pi - \theta(\omega_3) < 45^\circ$

2. Ringing Test for Passive Networks

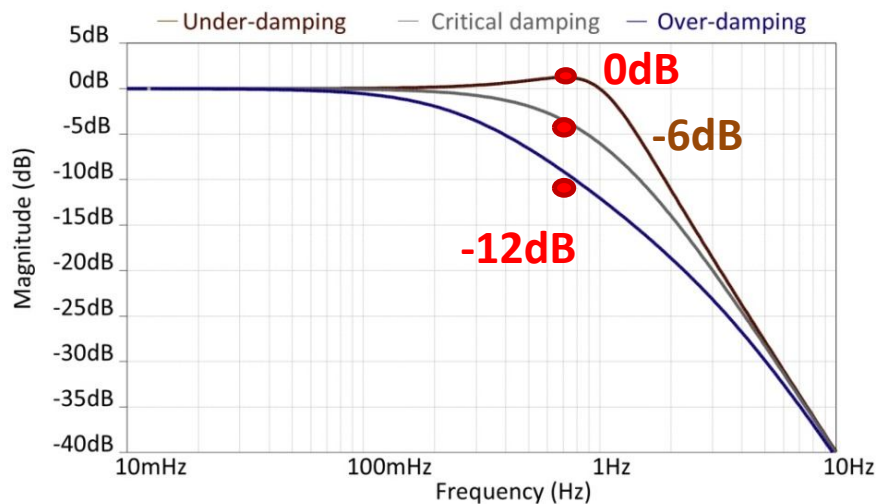
Operating Regions of 2nd-Order System

- **Under-damping:** $H_1(\omega) = \frac{1}{(j\omega)^2 + j\omega + 1}$;
 $L_1(\omega) = (j\omega)^2 + j\omega$;
- **Critical damping:** $H_2(\omega) = \frac{1}{(j\omega)^2 + 2j\omega + 1}$;
 $L_2(\omega) = (j\omega)^2 + 2j\omega$;
- **Over-damping:** $H_3(\omega) = \frac{1}{(j\omega)^2 + 3j\omega + 1}$;
 $L_3(\omega) = (j\omega)^2 + 3j\omega$;

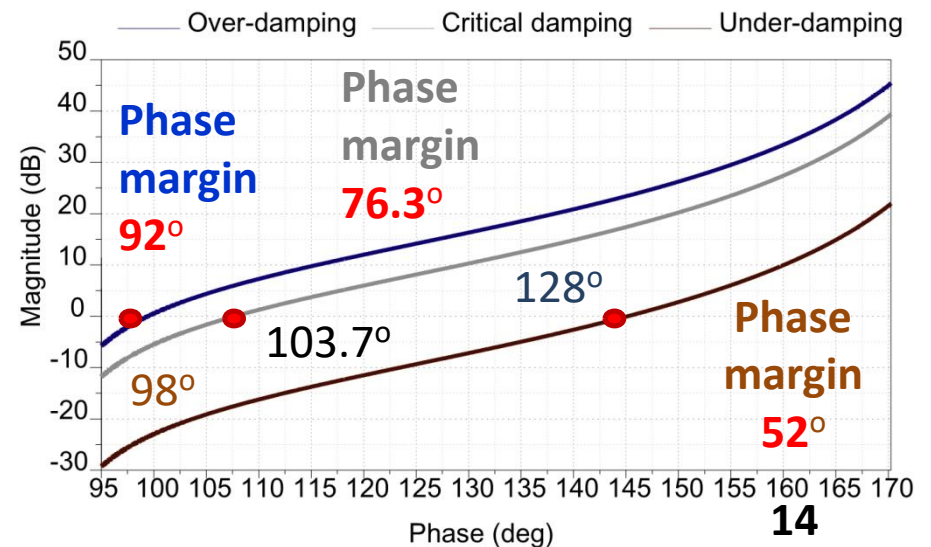
Transient response



Bode plot of transfer function



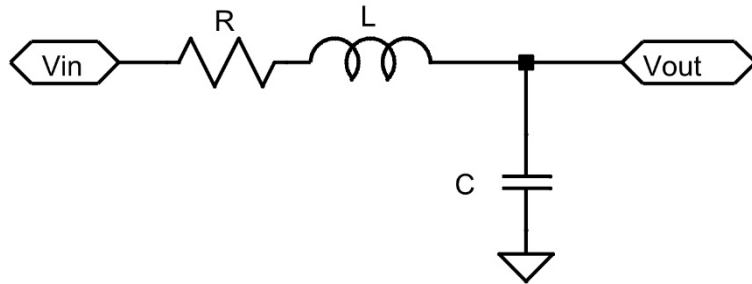
Nichols plot of self-loop function



2. Ringing Test for Passive Networks

Stability Test for Passive 2rd-Order RLC LPF

Passive RLC Low-pass Filter



Transfer function

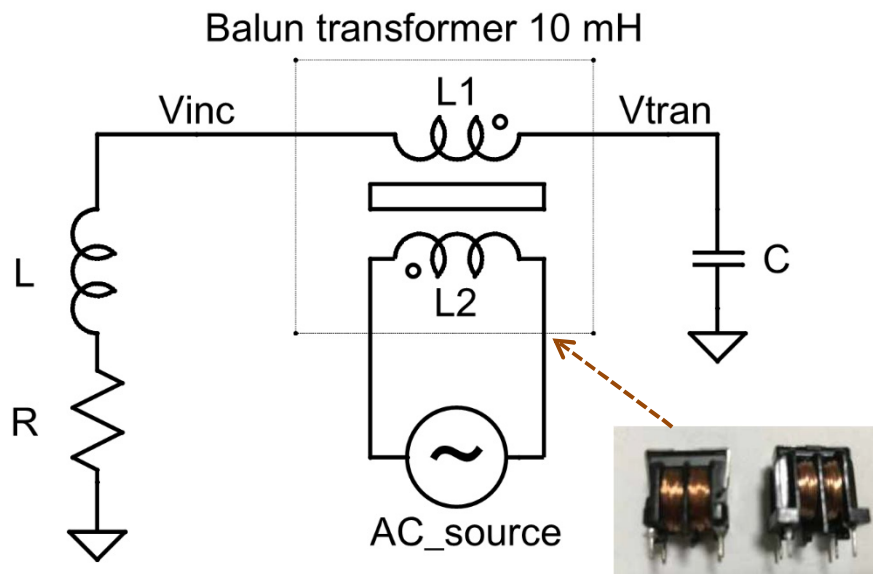
$$H(\omega) = \frac{V_{out}}{V_{in}} = \frac{1}{a_0 (j\omega)^2 + a_1 j\omega + 1};$$

Self-loop function

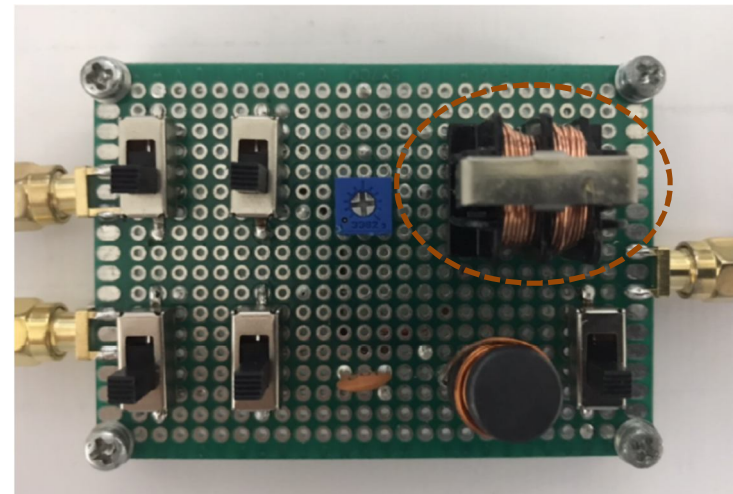
$$L(\omega) = a_0 (j\omega)^2 + a_1 j\omega;$$

where, $a_0 = LC$; $a_1 = RC$;

Derivation of self-loop function



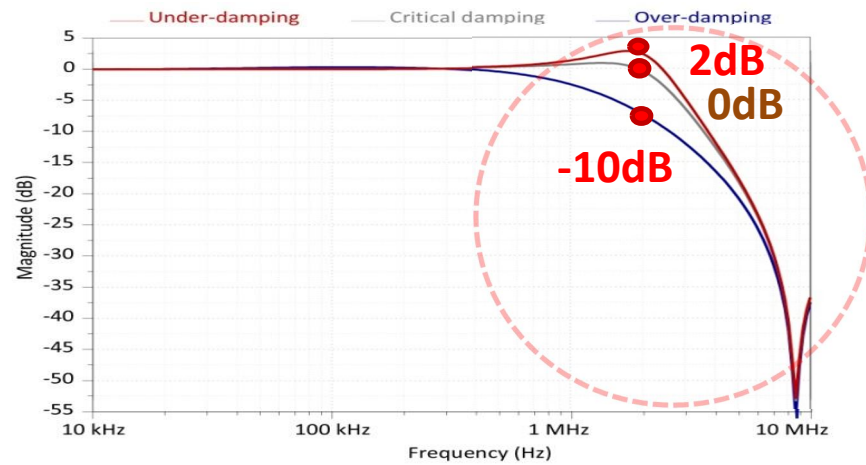
Implemented circuit



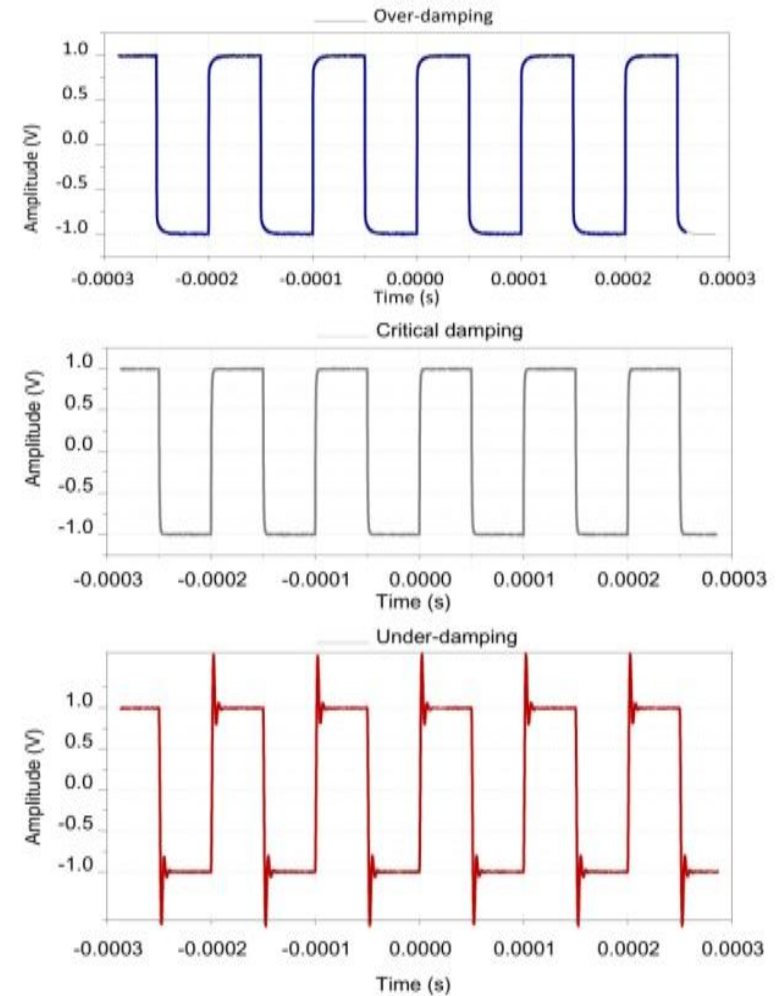
2. Ringing Test for Passive Networks

Stability Test for 2rd-Order Passive RLC LPF

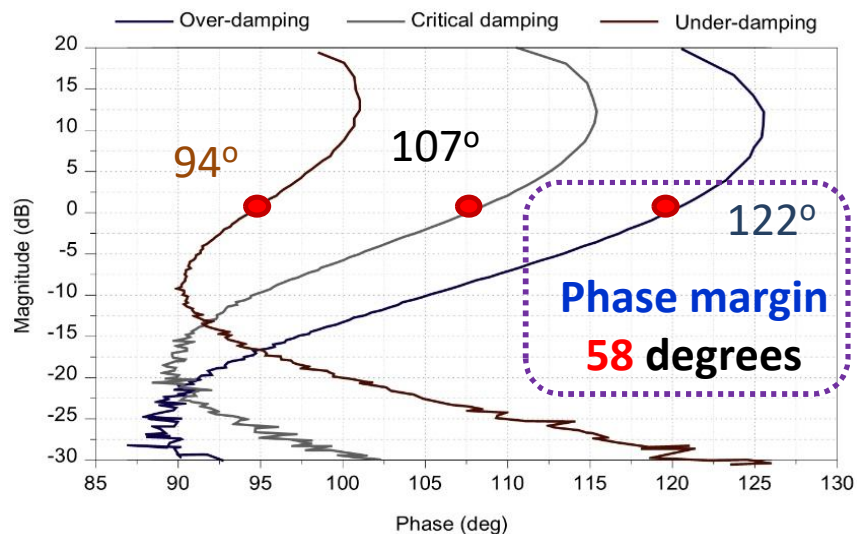
Bode plot of transfer function



Transient responses



Nichols plot of self-loop function



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2. Ringing Test for Passive Networks

- Stability test for RLC low-pass filters

3. Ringing Test for Ladder Low-Pass Filters

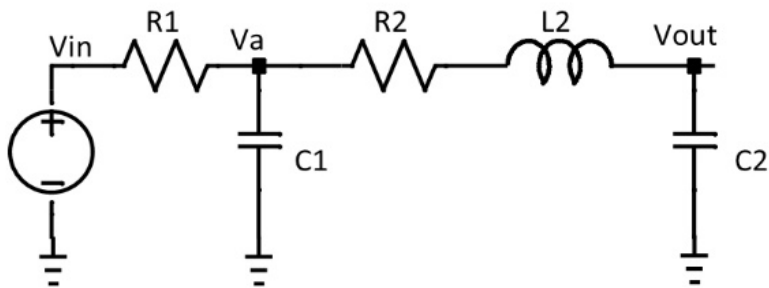
- **Stability test for active high-order ladder low-pass filters**

4. Conclusions

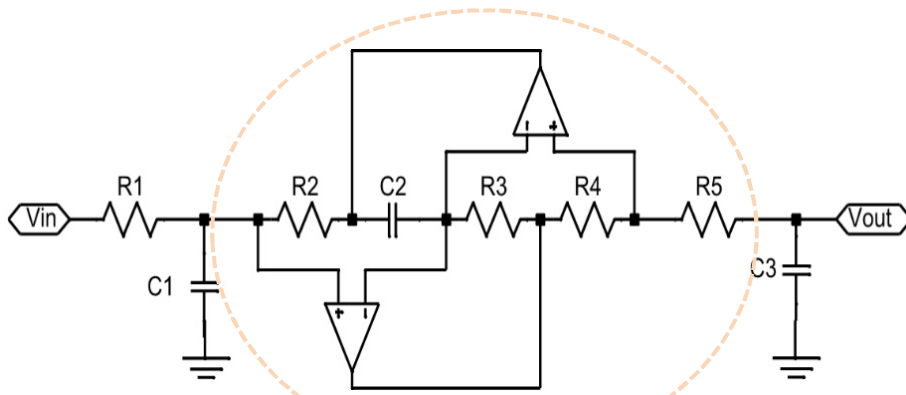
3. Ringing Test for Ladder Low-Pass Filters

Analysis of Active 3rd-Order Ladder LPF

Passive 3rd-order ladder LPF



Active 3rd-order ladder LPF



General impedance converter

Transfer function & self-loop function

$$H_{out}(\omega) = \frac{V_{out}}{V_{in}} = \frac{1}{a_0(j\omega)^3 + a_1(j\omega)^2 + a_2j\omega + 1};$$

$$L(\omega) = j\omega [a_0(j\omega)^2 + a_1j\omega + a_2]$$

where, $b_0 = L_2C_2; b_1 = R_2C_2;$

$a_0 = R_1C_1L_2C_2; a_1 = R_1C_1R_2C_2 + L_2C_2;$

$a_2 = R_1(C_1 + C_2) + R_2C_2;$

R1 = 100 Ω, R2 = 50 kΩ,

R3 = R4 = 50 kΩ, C1 = 5 nF, C2 = 10 nF,

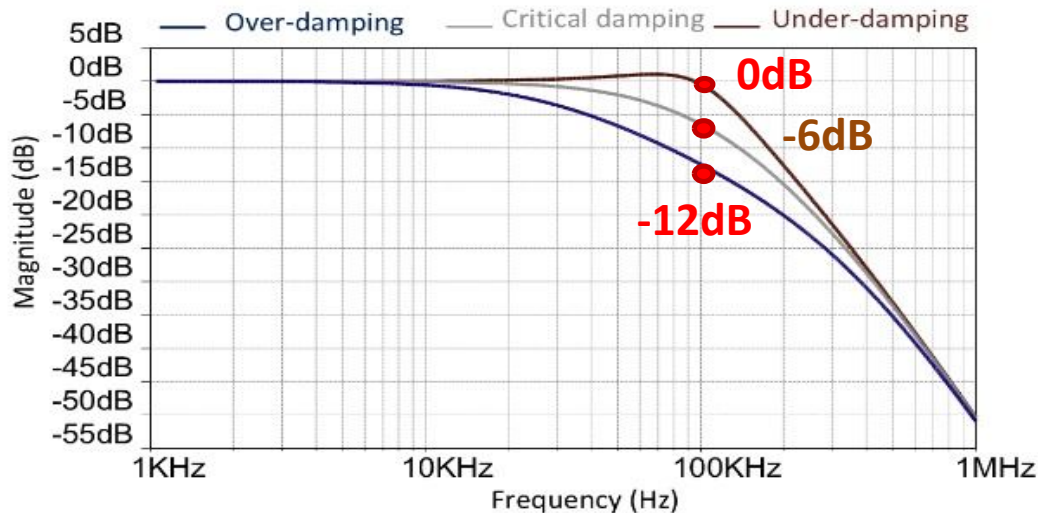
C3 = 3.18 nF, at $f_0 = 100$ kHz.

- **Over-damping** (R5 = 0.5 kΩ),
- **Critical damping** (R5 = 1 kΩ), and
- **Under-damping** (R5 = 2 kΩ).

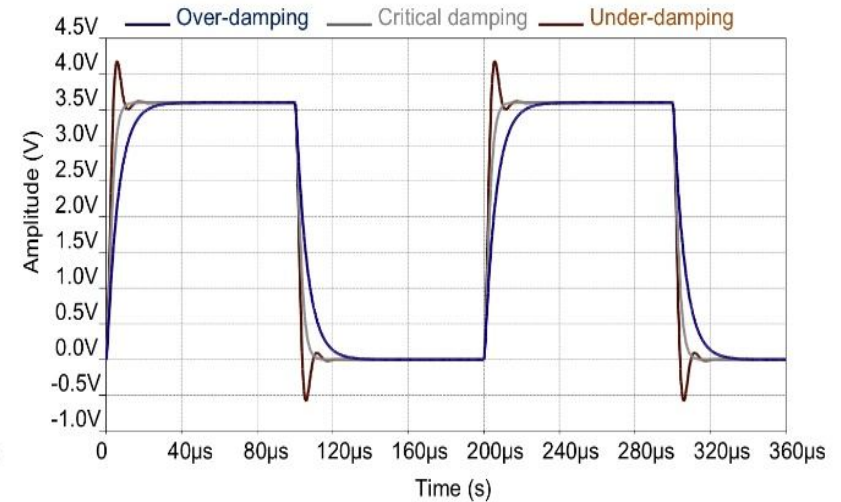
3. Ringing Test for Ladder Low-Pass Filters

Simulation Results of 3rd-Order Ladder LPF

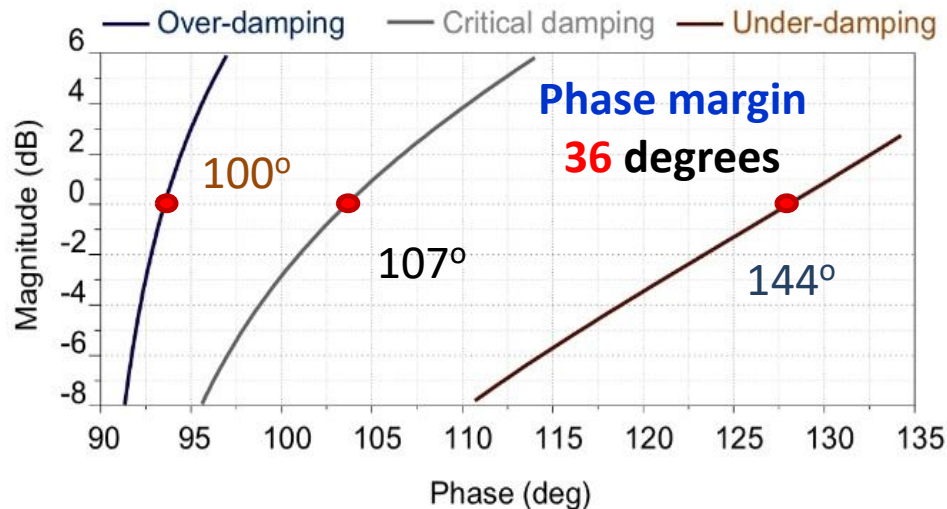
Bode plot of transfer function



Transient response



Nichols plot of self-loop function



Over-damping:

→ Phase margin is **80** degrees.

Critical damping:

→ Phase margin is **73** degrees.

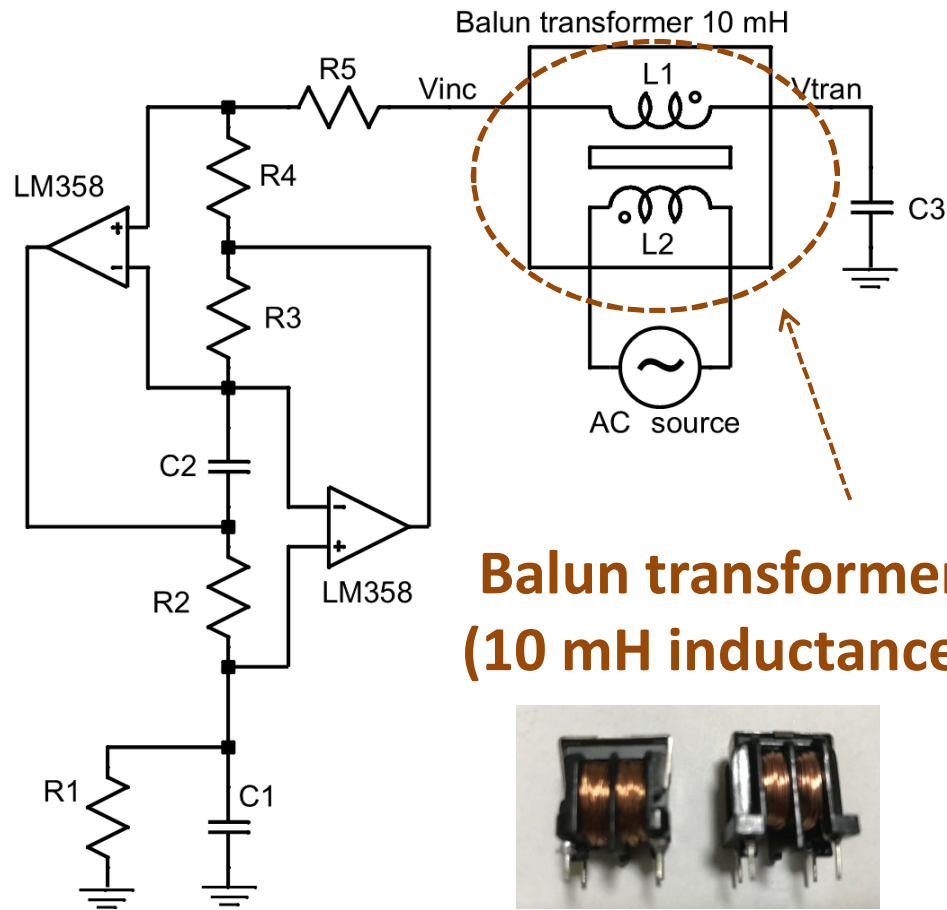
Under-damping:

→ Phase margin is **36** degrees.

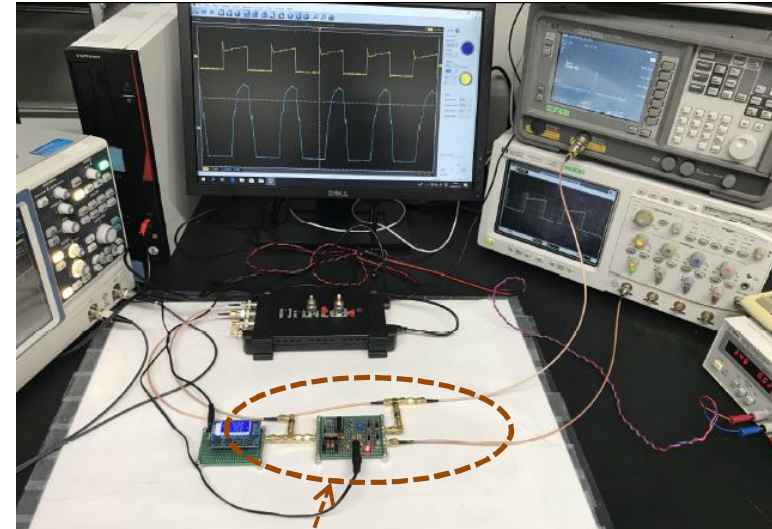
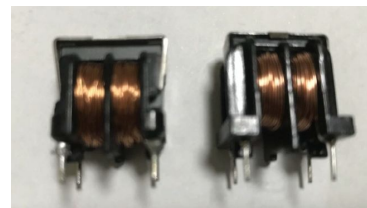
3. Ringing Test for Ladder Low-Pass Filters

Implemented Circuit of 3rd-Order Ladder LPF

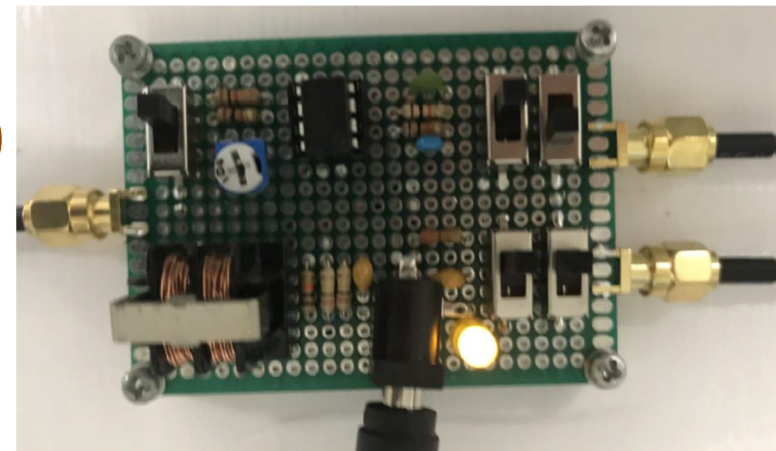
Measurement of self-loop function



**Balun transformer
(10 mH inductance)**



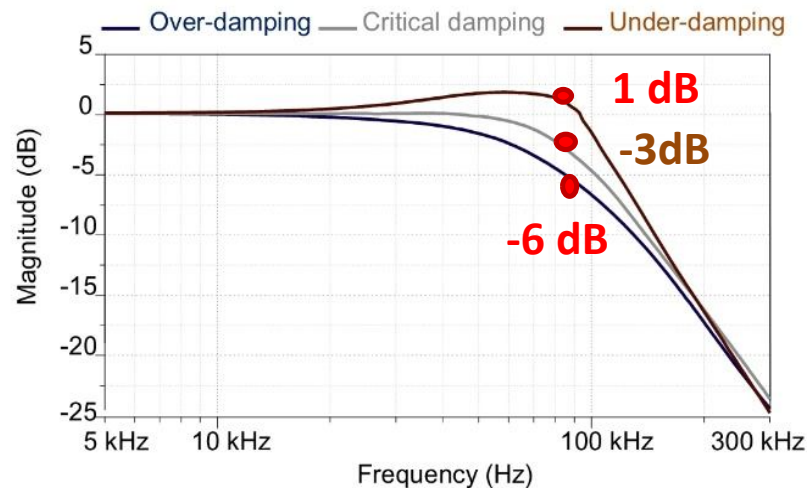
Device under test



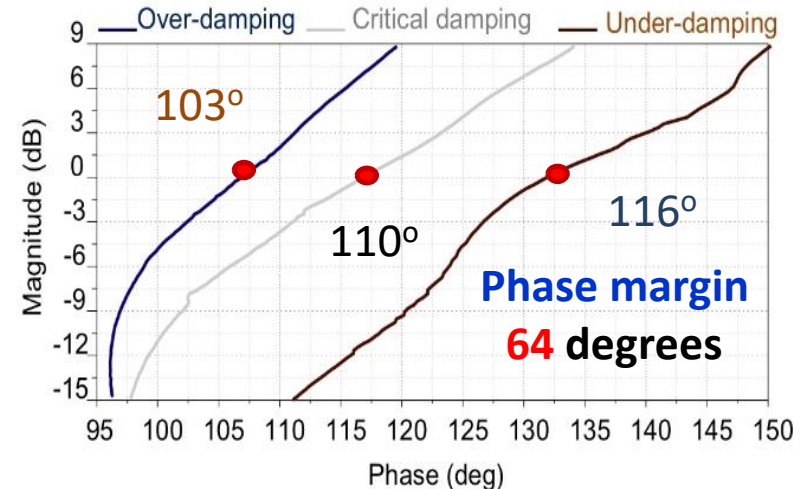
3. Ringing Test for Ladder Low-Pass Filters

Measurement Results of 3rd-order Ladder LPF

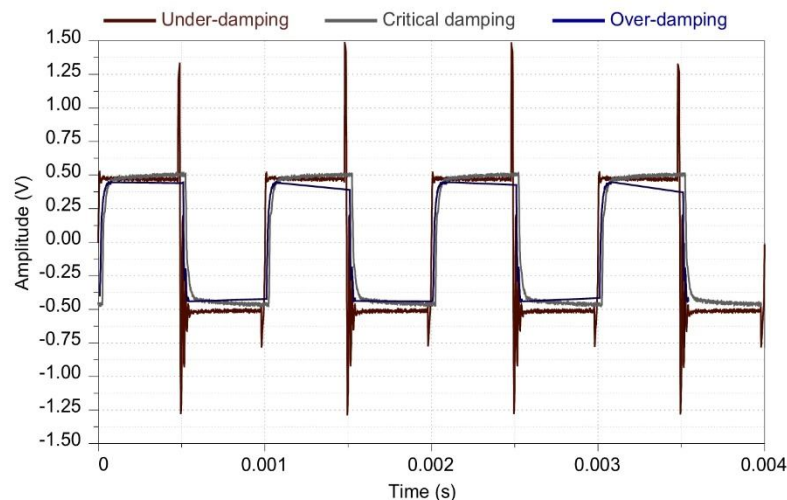
Bode plot of transfer function



Nichols plot of self-loop function



Transient response



Over-damping:

→ Phase margin is **77** degrees.

Critical damping:

→ Phase margin is **70** degrees.

Under-damping:

→ Phase margin is **64** degrees.

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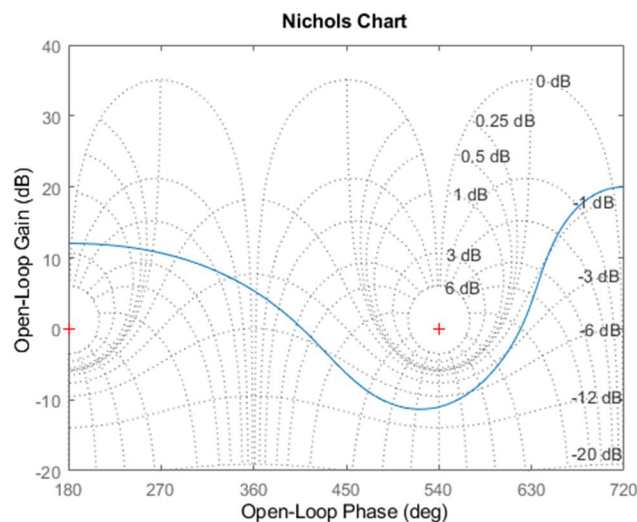
4. Comparison (Self-loop function)

Features	Alternating current conservation	Replica measurement	Middlebrook's method
Main objective	Self-loop function	Loop gain	Loop gain
Transfer function accuracy	Yes	No	No
Ringling Test	Yes	Yes	Yes
Operating region accuracy	Yes	No	No
Phase margin accuracy	Yes	No	No
Passive networks	Yes	No	No

4. Discussions (Self-loop function)

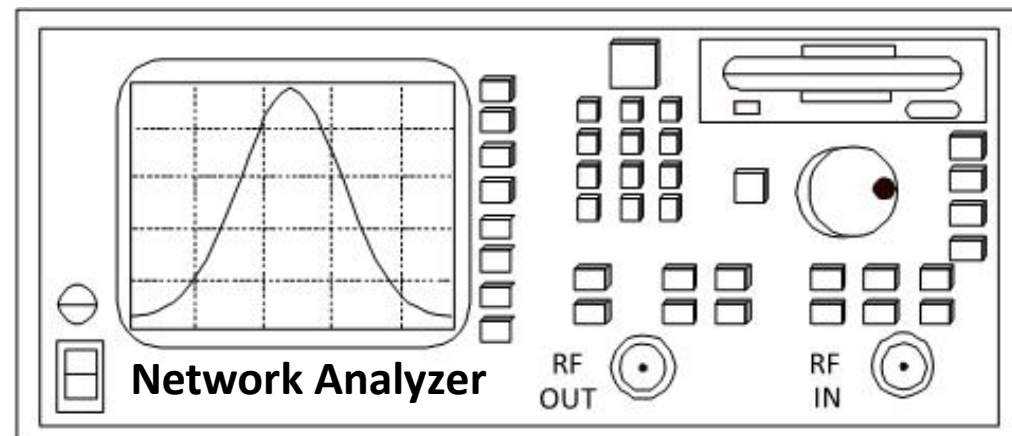
- Loop gain is **independent of** frequency variable.
- Loop gain in adaptive feedback network is **significantly different from** self-loop function in linear negative feedback network.

Nichols chart is **only used** in **MATLAB simulation**.



<https://www.mathworks.com/help/control/ref/nichols.html>

Nichols chart **isn't** used **widely** in practical measurements (**only used** in control theory).



➔ **(Technology limitations)**

4. Conclusions

This work:

- Proposal of alternating current conservation for deriving **self-loop function** in a transfer function
→ **Observation of self-loop function** can help us **optimize the behavior** of a high-order system.
- Implementations of circuits and measurements of self-loop functions for **passive & active low-pass filters**
→ **Theoretical concepts of stability test** are verified by **laboratory simulations** and **practical experiments**.

Future of work:

- **Stability test** for **parasitic components** in transmission lines, printed circuit boards, physical layout layers

References

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Thank you very much!

