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# RINGING TEST FOR THIRD-ORDER LADDER LOW-PASS FILTERS

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# Outline

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## 1. Research Background

- Motivation, objectives and achievements
- Self-loop function in a transfer function

## 2. Ringing Test for Passive Networks

- Stability test for RLC low-pass filters

## 3. Ringing Test for Unity-Gain Amplifiers

- Stability test for unity-gain amplifiers

## 4. Ringing Test for Ladder Low-Pass Filters

- Stability test for active 3<sup>rd</sup>-order ladder low-pass filters

## 5. Conclusions

# 1. Research Background

## Noise in Electronic Systems

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### Performance of a system

Signal to  
Noise Ratio:

$$\text{SNR} = \frac{\text{Signal power}}{\text{Noise power}}$$

### Common types of noise:

- Electronic noise
- Thermal noise,
- Intermodulation noise,
- Cross-talk,
- Impulse noise,
- Shot noise, and
- Transit-time noise.

### Performance of a device

Figure of  
Merit:

$$F = \frac{\text{Output SNR}}{\text{Input SNR}}$$

### Device noise:

- Flicker noise,
- Thermal noise,
- White noise.

### Linear networks

- Overshoot,
- Ringing
- Oscillation noise



# 1. Research Background

## Effects of Ringing on Electronic Systems

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**Ring**ing represents a **distortion** of a signal.

**Ring**ing is **overshoot/undershoot voltage** or current when it's seen on time domain.

**Ring**ing does the following things:

- **Causes** EMI noise,
- **Increases** current flow,
- **Consumes** the power,
- **Decreases the** performance, and
- **Damages** the devices.

Unstable system



**STABILITY TEST**

# 1. Research Background

## Objectives of Study

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- **Derivation of transfer function** in electronic systems using **superposition theorem**
- **Investigation of operating regions of linear negative feedback networks**
  - **Over-damping** (**high delay** in rising time)
  - **Critical damping** (max power propagation)
  - **Under-damping** (**overshoot and ringing**)
- **Stability test** for electronic networks based on **alternating current conservation**

# 1. Research Background

## Achievements of Study

### Superposition formula for multi-source networks

$$V_o(t) \sum_{i=1}^n \frac{1}{Z_i} + V_o(t) \sum_{i=1}^n \frac{1}{Z_{si}} + \frac{1}{\sum_{k=1}^n \frac{1}{Z_{pik}}} = \sum_{i=1}^n \left( \frac{V_i(t)}{Z_i} + I_{ai}(t) - I_{gi}(t) \right)$$

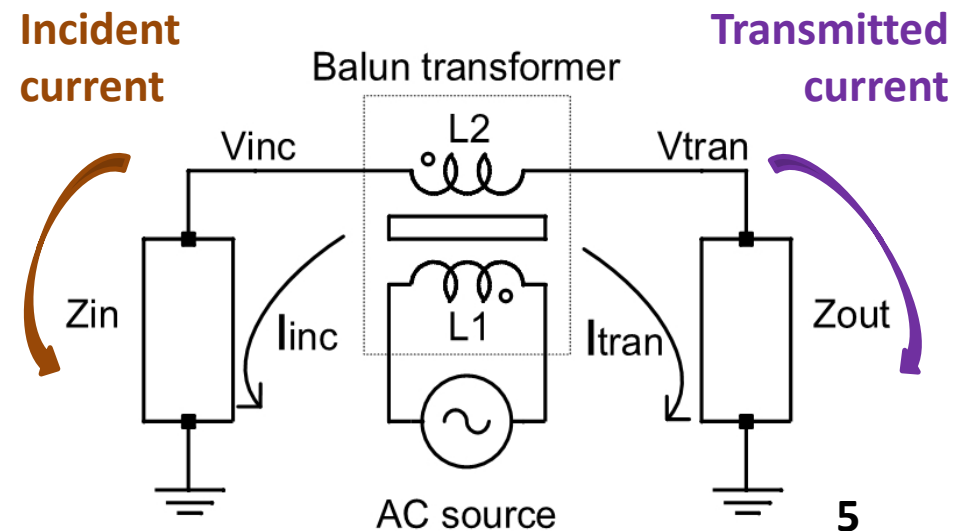
### Self-loop function

$$L(\omega) = -\frac{V_{inc}}{V_{trans}}$$



10 mH  
inductance

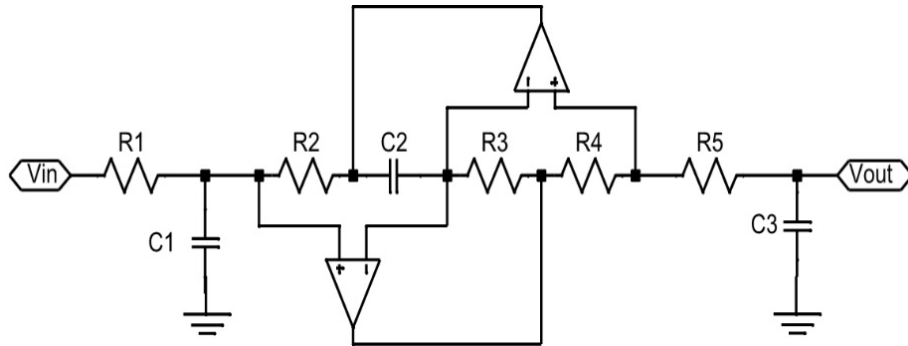
### Alternating current conservation



# 1. Research Background

## Approaching Methods

### 3<sup>rd</sup>-order ladder LPF

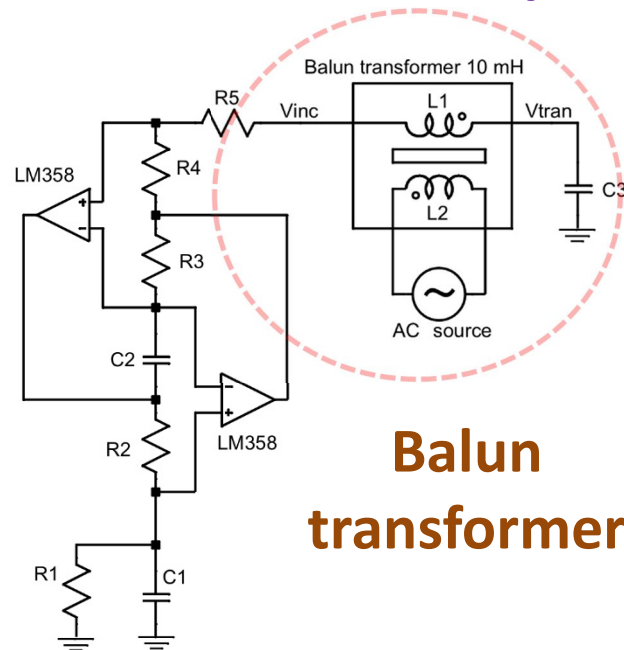


Balun transformer

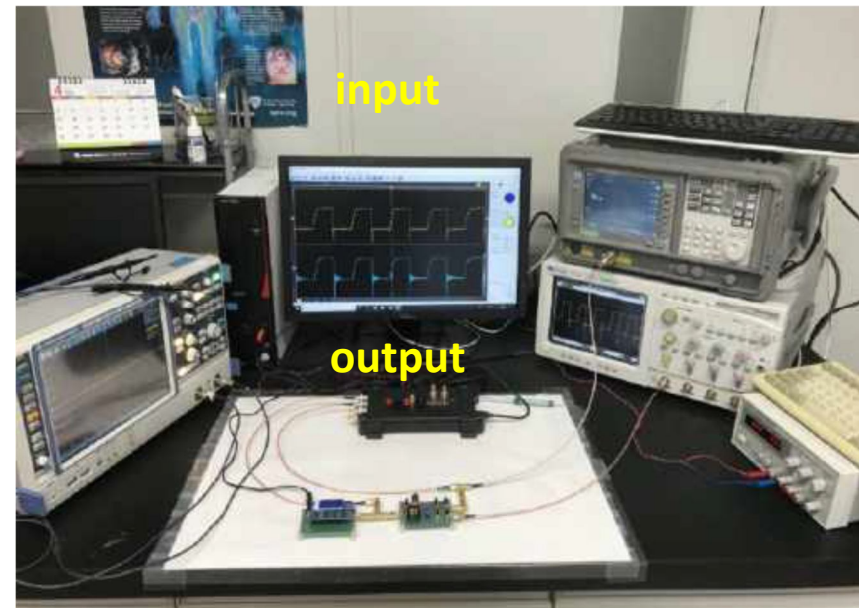


Implemented circuit

### Derivation of self-loop function



Balun transformer



# 1. Research Background

## Superposition Theorem for Multi-Source Systems

Superposition formula:

$$V_o(t) \sum_{i=1}^n \frac{1}{Z_i} + V_o(t) \sum_{i=1}^n \frac{1}{Z_{si}} + \frac{1}{\sum_{k=1}^n \frac{1}{Z_{pik}}} = \sum_{i=1}^n \left( \frac{V_i(t)}{Z_i} + I_{ai}(t) - I_{gi}(t) \right)$$

$V_o(t)$  : Voltage at one node

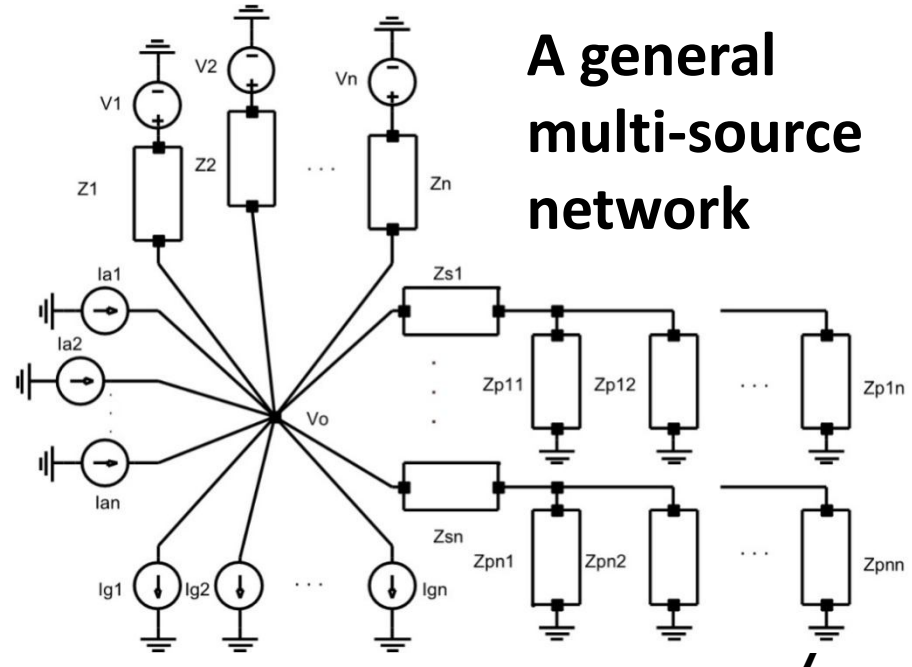
$V_i(t)$  : Input voltage sources

$I_{ai}(t)$  : Ahead-toward current sources

$I_{gi}(t)$  : Ground-toward current sources

$Z_i, s_i, p_i, (t)$ : Impedances at each branch

- Multi-source systems, feedback networks (op amps, amplifiers), polyphase filters, complex filters...

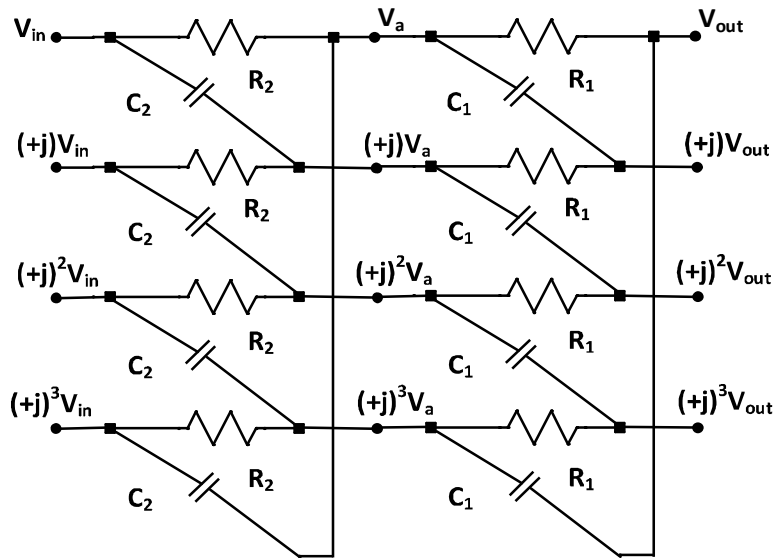




# 1. Research Background

## Analysis of 2<sup>nd</sup>-Order Polyphase Filter

### 2<sup>nd</sup>-order RC polyphase filter



Apply superposition at each node

$$V_{out} \left( \frac{1}{Z_{C1}} + \frac{1}{R_1} \right) = \frac{V_a}{R_1} + \frac{(+j)^3 V_a}{Z_{C1}};$$

$$V_a \left( \frac{1}{Z_{C2}} + \frac{1}{R_2} + \frac{2}{R_1 + Z_{C1}} \right) = \frac{V_{in}}{R_2} + \frac{(+j)^3 V_{in}}{Z_{C2}};$$

Transfer function for **positive** polyphase signal

$$H_P(\omega) = \frac{V_{out}}{V_{in}} = \frac{\left[ 1 + (+j)^3 b_1 j\omega \right] \left[ 1 + (+j)^3 b_2 j\omega \right]}{a_0 (j\omega)^2 + a_1 j\omega + 1};$$

Transfer function for **negative** polyphase signal

$$H_N(\omega) = \frac{V_{out}}{V_{in}} = \frac{\left[ 1 + (-j)^3 b_1 j\omega \right] \left[ 1 + (-j)^3 b_2 j\omega \right]}{a_0 (j\omega)^2 + a_1 j\omega + 1};$$

Here:  $b_0 = R_1 C_1; b_1 = R_2 C_2; a_0 = b_0 b_1; a_1 = b_0 + b_1 + 2 R_2 C_1;$

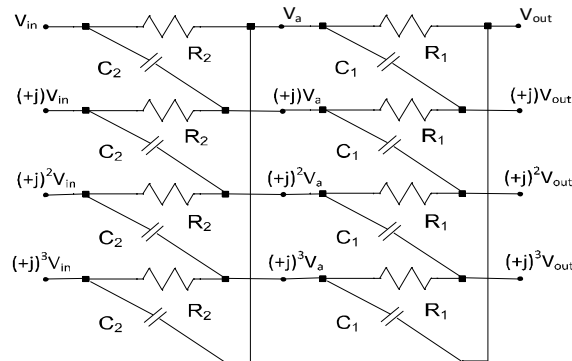
Image rejection ratio (IRR)

$$IRR(\omega) = \frac{|H_P(\omega)|}{|H_N(\omega)|} = \frac{|(1 + b_1 \omega)(1 + b_2 \omega)|}{|(1 - b_1 \omega)(1 - b_2 \omega)|};$$

# 1. Research Background

## Behaviors of 2<sup>nd</sup>-Order Polyphase Filter

2-order RC polyphase filter

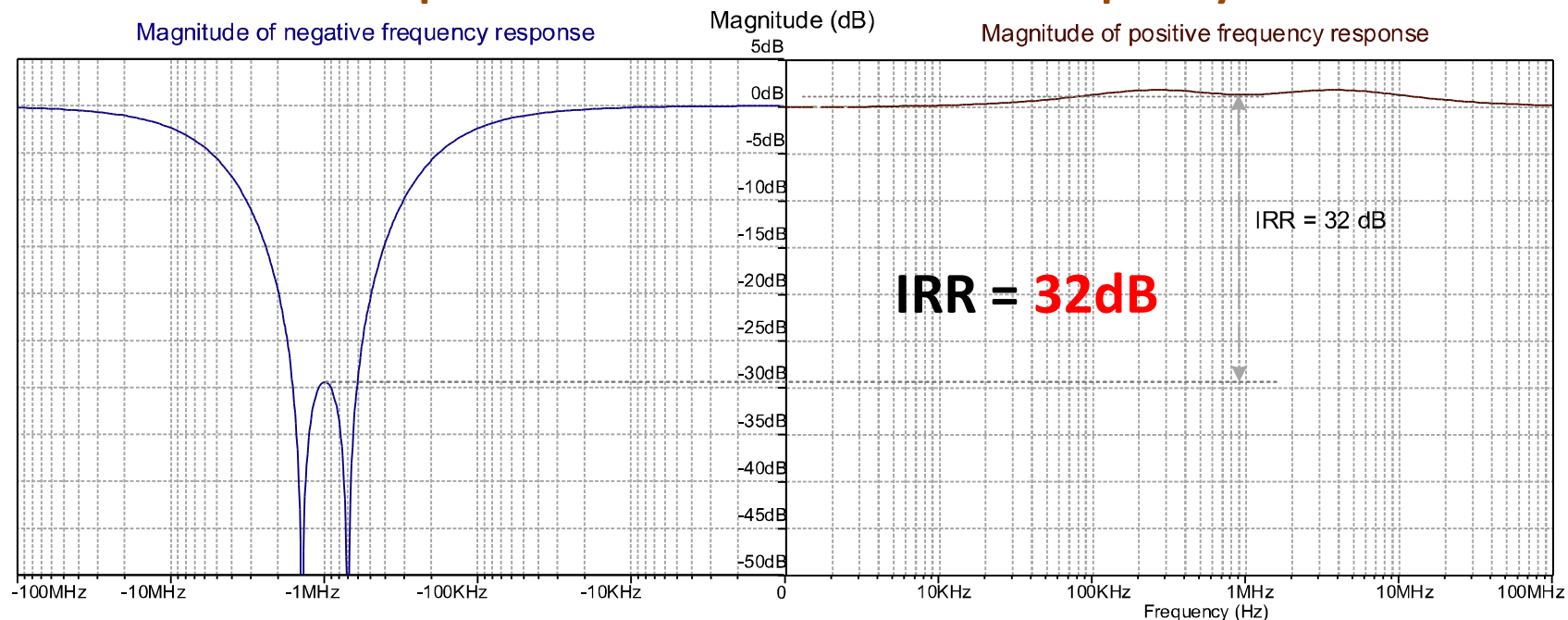


Transfer function in all frequency domain

$$|H(\omega)| = \frac{(1 + b_1\omega)(1 + b_2\omega)}{\sqrt{(1 - a_0\omega^2)^2 + (a_1\omega)^2}}; \omega \in R$$

Here, R1 = 1 kΩ, C1 = 227 pF, R2 = 1 kΩ, C2 = 114 pF, at f<sub>1</sub> = 700 kHz, f<sub>2</sub> = 1.4 MHz,

Bode plot of transfer function in all frequency domain



# 1. Research Background

## Behavior of 4<sup>th</sup>-order Complex Filter

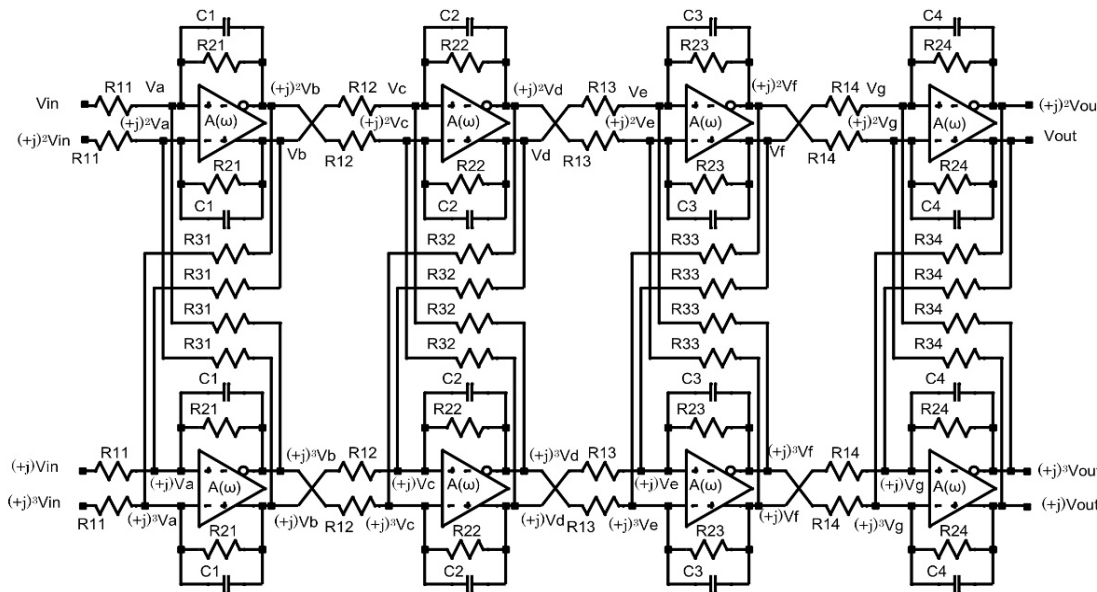
Transfer function  
for **positive**  
polyphase signals

$$H_P(\omega) = \frac{\frac{R_{21}}{R_{11}}}{\left[1 + j\left(\frac{\omega}{\omega_{cut1}} + \frac{R_{21}}{R_{31}}\right)\right]} \frac{\frac{R_{22}}{R_{12}}}{\left[1 + j\left(\frac{\omega}{\omega_{cut2}} + \frac{R_{22}}{R_{32}}\right)\right]} \frac{\frac{R_{23}}{R_{13}}}{\left[1 + j\left(\frac{\omega}{\omega_{cut3}} + \frac{R_{23}}{R_{33}}\right)\right]} \frac{\frac{R_{24}}{R_{14}}}{\left[1 + j\left(\frac{\omega}{\omega_{cut4}} + \frac{R_{24}}{R_{34}}\right)\right]}$$

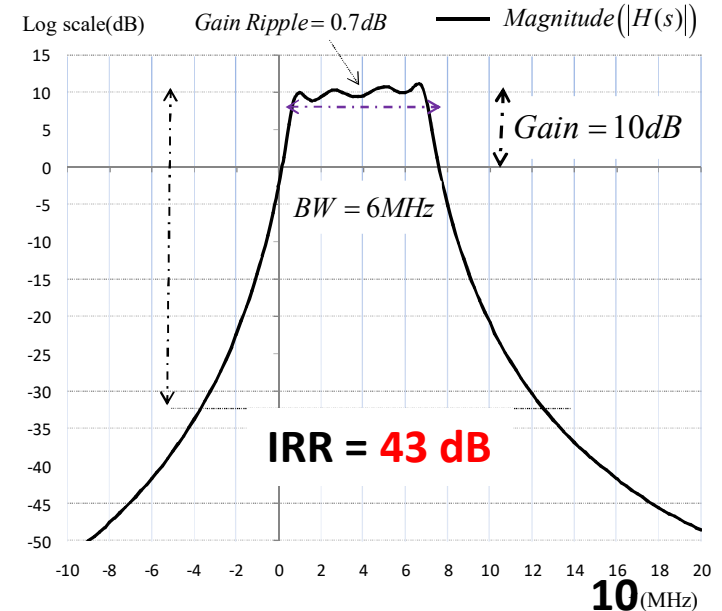
Transfer function  
for **negative**  
polyphase signals

$$H_N(\omega) = \frac{\frac{R_{21}}{R_{11}}}{\left[1 + j\left(\frac{\omega}{\omega_{cut1}} - \frac{R_{21}}{R_{31}}\right)\right]} \frac{\frac{R_{22}}{R_{12}}}{\left[1 + j\left(\frac{\omega}{\omega_{cut2}} - \frac{R_{22}}{R_{32}}\right)\right]} \frac{\frac{R_{23}}{R_{13}}}{\left[1 + j\left(\frac{\omega}{\omega_{cut3}} - \frac{R_{23}}{R_{33}}\right)\right]} \frac{\frac{R_{24}}{R_{14}}}{\left[1 + j\left(\frac{\omega}{\omega_{cut4}} - \frac{R_{24}}{R_{34}}\right)\right]}$$

### 4<sup>th</sup>-order complex filter



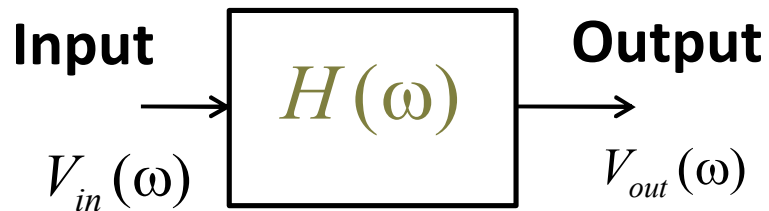
### Bode plot of transfer function



# 1. Research Background

## Self-loop Function in A Transfer Function

### Linear system



### Model of a linear system

$$H(\omega) = \frac{b_0(j\omega)^n + \dots + b_{n-1}(j\omega) + b_n}{a_0(j\omega)^n + \dots + a_{n-1}(j\omega) + a_n}$$

### Transfer function

$$H(\omega) = \frac{V_{out}(\omega)}{V_{in}(\omega)} = \frac{A(\omega)}{1 + L(\omega)}$$

$A(\omega)$  : Open loop function

$H(\omega)$  : Transfer function

$L(\omega)$  : Self-loop function

Variable: angular frequency ( $\omega$ )

○ Polar chart → Nyquist chart

○ Magnitude-frequency plot

○ Angular-frequency plot

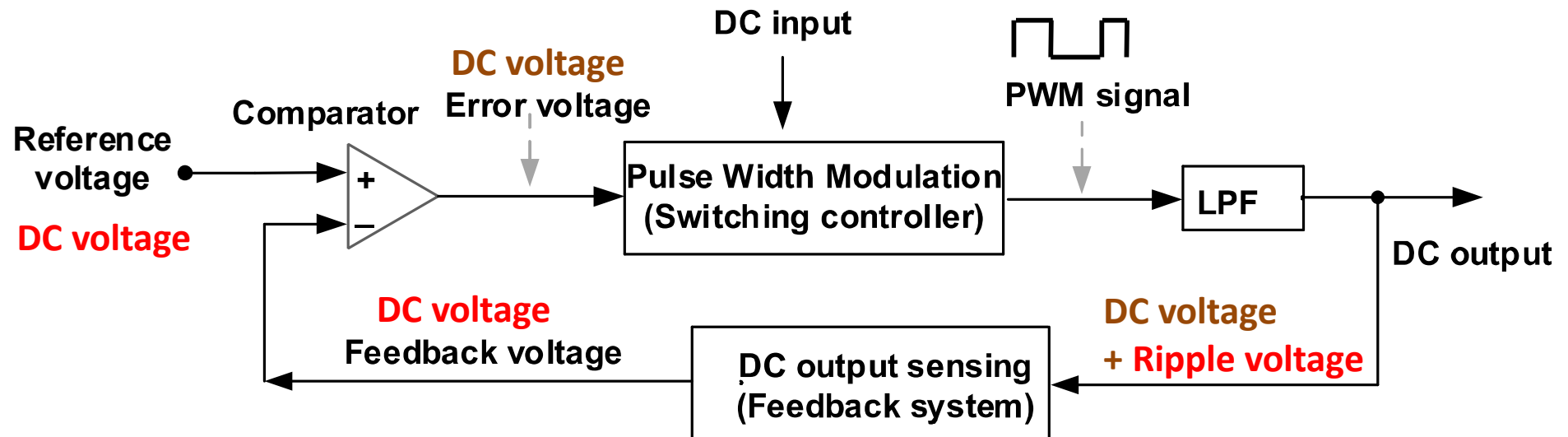
○ Magnitude-angular diagram → Nichols diagram

Bode plots

# 1. Research Background

## Characteristics of Adaptive Feedback Network

Block diagram of a typical adaptive feedback system



**Adaptive feedback** is used to control the output source along with the decision source (**DC-DC Buck converter**).

Transfer function of an **adaptive feedback network** is **significantly different from** transfer function of a **linear negative feedback network**.

→ **Loop gain is independent** of frequency variable (**referent voltage, feedback voltage, and error voltage are DC voltages**).

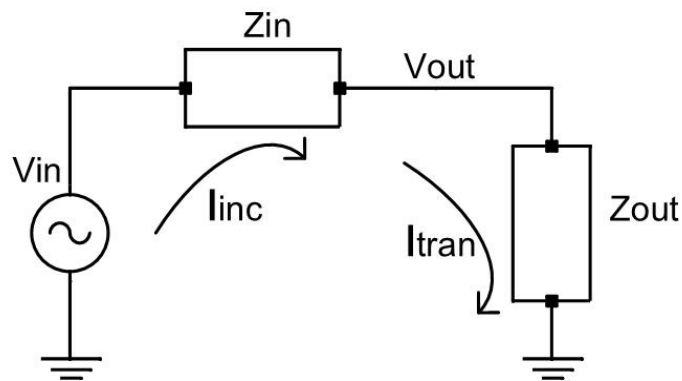
# 1. Research Background

## Alternating Current Conservation

### Transfer function

$$H(\omega) = \frac{V_{out}(\omega)}{V_{in}(\omega)} = \frac{1}{1 + \frac{Z_{in}}{Z_{out}}}$$

$$\Rightarrow L(\omega) = \frac{Z_{in}}{Z_{out}};$$



Simplified linear system

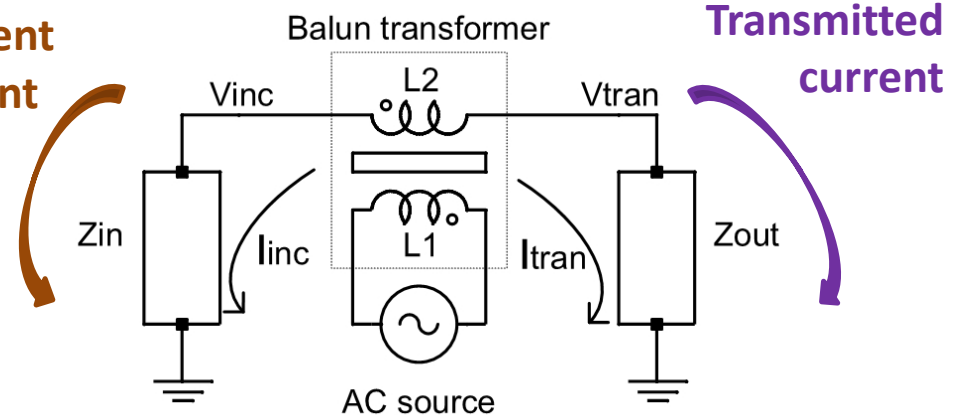
### Self-loop function

$$\frac{V_{inc}}{Z_{in}} = -\frac{V_{trans}}{Z_{out}} \Rightarrow L(\omega) = -\frac{V_{inc}}{V_{trans}} = \frac{Z_{in}}{Z_{out}}$$



10 mH  
inductance

Incident  
current



Derivation of self-loop function

# 1. Research Background

## Limitations of Conventional Methods

- **Middlebrook's measurement of loop gain**
  - Applying only in feedback systems (**DC-DC converters**).
- **Replica measurement of loop gain**
  - Using two identical networks (**not real measurement**).
- **Nyquist's stability condition**
  - Theoretical analysis for feedback systems (**Lab tool**).
- **Nichols chart of loop gain**
  - Only used in feedback control theory (**Lab tool**).
- **Conventional superposition**
  - Solving for every source (**several times**).

# Outline

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- Motivation, objectives and achievements
- Self-loop function in a transfer function

## 2. Ringing Test for Passive Networks

- **Stability test for RLC low-pass filters**

## 3. Ringing Test for Unity-Gain Amplifiers

- Stability test for unity-gain amplifiers

## 4. Ringing Test for Ladder Low-Pass Filters

- Stability test for active 3<sup>rd</sup>-order ladder low-pass filters

## 5. Conclusions



## 2. Ringing Test for Passive Networks

### Characteristics of 2<sup>nd</sup>-order Transfer Function

Second-order transfer function: 
$$H(\omega) = \frac{1}{1 + a_0(j\omega)^2 + a_1j\omega}$$

Case	Over-damping	Critical damping	Under-damping
<b>Delta</b> ( $\Delta$ )	$\frac{1}{a_0} < \left(\frac{a_1}{2a_0}\right)^2 \Rightarrow \Delta = a_1^2 - 4a_0 > 0$	$\frac{1}{a_0} = \left(\frac{a_1}{2a_0}\right)^2 \Rightarrow \Delta = a_1^2 - 4a_0 = 0$	$\frac{1}{a_0} > \left(\frac{a_1}{2a_0}\right)^2 \Rightarrow \Delta = a_1^2 - 4a_0 < 0$
<b>Module</b> $ H(\omega) $	$\frac{1}{a_0} \sqrt{\omega^2 + \left(\frac{a_1}{2a_0} - \sqrt{\left(\frac{a_1}{2a_0}\right)^2 - \frac{1}{a_0}}\right)^2} \sqrt{\omega^2 + \left(\frac{a_1}{2a_0} + \sqrt{\left(\frac{a_1}{2a_0}\right)^2 - \frac{1}{a_0}}\right)^2}$	$\frac{1}{a_0} \sqrt{\omega^2 + \left(\frac{a_1}{2a_0}\right)^2} = \frac{1}{2} = -6dB$	$\frac{1}{a_0} \sqrt{\left(\omega - \sqrt{\frac{1}{a_0} - \left(\frac{a_1}{2a_0}\right)^2}\right)^2 + \left(\frac{a_1}{2a_0}\right)^2} \sqrt{\left(\omega + \sqrt{\frac{1}{a_0} - \left(\frac{a_1}{2a_0}\right)^2}\right)^2 + \left(\frac{a_1}{2a_0}\right)^2}$
<b>Angular</b> $\theta(\omega)$	$-\arctan\left(\frac{\omega}{\frac{a_1}{2a_0} - \sqrt{\left(\frac{a_1}{2a_0}\right)^2 - \frac{1}{a_0}}}\right) - \arctan\left(\frac{\omega}{\frac{a_1}{2a_0} + \sqrt{\left(\frac{a_1}{2a_0}\right)^2 - \frac{1}{a_0}}}\right)$	$-2 \arctan\left(\frac{2a_0\omega}{a_1}\right)$	$-\arctan\left(\frac{\omega - \sqrt{\frac{1}{a_0} - \left(\frac{a_1}{2a_0}\right)^2}}{\frac{a_1}{2a_0}}\right) - \arctan\left(\frac{\omega + \sqrt{\frac{1}{a_0} - \left(\frac{a_1}{2a_0}\right)^2}}{\frac{a_1}{2a_0}}\right)$
$\omega_{cut} = \frac{a_1}{2a_0}$	$ H(\omega_{cut})  < \frac{2a_0}{a_1}$ $\theta(\omega_{cut}) > -\frac{\pi}{2}$	$ H(\omega_{cut})  = \frac{2a_0}{a_1}$ $\theta(\omega_{cut}) = -\frac{\pi}{2}$	$ H(\omega_{cut})  > \frac{2a_0}{a_1}$ $\theta(\omega_{cut}) < -\frac{\pi}{2}$

## 2. Ringing Test for Passive Networks

### Characteristics of 2<sup>nd</sup>-order Self-loop Function

Second-order self-loop function:  $L(\omega) = j\omega[a_0 j\omega + a_1]$

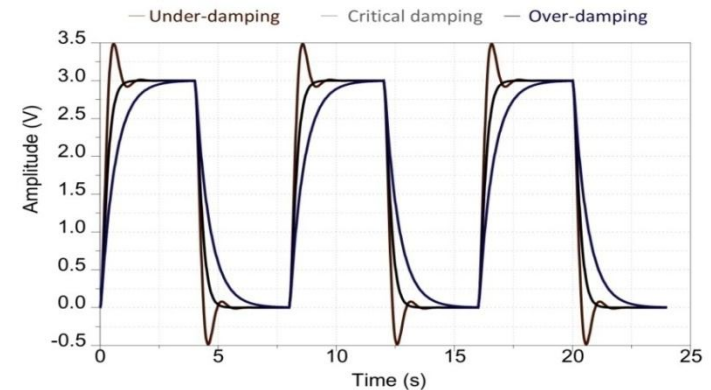
Case	Over-damping	Critical damping	Under-damping			
<b>Delta (<math>\Delta</math>)</b>	$\Delta = a_1^2 - 4a_0 > 0$	$\Delta = a_1^2 - 4a_0 = 0$	$\Delta = a_1^2 - 4a_0 < 0$			
$ L(\omega) $	$\omega\sqrt{(a_0\omega)^2 + a_1^2}$	$\omega\sqrt{(a_0\omega)^2 + a_1^2}$	$\omega\sqrt{(a_0\omega)^2 + a_1^2}$			
$\theta(\omega)$	$\frac{\pi}{2} + \arctan \frac{a_0\omega}{a_1}$	$\frac{\pi}{2} + \arctan \frac{a_0\omega}{a_1}$	$\frac{\pi}{2} + \arctan \frac{a_0\omega}{a_1}$			
$\omega_1 = \frac{a_1}{2a_0}\sqrt{\sqrt{5}-2}$	$ L(\omega_1)  > 1$	$\pi - \theta(\omega_1) > 76.3^\circ$	$ L(\omega_1)  < 1$	$\pi - \theta(\omega_1) < 76.3^\circ$		
$\omega_2 = \frac{a_1}{2a_0}$	$ L(\omega_2)  > \sqrt{5}$	$\pi - \theta(\omega_2) > 63.4^\circ$	$ L(\omega_2)  = \sqrt{5}$	$\pi - \theta(\omega_2) = 63.4^\circ$	$ L(\omega_2)  < \sqrt{5}$	$\pi - \theta(\omega_2) < 63.4^\circ$
$\omega_3 = \frac{a_1}{a_0}$	$ L(\omega_3)  > 4\sqrt{2}$	$\pi - \theta(\omega_3) > 45^\circ$	$ L(\omega_3)  = 4\sqrt{2}$	$\pi - \theta(\omega_3) = 45^\circ$	$ L(\omega_3)  < 4\sqrt{2}$	$\pi - \theta(\omega_3) < 45^\circ$

# 2. Ringing Test for Passive Networks

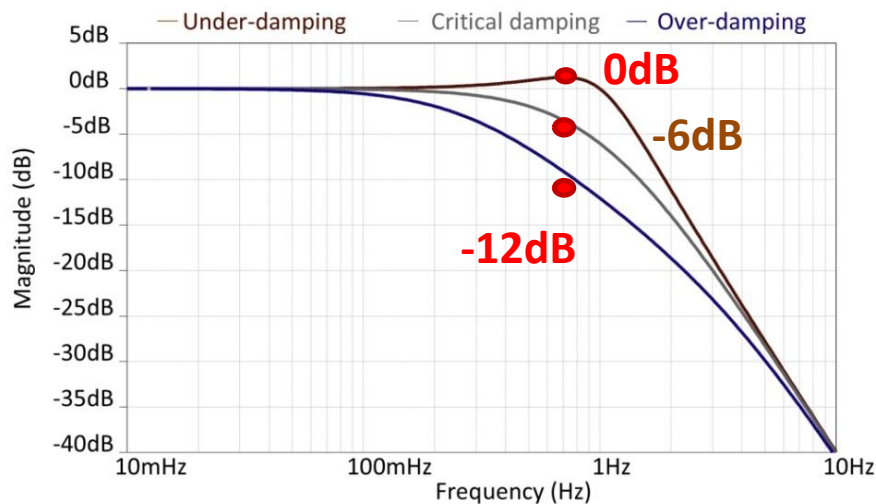
## Operating Regions of 2<sup>nd</sup>-Order System

- **Under-damping:**  $H_1(\omega) = \frac{1}{(j\omega)^2 + j\omega + 1}$ ;  
 $L_1(\omega) = (j\omega)^2 + j\omega$ ;
- **Critical damping:**  $H_2(\omega) = \frac{1}{(j\omega)^2 + 2j\omega + 1}$ ;  
 $L_2(\omega) = (j\omega)^2 + 2j\omega$ ;
- **Over-damping:**  $H_3(\omega) = \frac{1}{(j\omega)^2 + 3j\omega + 1}$ ;  
 $L_3(\omega) = (j\omega)^2 + 3j\omega$ ;

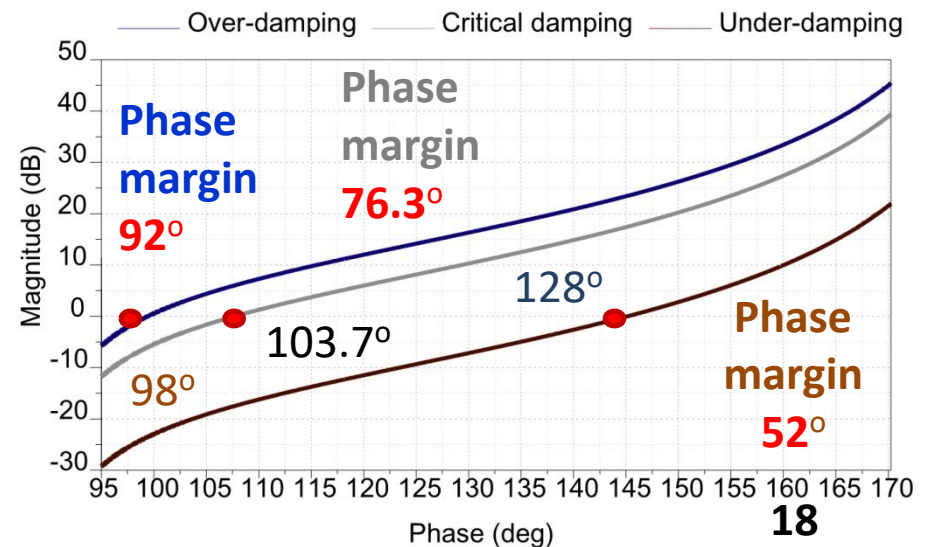
Transient response



Bode plot of transfer function



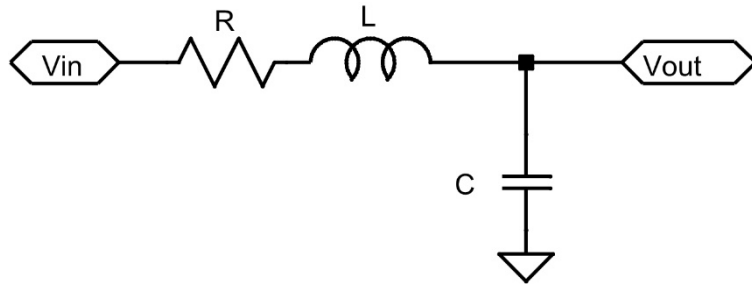
Nichols plot of self-loop function



## 2. Ringing Test for Passive Networks

### Stability Test for Passive 2<sup>rd</sup>-Order RLC LPF

Passive RLC Low-pass Filter



Transfer function

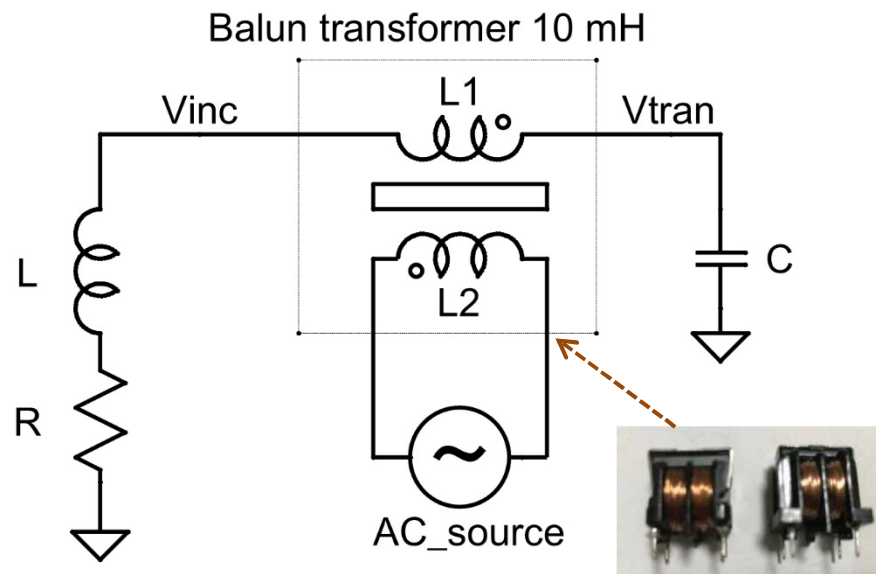
$$H(\omega) = \frac{V_{out}}{V_{in}} = \frac{1}{a_0 (j\omega)^2 + a_1 j\omega + 1};$$

Self-loop function

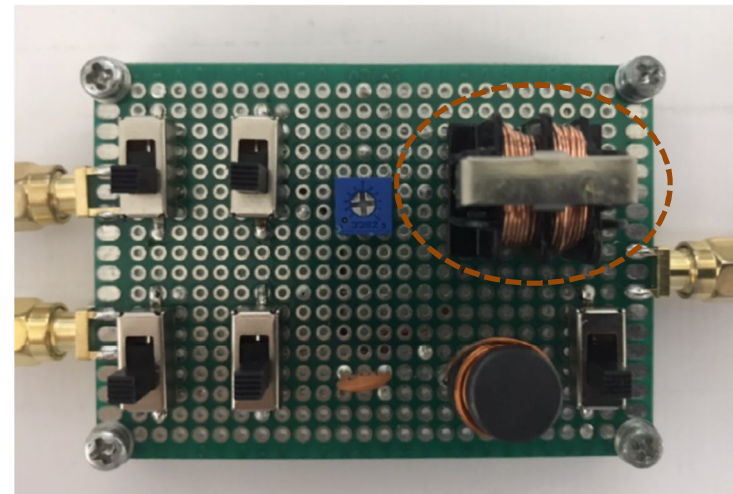
$$L(\omega) = a_0 (j\omega)^2 + a_1 j\omega;$$

where,  $a_0 = LC$ ;  $a_1 = RC$ ;

Derivation of self-loop function



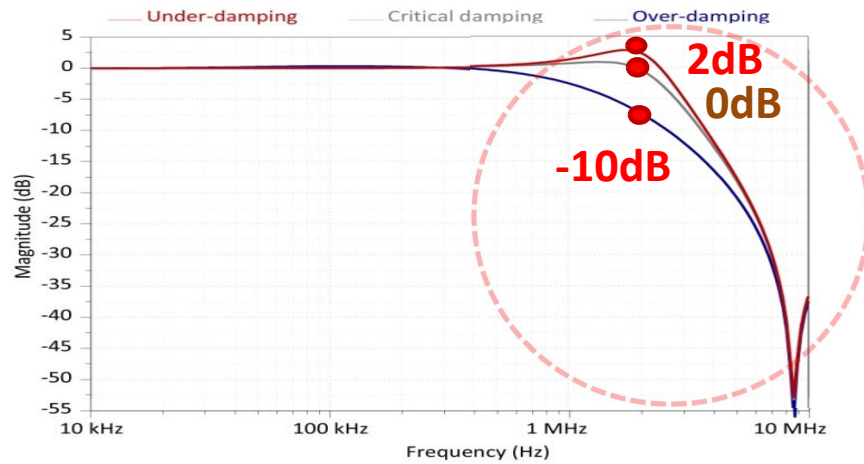
Implemented circuit



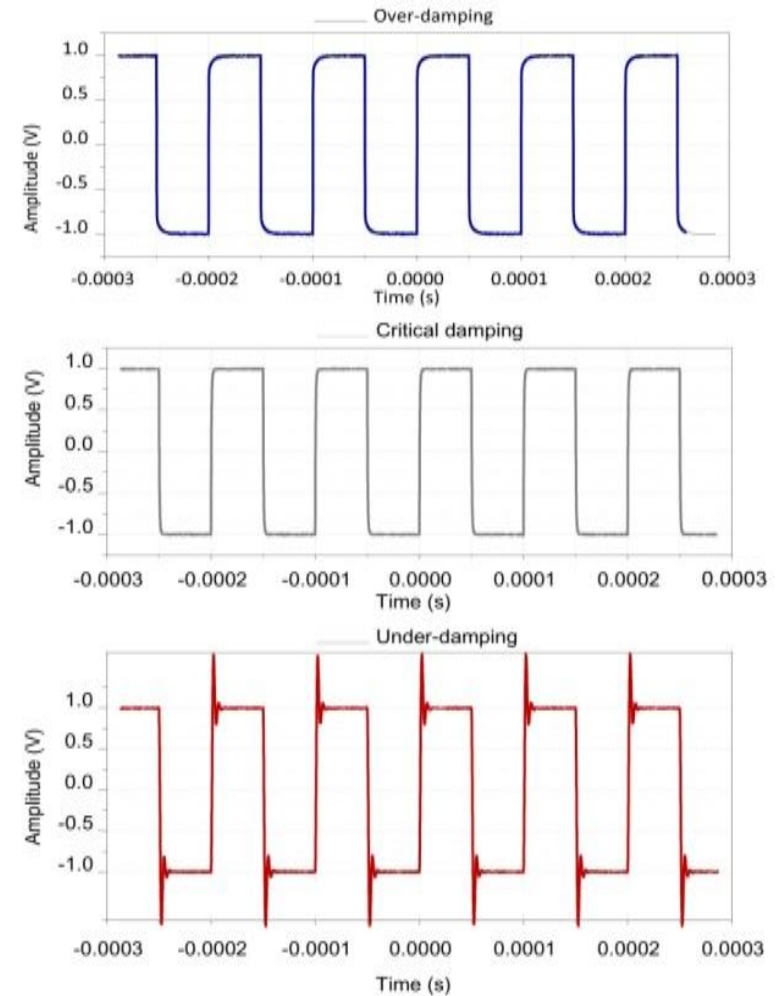
# 2. Ringing Test for Passive Networks

## Stability Test for 2<sup>rd</sup>-Order Passive RLC LPF

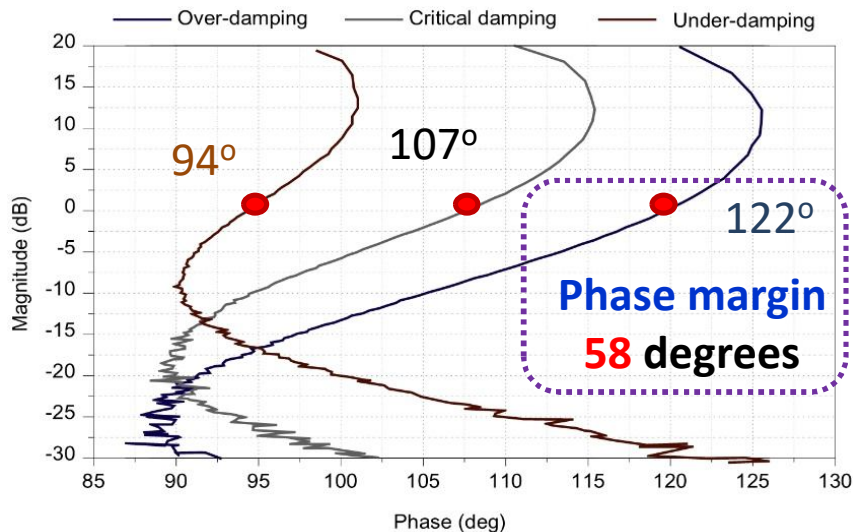
**Bode plot** of transfer function



**Transient responses**



**Nichols plot** of self-loop function



# Outline

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## 1. Research Background

- Motivation, objectives and achievements
- Self-loop function in a transfer function

## 2. Ringing Test for Passive Networks

- Stability test for RLC low-pass filters

## **3. Ringing Test for Unity-Gain Amplifiers**

- **Stability test for unity-gain amplifiers**

## 4. Ringing Test for Ladder Low-Pass Filters

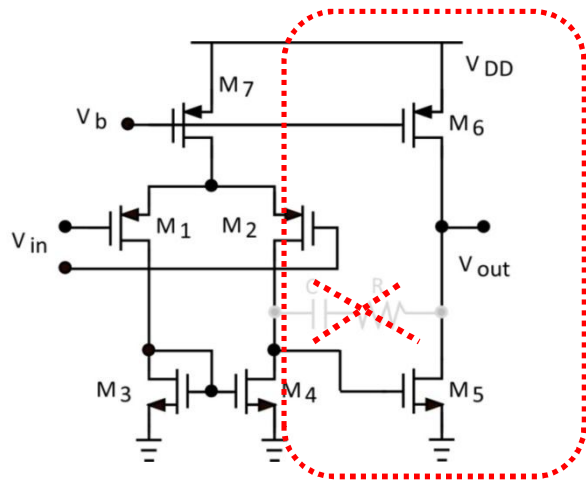
- Stability test for active 3<sup>rd</sup>-order ladder low-pass filters

## 5. Conclusions

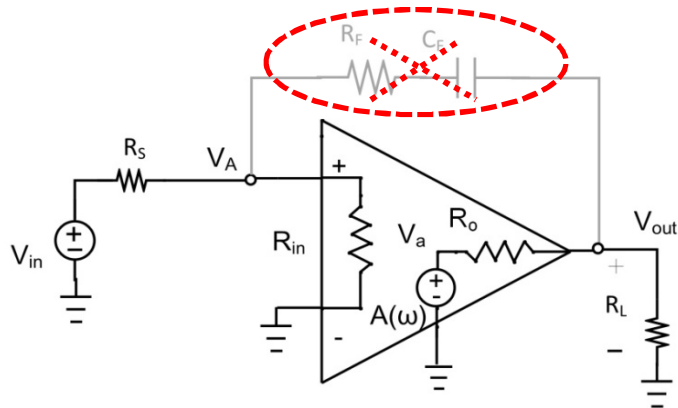
# 3. Ringing Test for Unity-Gain Amplifiers

## Analysis of Op Amp without Miller's Capacitor

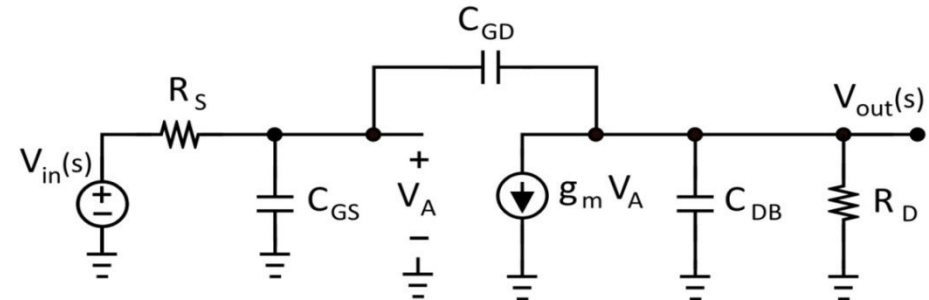
**Without** frequency compensation



**Simplified model**



**Small signal model**



**Transfer function  $H(\omega)$  and self-loop function  $L(\omega)$**

$$H(\omega) = \frac{b_0 j\omega + b_1}{a_0 (j\omega)^2 + a_1 j\omega + 1};$$

$$L(\omega) = a_0 (j\omega)^2 + a_1 j\omega$$

**Where,**

$$b_0 = R_D R_S \left[ (C_{GD} + C_{DB})(C_{GS} + C_{GD}) - C_{GD}^2 \right]$$

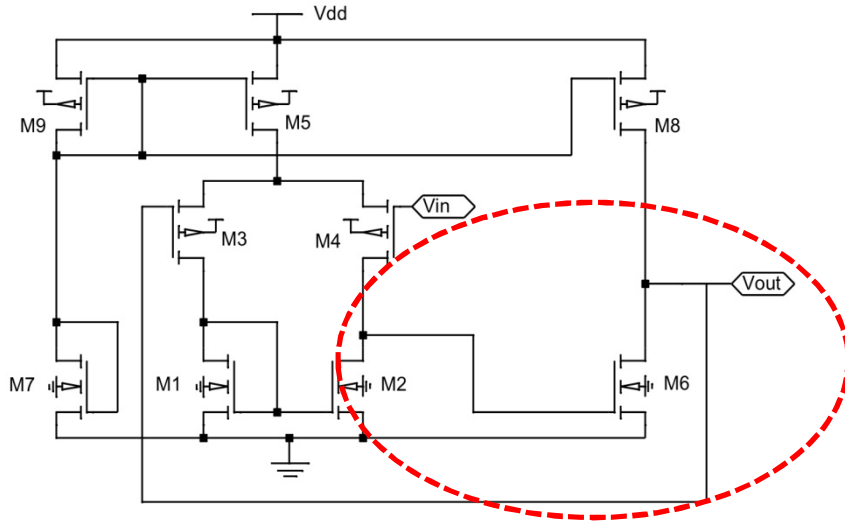
$$b_1 = \left[ R_D (C_{GD} + C_{DB}) + R_S (C_{GS} + C_{GD}) + R_D R_S g_m C_{GD} \right]$$

$$a_0 = R_D C_{GD}; \quad a_1 = -R_D g_m;$$

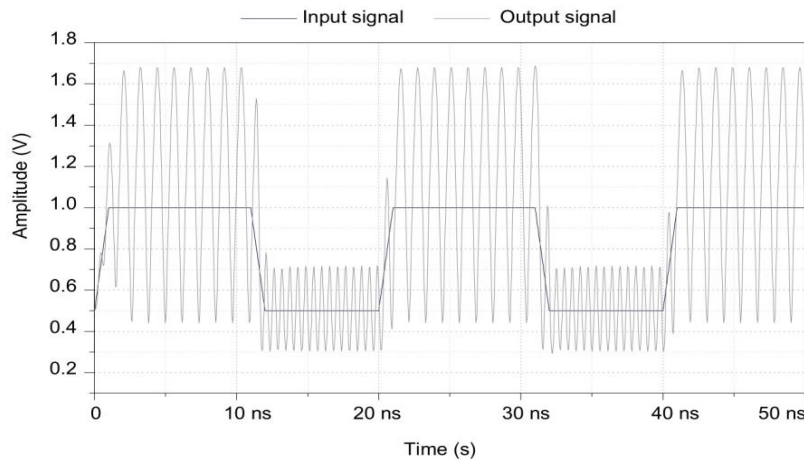
# 3. Ringing Test for Unity-Gain Amplifiers

## Unity-Gain Amplifier without Miller's Capacitor

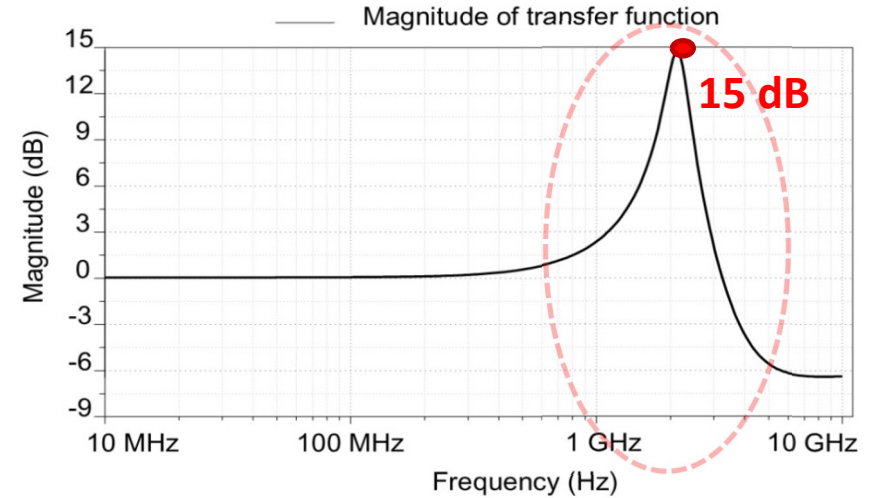
### Unity-Gain Amplifier



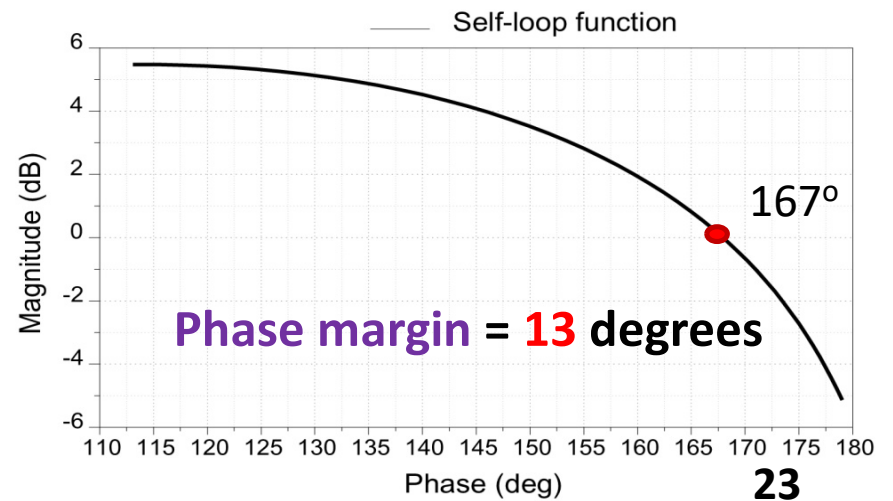
### Transient response



### Bode plot of transfer function $H(\omega)$



### Nichols plot of self-loop function $L(\omega)$

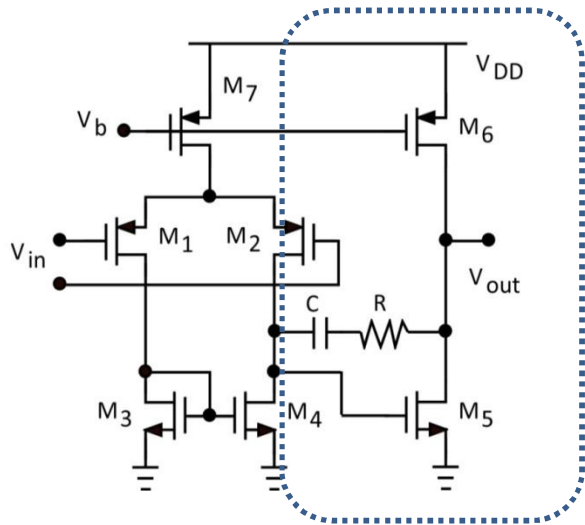




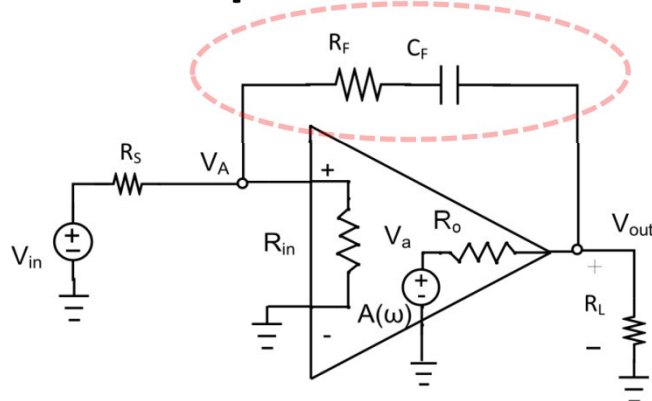
# 3. Ringing Test for Unity-Gain Amplifiers

## Two-stage Op Amp with Frequency Compensation

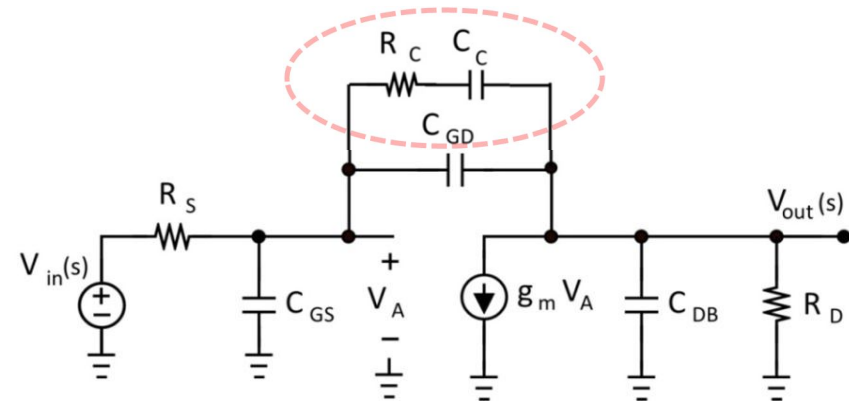
With Miller's capacitor and resistor



Simplified model



Small signal model



Transfer function  $H(\omega)$

$$H(\omega) = \frac{b_0 (j\omega)^3 + b_1 (j\omega)^2 + b_2 j\omega + b_3}{a_0 (j\omega)^4 + a_1 (j\omega)^3 + a_2 (j\omega)^2 + a_3 j\omega + 1};$$

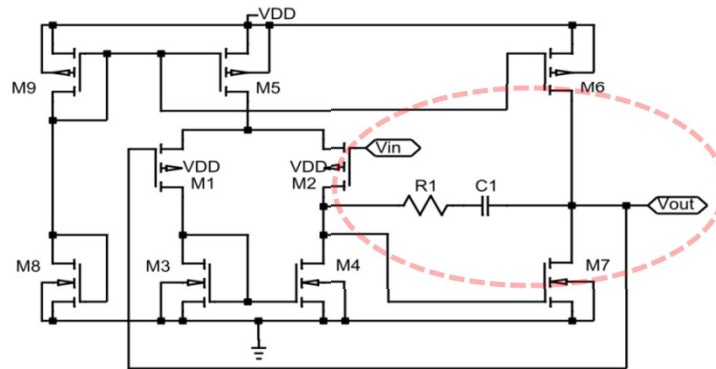
Self-loop function  $L(\omega)$

$$L(\omega) = a_0 (j\omega)^4 + a_1 (j\omega)^3 + a_2 (j\omega)^2 + a_3 j\omega$$

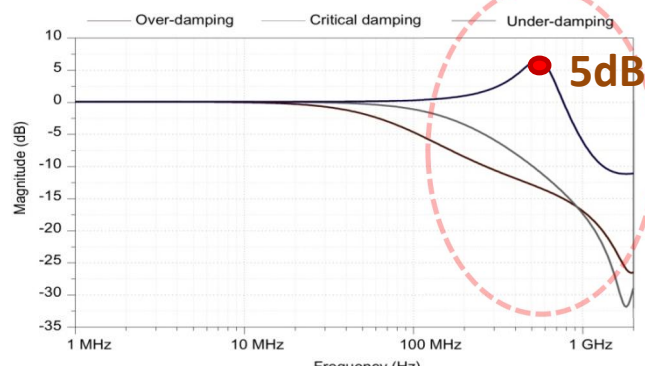
# 3. Ringing Test for Unity-Gain Amplifiers

## Stability Test for Op Amp with Miller's Capacitor

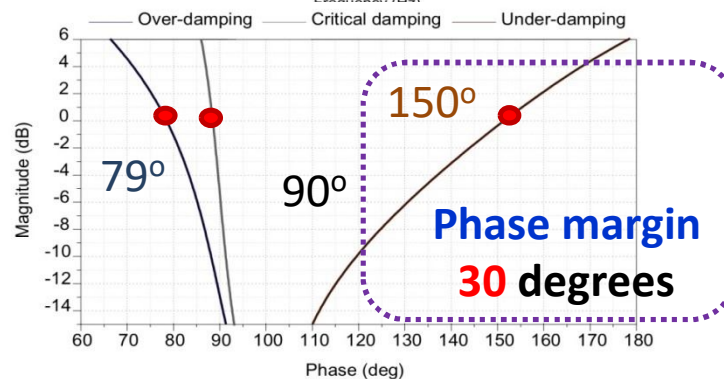
Unity-gain amplifier with Miller's capacitor



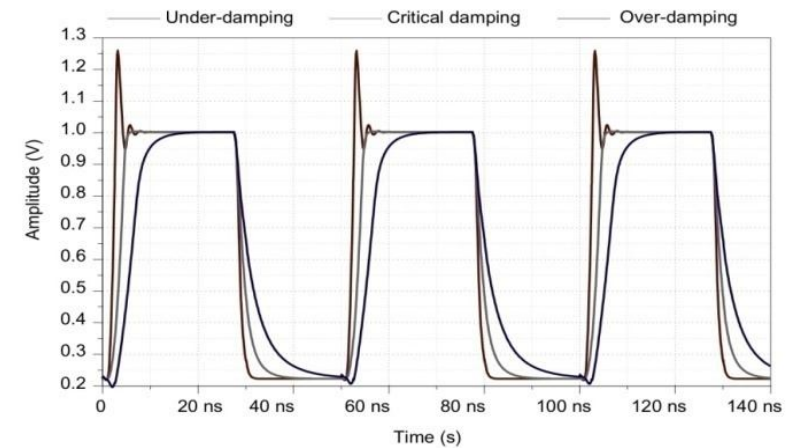
Bode plot of transfer function



Nichols plot of self-loop function



Simulated transient response



Operating regions

**Under-damping:**

**$R1 = 2 \text{ k}\Omega$ ,  $C1 = 1 \text{ pF}$**

**Critical damping:**

**$R1 = 3.5 \text{ k}\Omega$ ,  $C1 = 0.2 \text{ pF}$**

**Over-damping:**

**$R1 = 3.5 \text{ k}\Omega$ ,  $C1 = 0.8 \text{ pF}$**

# Outline

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## 1. Research Background

- Motivation, objectives and achievements
- Self-loop function in a transfer function

## 2. Ringing Test for Passive Networks

- Stability test for RLC low-pass filters

## 3. Ringing Test for Unity-Gain Amplifiers

- Stability test for unity-gain amplifiers

## 4. Ringing Test for Ladder Low-Pass Filters

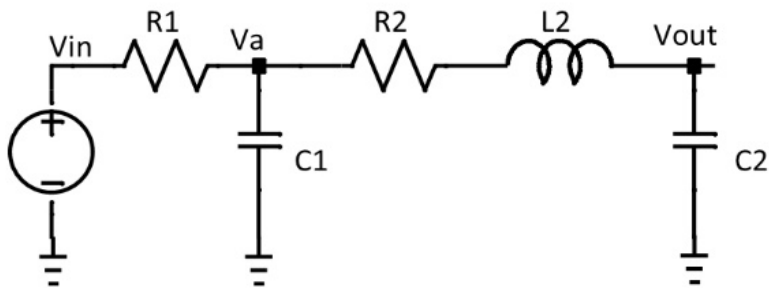
- **Stability test for active 3<sup>rd</sup>-order ladder low-pass filters**

## 5. Conclusions

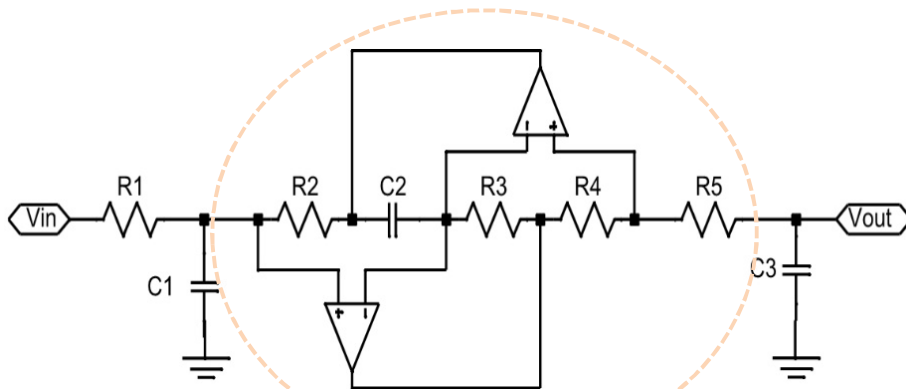
# 4. Ringing Test for Ladder Low-Pass Filters

## Analysis of Active 3<sup>rd</sup>-Order Ladder LPF

### Passive 3<sup>rd</sup>-order ladder LPF



### Active 3<sup>rd</sup>-order ladder LPF



General impedance converter

### Transfer function & self-loop function

$$H_{out}(\omega) = \frac{V_{out}}{V_{in}} = \frac{1}{a_0(j\omega)^3 + a_1(j\omega)^2 + a_2j\omega + 1};$$

$$L(\omega) = j\omega [a_0(j\omega)^2 + a_1j\omega + a_2]$$

where,  $b_0 = L_2C_2; b_1 = R_2C_2;$

$a_0 = R_1C_1L_2C_2; a_1 = R_1C_1R_2C_2 + L_2C_2;$

$a_2 = R_1(C_1 + C_2) + R_2C_2;$

**R1 = 100 Ω, R2 = 50 kΩ,**

**R3 = R4 = 50 kΩ, C1 = 5 nF, C2 = 10 nF,**

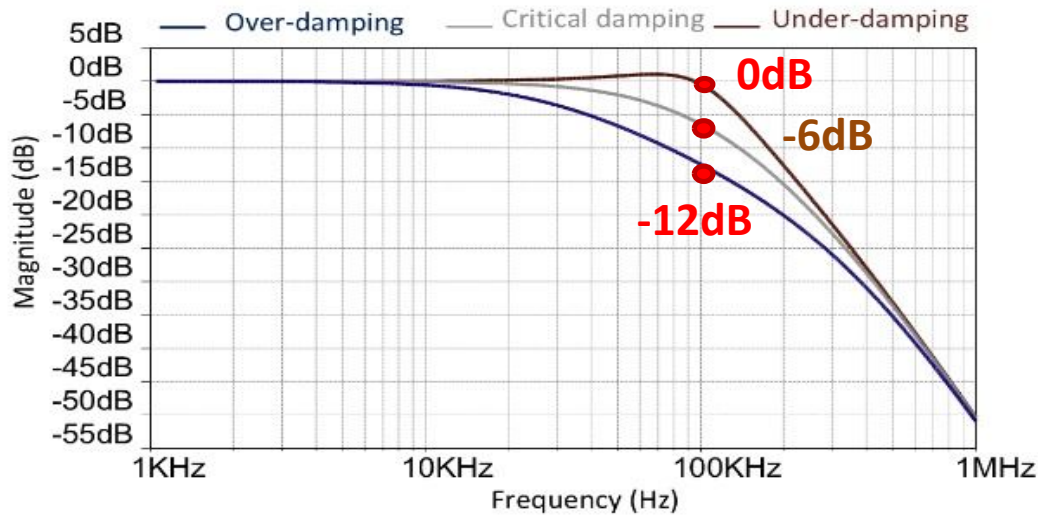
**C3 = 3.18 nF, at  $f_0 = 100$  kHz.**

- **Over-damping** (R5 = 0.5 kΩ),
- **Critical damping** (R5 = 1 kΩ), and
- **Under-damping** (R5 = 2 kΩ).

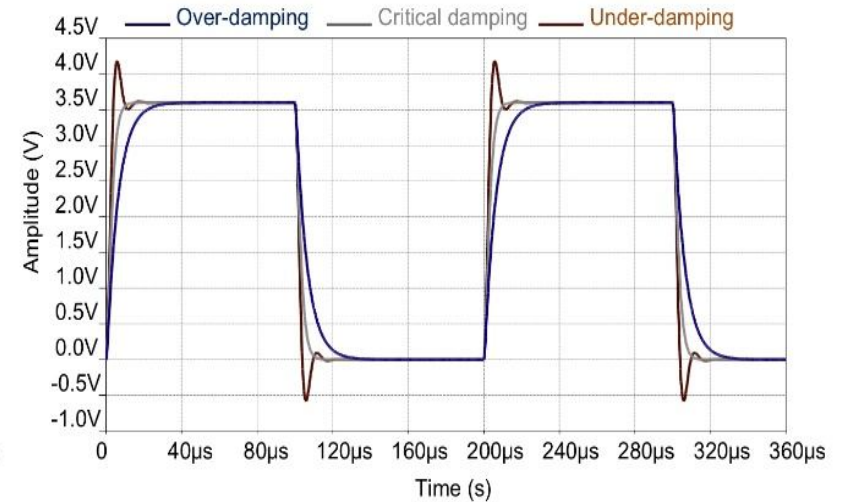
# 4. Ringing Test for Ladder Low-Pass Filters

## Simulation Results of 3<sup>rd</sup>-Order Ladder LPF

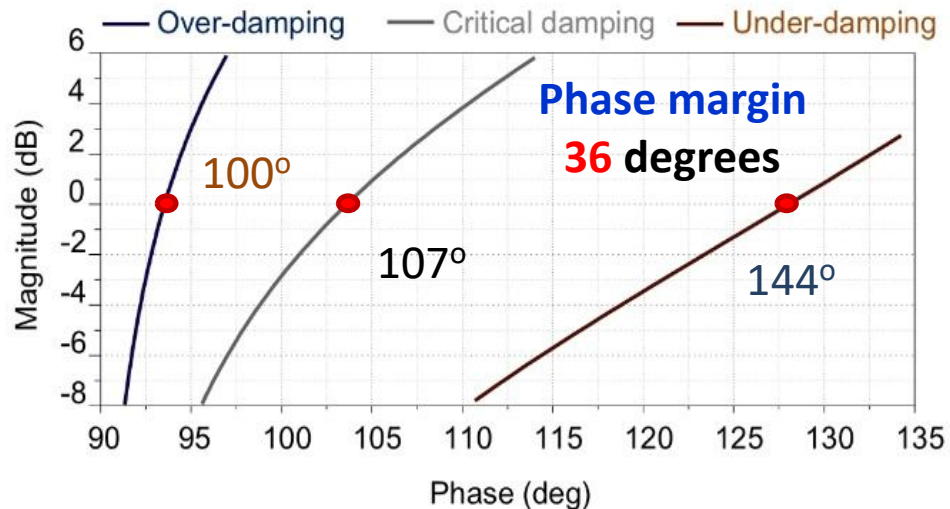
**Bode plot** of transfer function



**Transient response**



**Nichols plot** of self-loop function



**Over-damping:**

→ Phase margin is **80** degrees.

**Critical damping:**

→ Phase margin is **73** degrees.

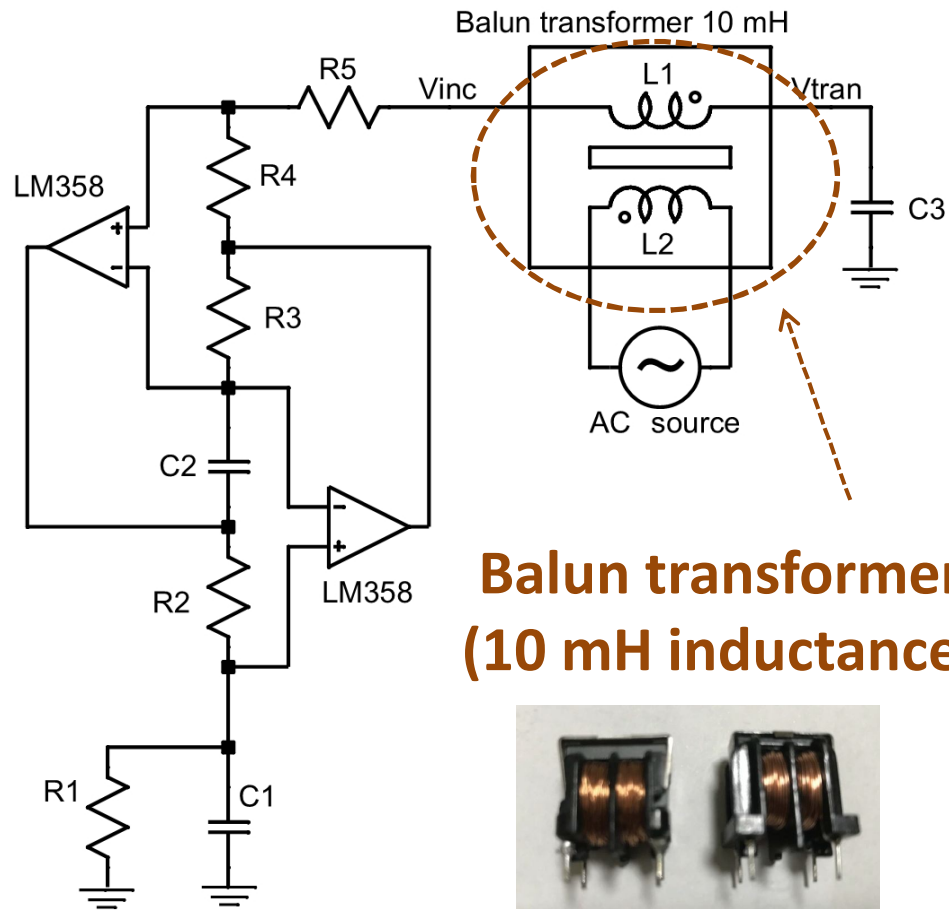
**Under-damping:**

→ Phase margin is **36** degrees.

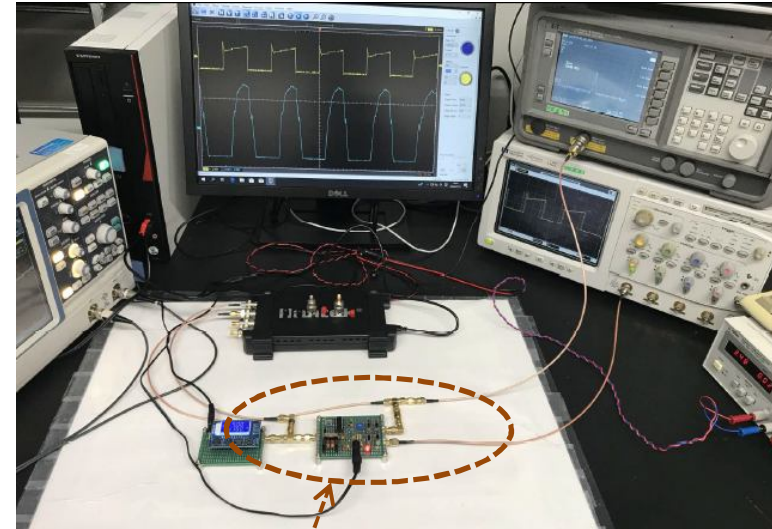
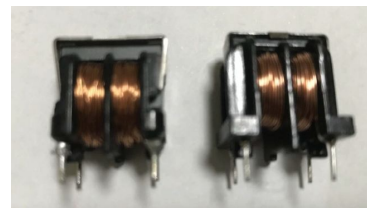
# 4. Ringing Test for Ladder Low-Pass Filters

## Implemented Circuit of 3<sup>rd</sup>-Order Ladder LPF

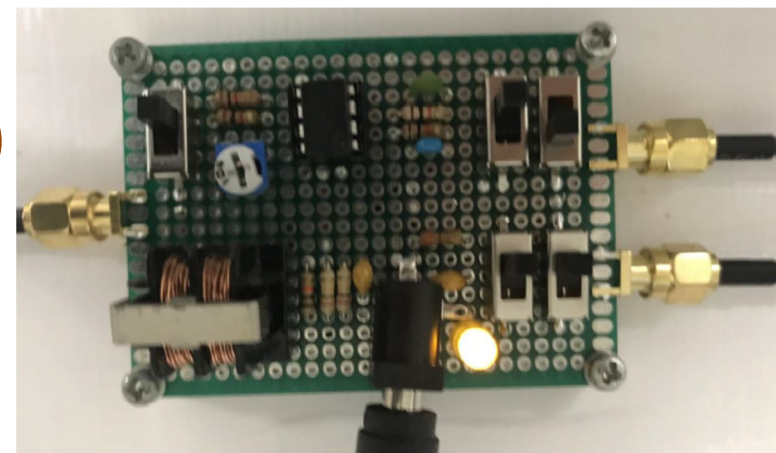
### Measurement of self-loop function



**Balun transformer  
(10 mH inductance)**



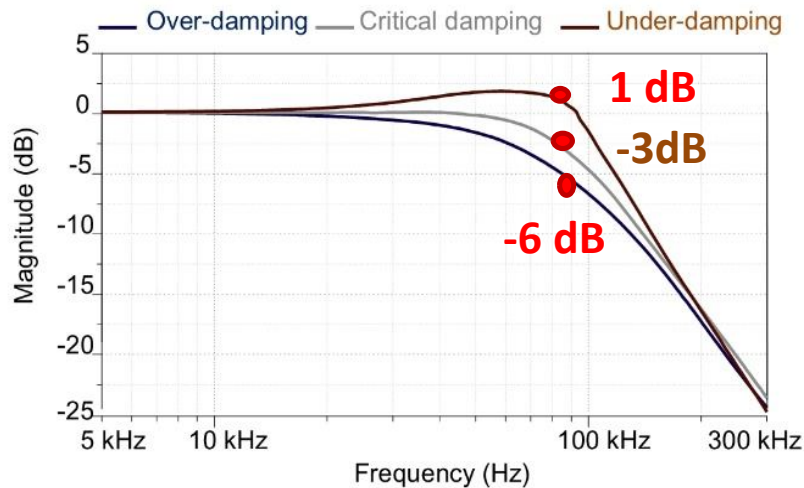
**Device under test**



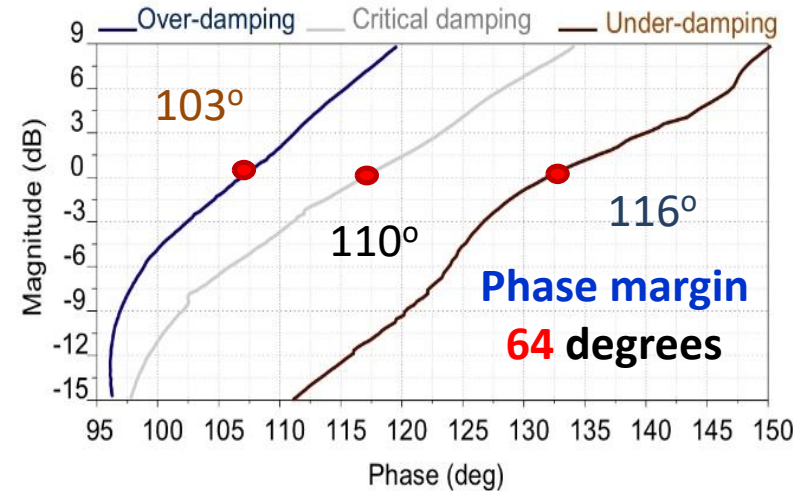
# 4. Ringing Test for Ladder Low-Pass Filters

## Measurement Results of 3<sup>rd</sup>-order Ladder LPF

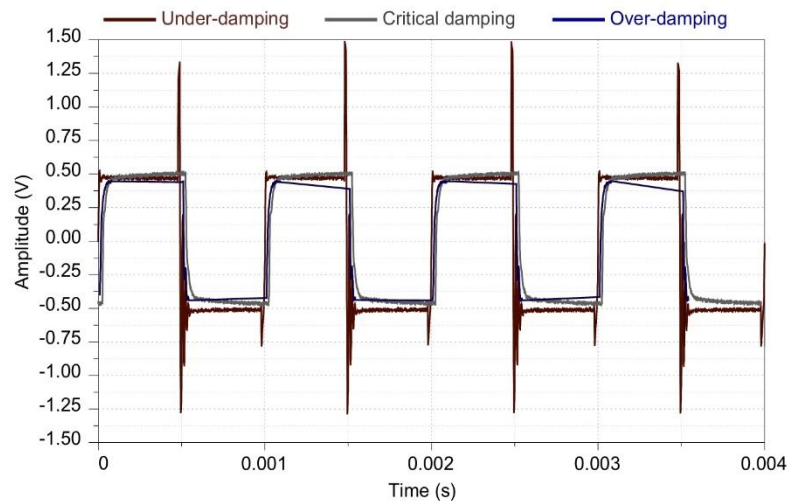
**Bode plot** of transfer function



**Nichols plot** of self-loop function



**Transient response**



**Over-damping:**

→ Phase margin is **77** degrees.

**Critical damping:**

→ Phase margin is **70** degrees.

**Under-damping:**

→ Phase margin is **64** degrees.

# Outline

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## 1. Research Background

- Motivation, objectives and achievements
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## 3. Ringing Test for Unity-Gain Amplifiers

- Stability test for unity-gain amplifiers

## 4. Ringing Test for Ladder Low-Pass Filters

- Stability test for active 3<sup>rd</sup>-order ladder low-pass filters

## 5. Conclusions



## 5. Comparison (Superposition formula)

<b>Features</b>	<b>Superposition formula</b>	<b>Conventional Superposition</b>	<b>Millan's theorem</b>
<b>Effects of all actuating sources</b>	<b>At one time</b>	<b>Several times</b>	<b>At one time</b>
<b>Transfer function accuracy</b>	<b>Yes</b>	<b>No</b>	<b>No</b>
<b>Single-input network analysis</b>	<b>Yes</b>	<b>Yes</b>	<b>Yes</b>
<b>Polyphase network analysis</b>	<b>Yes</b>	<b>No</b>	<b>No</b>
<b>Complex network analysis</b>	<b>Yes</b>	<b>No</b>	<b>No</b>
<b>Image rejection ratio accuracy</b>	<b>Yes</b>	<b>No</b>	<b>No</b>

## 5. Discussions (Superposition formula)

---

**Transfer function** and **image rejection ratio** give useful information about the behaviors of **polyphase filters** and **complex filters**.

**Fundamental network analysis theory** for **multi-source systems**:

- **Compute** the effects of all sources at one time,
- **Reduce** the wasteful time,
- **Decrease** the hand calculation times,
- **Get** the transfer function faster, and
- **Reduce** the network complexity.

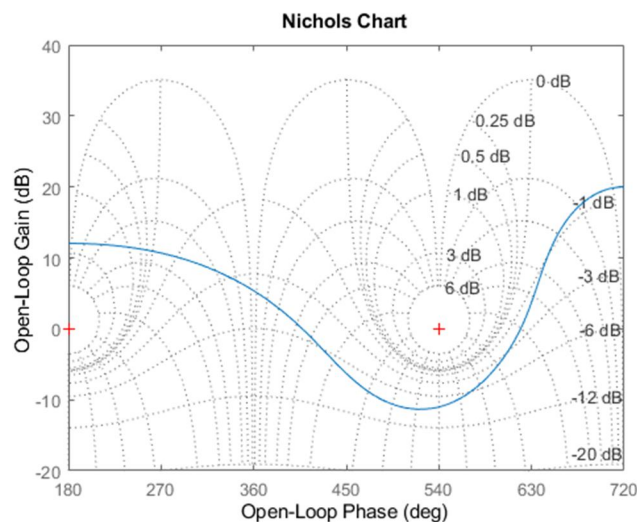
## 5. Comparison (Self-loop function)

<b>Features</b>	<b>Alternating current conservation</b>	<b>Replica measurement</b>	<b>Middlebrook's method</b>
<b>Main objective</b>	<b>Self-loop function</b>	<b>Loop gain</b>	<b>Loop gain</b>
<b>Transfer function accuracy</b>	<b>Yes</b>	<b>No</b>	<b>No</b>
<b>Ringing Test</b>	<b>Yes</b>	<b>Yes</b>	<b>Yes</b>
<b>Operating region accuracy</b>	<b>Yes</b>	<b>No</b>	<b>No</b>
<b>Phase margin accuracy</b>	<b>Yes</b>	<b>No</b>	<b>No</b>
<b>Passive networks</b>	<b>Yes</b>	<b>No</b>	<b>No</b>

# 5. Discussions (Self-loop function)

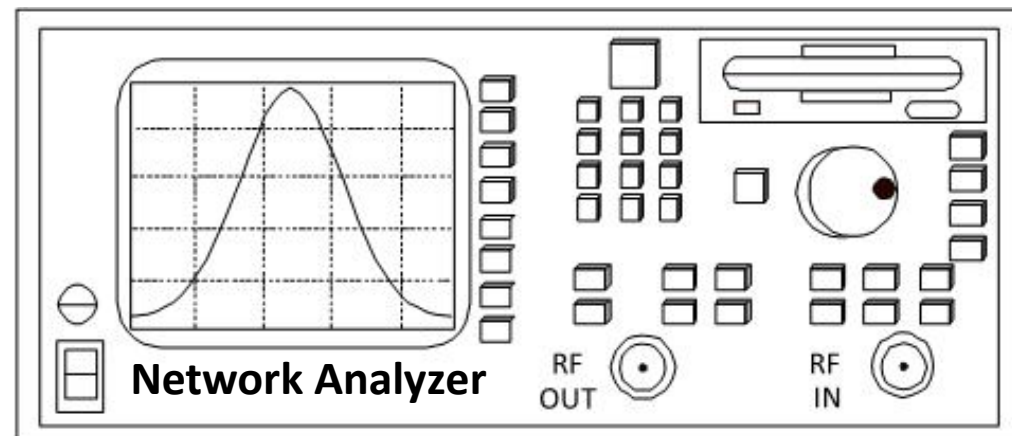
- Loop gain is **independent of** frequency variable.
- Loop gain in adaptive feedback network is **significantly different from** self-loop function in linear negative feedback network.

Nichols chart is **only used** in **MATLAB simulation**.



<https://www.mathworks.com/help/control/ref/nichols.html>

Nichols chart **isn't** used **widely** in practical measurements (**only used** in control theory).



➔ **(Technology limitations)**

# 5. Conclusions

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## This work:

- Proposal of alternating current conservation for deriving **self-loop function** in a transfer function  
→ **Observation of self-loop function** can help us **optimize the behavior** of a high-order system.
- Implementations of circuits and measurements of self-loop functions for **passive & active low-pass filters**  
→ **Theoretical concepts of stability test** are verified by **laboratory simulations** and **practical experiments**.

## Future of work:

- **Stability test** for **parasitic components** in transmission lines, printed circuit boards, physical layout layers

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Thank you very much!



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